16-Channel Short Haul T1/E1/J1 Line Interface Unit IDT82P20416

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16-Channel Short Haul T1/E1/J1 Line Interface Unit

FEATURES

- Integrates 16 channels T1/E1/J1 short haul line interface units for 100 Ω T1, 120 Ω E1, 110 Ω J1 twisted pair cable and 75 Ω E1 coaxial cable applications
- Per-channel configurable Line Interface options
 - Fully integrated and software selectable receive and transmit termination
 - Option 1: Fully Internal Impedance Matching with integrated receive termination resistor
 - Option 2: Partially Internal Impedance Matching with common external resistor for improved device power dissipation
 - Option 3: External impedance Matching termination
 - Supports global configuration and per-channel configuration to T1, E1 or J1 mode

Per-channel programmable features

- Provides T1/E1/J1 short haul waveform templates and userprogrammable arbitrary waveform templates
- Provides two JAs (Jitter Attenuator) for each channel of receiver and transmitter
- Supports AMI/B8ZS (for T1/J1) and AMI/HDB3 (for E1) encoding and decoding

Per-channel System Interface options

- Supports Single Rail, Dual Rail with clock or without clock and sliced system interface
- Integrated Clock Recovery for the transmit interface to recover transmit clock from system transmit data

• Per-channel system and diagnostic functions

- Provides transmit driver over-current detection and protection with optional automatic high impedance of transmit interface
- Detects and generates PRBS (Pseudo Random Bit Sequence), ARB (Arbitrary Pattern) and IB (Inband Loopback) in either receive or transmit direction
- Provides defect and alarm detection in both receive and transmit directions.
 - Defects include BPV (Bipolar Violation) /CV (Code Violation) and EXZ (Excessive Zeroes)
 - Alarms include LLOS (Line LOS), SLOS (System LOS), TLOS (Transmit LOS) and AIS (Alarm Indication Signal)
- Programmable LLOS detection /clear levels. Compliant with ITU and ANSI specifications

- Various pattern, defect and alarm reporting options – Serial hardware LLOS reporting (LLOS, LLOS0) for all 16 channels
 - Register access to individual registers or 16-bit error counters
- Supports Analog Loopback, Digital Loopback and Remote Loopback
- Supports T1.102 line monitor
- Hitless Protection Switching (HPS) without external Relays
 - Supports 1+1 and 1:1 hitless protection switching
 - Asynchronous hardware control (OE, RIM) for fast global high impedance of receiver and transmitter (hot switching between working and backup board)
 - · High impedance transmitter and receiver while powered down
 - Per-channel register control for high impedance, independent for receiver and transmitter

Clock Inputs and Outputs

- + Flexible master clock (N x 1.544 MHz or N x 2.048 MHz) (1 \leq N \leq 8, N is an integer number)
- Integrated clock synthesizer can multiply or divide the reference clock to a wide range of frequencies: 8 KHz, 64 KHz, 2.048 MHz, 4.096 MHz, 8.192 MHz, 19.44 MHz and 32.768 MHz

Microprocessor Interface

· Supports Serial microprocessor interface

• Other Key Features

- IEEE1149.1 JTAG boundary scan
- Two general purpose I/O pins
- · 3.3 V I/O with 5 V tolerant inputs
- 3.3 V and 1.8 V power supply
- Package: 484-pin Fine Pitch BGA (19 mm X 19 mm)

Applicable Standards

- AT&T Pub 62411 Accunet T1.5 Service
- ANSI T1.102, T1.403 and T1.231
- · Bellcore TR-TSY-000009, GR-253-CORE and GR-499-CORE
- ETSI CTR12/13
- ETS 300166 and ETS 300 233
- G.703, G.735, G.736, G.742, G.772, G.775, G.783 and G.823
- 0.161
- ITU I.431 and ITU O.171

APPLICATIONS

- SDH/SONET multiplexers
- Central office or PBX (Private Branch Exchange)
- Digital access cross connects
- Remote wireless modules
- Microwave transmission systems

DESCRIPTION

The IDT82P20416 is a 16-channel high-density T1/E1/J1 short haul Line Interface Unit. Each channel of the IDT82P20416 can be independently configured. The configuration is performed through a Serial microprocessor interface.

In the receive path, through a Single Ended or Differential line interface, the received signal is processed by an adaptive Equalizer and then sent to a Slicer. Clock and data are recovered from the digital pulses output from the Slicer. After passing through an enabled or disabled Receive Jitter Attenuator, the recovered data is decoded using B8ZS/ AMI/HDB3 line code rule in Single Rail NRZ Format mode and output to the system, or output to the system without decoding in Dual Rail NRZ Format mode and Dual Rail RZ Format mode. In the transmit path, the data to be transmitted is input on TDn in Single Rail NRZ Format mode or TDPn/TDNn in Dual Rail NRZ Format mode and Dual Rail RZ Format mode, and is sampled by a transmit reference clock. The clock can be supplied externally from TCLKn or recovered from the input transmit data by an internal Clock Recovery. A selectable JA in Tx path is used to de-jitter gapped clocks. To meet T1/ E1/J1 waveform standards, five preset T1 templates, two E1 templates and one J1 template, as well as an arbitrary waveform generator are provided. The data through the Waveform Shaper, the Line Driver and the Tx Transmitter is output on TTIPn and TRINGn.

Alarms (including LOS, AIS) and defects (including BPV, EXZ) are detected in both receive line side and transmit system side. AIS alarm, PRBS, ARB and IB patterns can be generated /detected in receive / transmit direction for testing purpose. Analog Loopback, Digital Loopback and Remote Loopback are all integrated for diagnostics.

A line monitor function per T1.102 is available to provide a Non-Intrusive Monitoring of channels of other devices.

JTAG per IEEE 1149.1 is also supported by the IDT82P20416.

BLOCK DIAGRAM

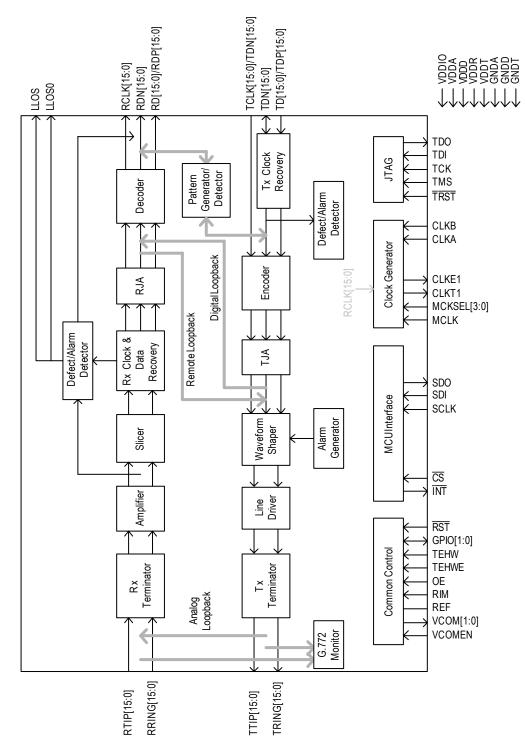


Figure-1 Functional Block Diagram

1 PIN ASSIGNMENT

r	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	-
A	NC	NC	NC	NC	RD15/ RDP15	(TDN14	RCLK1 4	TDN13	RCLK1 3	TDN12	RCLK1 2	RCLK1	TDN11	RCLK1	(TDN10)	RCLK9	TDN9	RCLK8	NC	NC	NC	NC	A
В	NC	NC	NC	NC	TCLK15 /TDN15	TD14/ TDP14	(RDN14)	TD13/ TDP13	RDN13	TD12/ TDP12	RDN12	RDN11	TD11/ TDP11	RDN10	TD10/ TDP10	RDN9	TD9/ TDP9	RD8/ RDP8	NC	NC	NC	NC	В
С	TRING 12	NC	NC	NC	TD15/ TDP15	GNDD	GNDD	GNDD	GNDD	GNDD	GNDD	GNDD	GNDD	GNDD	GNDD	GNDD	GNDD	(TDN8	NC	NC	NC	NC	c
D	(TTIP12)	NC	NC	NC	TDN15	RDN15	TCLK14 /TDN14	VDDIO	TCLK13 /TDN13	GNDD	TCLK12 /TDN12		VDDIO	TCLK10 /TDN10	VDDIO	TCLK9/ TDN9	VDDIO	TD8/ TDP8	NC	NC	NC	NC	D
Е	TRING 13	GNDA	GNDA	GNDA	NC	RCLK1 5	RD14/ RDP14	GNDD	RD13/ RDP13	VDDIO	RD12/ RDP12	RD11/ RDP11	GNDD	RD10/ RDP10		RD9/ RDP9	RDN8	TCLK8/ TDN8	GNDA	GNDA	NC	TRING 11	E
F	(TTIP13)	NC	GNDA	GNDA	VDDD			VDDD	NC	VDDD	GNDD	VDDIO	VDDD	VDDD	VDDD	VDDD	VDDD	VDDD	GNDA	GNDA	NC	TTIP11	F
G	(TRING 14	NC	RTIP12	RRING 12				GNDD		GNDD	VDDD	GNDD		GNDD				NC	RRING	RTIP11	NC	TRING 10	G
н	(TTIP14)	NC		(RRING 13		VDDR1		GNDD		GNDD	GNDD	GNDD		GNDD				NC	RRING	RTIP10	NC	TTIP10	H
J	(TRING 15	NC	RTIP14	(RRING 14	VDDT	VDDT	GNDT	GNDT	GNDT	GNDT	GNDT	GNDT	GNDT	GNDT					RRING 9	RTIP9	NC) J
К	(TTIP15)	NC	RTIP15	(RRING 15	VDDT	VDDT				GNDT	GNDT	GNDT								RTIP8	NC	TTIP9	ĸ
L						VDDR1		GNDT	VDDT	GNDT	GNDT	GNDT					VDDR8	NC			NC	TRING 8	L
М								GNDT					GNDA					VDDA					M
Ν				(RRING 2			(VDDR3)	GNDT	GNDT	GNDT				NC					RRING	RTIP7		TRING 7	N
Р			(RTIP3	(RRING 3						GNDD	NC			NC			VDDR5	NC		RTIP6	NC		P
R	(TRING 2			(RRING 4				NC											RRING	RTIP5		TRING 6	R
Т					NC	NC		VDDD															Т
U	(TRING 3				NC															GNDA	NC	TRING 5	U
V				(TD1/ TDP1		(TCLK2/) TDN2	TCLK3/ TDN3	TCLK4/ TDN4		GPIO0		TEHWE		ССКВ	MCK SEL3	TCLK6/ TDN6		RDN7	TD6/ TDP6				v
W	(TRING 4		TCLK1/ TDN1			(TD2/ TDP2		RDN3	RDN4	(TCLK0/) TDN0		ТСК	SDI	NC			TCLK5/ TDN5	RCLK7			NC	NC	w
Y	(TTIP4)	RD1/ RDP1						NC	NC			NC		NC					NC		(NC		Y
AA	NC	RD2/ RDP2		(TD3/ TDP3	RD3/ RDP3	TD4/ TDP4		RD0/ RDP0		GPI01	RST	SDO				TD5/ TDP5	RCLK5	RD6/ RDP6		TCLK7/ TDN7	RD7/ RDP7	, NC	AA
AB		RCLK3		RD4/ RDP4	(TD0/ TDP0		RCLK0	TRST	TDO	TEHW	RIM	SCLK		LLOS	NC			MCLK		RD5/ RDP5		RCLK6	AB
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	2	1 22	

Figure-2 484-Pin Fine Pitch BGA (Top View)

2 PIN DESCRIPTION

Name	I/O	Pin No. ¹	Description					
Line Interface								
RTIPn RRINGn (n=0~15)	Input	L3, M3, N3, P3, R3, R20, P20, N20, K20, J20, H20, G20, G3, H3, J3, K3 L4, M4, N4, P4, R4, R19, P19, N19, K19, J19, H19, G19, G4, H4, J4, K4	mode. In Receive Differential mode, the received signal is coupled into RTIPn and RRINGn via a 1:1					
TTIPn TRINGn (n=0~15)	Output	M1, P1, T1, V1, Y1, V22, T22, P22, M22, K22, H22, F22, D1, F1, H1, K1 L1, N1, R1, U1, W1, U22, R22, N22, L22, J22, G22, E22, C1, E1, G1, J1	 TTIPn / TRINGn: Transmit Bipolar Tip /Ring for Channel 0 ~ 15 The transmit line interface supports both Transmit Differential mode and Transmit Single Ended mode. In Transmit Differential mode, TTIPn outputs a positive differential pulse while TRINGn outputs a negative differential pulse. The pulses are coupled to the line side via a 1:2 (step up) transformer or without a transformer (transformer-less). In Transmit Single Ended mode, TRINGn should be left open (it is shorted to ground internally). The signal presented at TTIPn is output to the line side via a 1:2 (step up) transformer. These pins will become High-Z globally or channel specific in the following conditions: Global High-Z: Connecting the OE pin to low; Loss of MCLK; During and after power-on reset, hardware reset or global software reset; Per-channel High-Z Writing '0' to the OE bit (b6, TCF0,)²; Loss of TCLKn in Transmit Single Rail NRZ Format mode or Transmit Dual Rail NRZ Format mode, except that the channel is in Remote Loopback or transmit internal pattern with XCLK³; Transmitter power down by writing '1' to the T_OFF bit (b5, TCF0,); Per-channel software reset; The THZ_OC bit (b4, TCF0,) is set to '1' and the transmit driver over-current is detected. 					

Note:

1. The pin number of the pins with the footnote 'n' is listed in order of channel (CH0 ~ CH15).

2. The content in the brackets indicates the position and the register name of the preceding bit. After the register name, if the punctuation ',...' is followed, this bit is in a per-channel register. The addresses and details are included in Chapter 5 Programming Information.

3. XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.



Name	I/O	Pin No.	Description
			System Interface
RDn / RDPn (n=0~15)	Output	AA8, Y2, AA2, AA5, AB4, AB20, AA18, AA21, B18, E16, E14, E12, E11, E9, E7, A5	 RDn: Receive Data for Channel 0 ~ 15 When the receive system interface is configured to Single Rail NRZ Format mode, this multiplex pin is used as RDn. The decoded NRZ data is updated on the active edge of RCLKn. The active level on RDn is selected by the RD_INV bit (b3, RCF1,). When the receiver is powered down, RDn will be in High-Z state or low, as selected by the RHZ bit (b6, RCF0,). RDPn: Positive Receive Data for Channel 0 ~ 15 When the receive system interface is configured to Dual Rail NRZ Format mode, Dual Rail RZ Format mode or Dual Rail Sliced mode, this multiplex pin is used as RDPn. In Receive Dual Rail Sliced mode, the un-decoded NRZ data is output on RDPn and RDNn and updated on the active edge of RCLKn. In Receive Dual Rail RZ Format mode, the un-decoded RZ data is output on RDPn and RDNn and updated on the active edge of RCLKn. In Receive Dual Rail Sliced mode, the raw RZ sliced data is output on RDPn and RDNn. For Receive Differential line interface, an active level on RDPn indicates the receipt of a positive pulse on RTIPn and a negative pulse on RRINGn; while an active level on RDNn indicates the receipt of a positive pulse on RTIPn. For Receive Single Ended line interface, an active level on RDPn indicates the receipt of a positive pulse on RTIPn; while an active level on RDPn indicates the receipt of a positive pulse on RTIPn; while an active level on RDPn indicates the receipt of a negative pulse on RTIPn. The active level on RDPn and RDNn is selected by the RD_INV bit (b3, RCF1,). When the receiver is powered down, RDPn and RDNn will be in High-Z state or low, as
RDNn (n=0~15)	Output	V9, V5, W7, W8, W9, W19, T15, V18, E17, B16, B14, B12, B11, B9, B7, D6	selected by the RHZ bit (b6, RCF0,). RDNn: Negative Receive Data for Channel 0 ~ 15 When the receive system interface is configured to Dual Rail NRZ Format mode, Dual Rail RZ Format mode or Dual Rail Sliced mode, this multiplex pin is used as RDNn. (Refer to the description of RDPn for details).
RCLKn (n=0~15)	Output	AB7, W4, AA3, AB2, AA7, AA17, AB22, W18, A18, A16, A14, A12, A11, A9, A7, E6	RCLKn: Receive Clock for Channel 0 ~ 15 When the receive system interface is configured to Single Rail NRZ Format mode, Dual Rail NRZ Format mode or Dual Rail RZ Format mode, this multiplex pin is used as RCLKn. RCLKn outputs a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock which is recovered from the received signal. The data output on RDPn/RDNn (in Receive Dual Rail NRZ Format mode, Receive Dual Rail RZ Format mode and Receive Dual Rail Sliced) is updated on the active edge of RCLKn. The active edge is selected by the RCK_ES bit (b4, RCF1,). In LLOS condition, RCLKn output high or XCLK, as selected by the RCKH bit (b7, RCF0,) (refer to Section 3.5.3.1 Line LOS (LLOS) for details). When the receiver is powered down, RCLKn will be in High-Z state or low, as selected by the RHZ bit (b6, RCF0,).
LLOS	Output	AB14	LLOS: Receive Line Loss Of Signal LLOS synchronizes with the output of CLKE1 and can indicate the LLOS (Line LOS) status of all 16 channels in a serial format. When the clock output on CLKE1 is enabled, LLOS indicates the LLOS status of the 16 chan- nels in a serial format and repeats every seventeen cycles. The start filler is positioned by LLOS0. Refer to the description of LLOS0 below for details. LLOS is updated on the rising edge of CLKE1 and is always active high. When the clock output of CLKE1 is disabled, LLOS will be held in High-Z state. (Refer to Section 3.5.3.1 Line LOS (LLOS) for details.)

Name	I/O	Pin No.			Description	
LLOSO	Output	AA13	LLOS0 can indic When the clock of indicate the start high for one 8 K outputs 2.048 M seventeen 2.048 When the clock of	ate the start posi putput on CLKE1 position on the I Hz clock cycle (` Hz clock, LLOS0 MHz clock cycle putput on CLKE1	LOS pin. When CLKE1 out 125 µs) every seventeen 8	
TDn / TDPn (n=0~15)	Input	AB5, V4, W6, AA4, AA6, AA16, V19, V17, D18, B17, B15, B13, B10, B8, B16, C5	plex pin is used a TDn accepts Sin TCLKn. The active level of TDPn: Positive When the transm RZ Format mode In Transmit Dual TDNn and samp In Transmit Dual	nit system interfa as TDn. gle Rail NRZ dat on TDn is selecte Transmit Data f nit system interfa e, this multiplex p I Rail NRZ Form led on the active Rail RZ Format i	ce is configured to Single R a. The data is sampled into ed by the TD_INV bit (b3, To or Channel 0 ~ 15 ce is configured to Dual Ra in is used as TDPn. at mode, the pre-encoded edge of TCLKn.	il NRZ Format mode or Dual Rail NRZ data is input on TDPn and data is input on TDPn and TDNn.
			TDPn	TDNn	Output Pulse on TTIPn	Output Pulse on TRINGn *
			0	0	Space	Space
			0	1	Negative Pulse	Positive Pulse
			1	0	Positive Pulse	Negative Pulse
			1	1	Space	Space
			Note: * For Transmit :	Single Ended line i	nterface, TRINGn should be op	en.
			The active level	on TDPn and TD	Nn is selected by the TD_I	NV bit (b3, TCF1,).
TDNn (n=0~15)	Input / Output	AB6, Y3, W5, AB1, AB3, AB19, AB21, AA19, C18, A17, A15, A13, A10, A8, A6, D5		nit system interfa as TDNn.		ail NRZ Format mode, this multi-
TCLKn / TDNn (n=0~15)	Input	W10, W3, V6, V7, V8, W17, V16, AA20, E18, D16, D14, D12, D11, D9, D7, B5	Rail NRZ Format TCLKn inputs a The data input o mit Dual Rail NR	nit system interfa t mode, this multi 1.544 MHz (in T1 n TDn (in Transr Z Format mode) • Transmit Data • it system interfa DNn.	ace is configured to Single iplex pin is used as TCLKn. /J1 mode) or 2.048 MHz (ir nit Single Rail NRZ Format is sampled on the active ec for Channel 0 ~ 15 ce is configured to Dual Rai	mode) or TDPn/TDNn (in Trans-



Name	1/0	Pin No.	Description				
	· ·		Clock				
MCLK MCKSEL[0]	Input	AB18 AB17	MCLK: Master Clock Input MCLK provides a stable reference timing +/-32 ppm (in T1/J1 mode) or +/-50 ppm is informed to the device by MCKSEL[3: If MCLK misses (duty cycle is less than reset automatically. MCKSEL[3:0]: Master Clock Selection	(in E1 mode) accuracy. The clock frequ)]. 30% for 10 μs) and then recovers, the	ency of MCL		
	input		These four pins inform the device of the				
MCKSEL[1]		W16	MCKSEL[3:0] [*]	Frequency (MHz)			
MCKSEL[2]		AA15	0000	1.544			
MCKSEL[3]		V15	0001	1.544 X 2			
			0010	1.544 X 3			
			0011	1.544 X 4			
			0100	1.544 X 5			
			0101	1.544 X 6			
			0110	1.544 X 7			
			0111	1.544 X 8			
			1000	2.048			
			1001	2.048 X 2			
			1010	2.048 X 3			
			1011	2.048 X 4			
			1100	2.048 X 5			
			1101	2.048 X 6			
			1110	2.048 X 7			
			1111	2.048 X 8			
			Note: 0: GNDD 1: VDDIO				
CLKT1	Output	W15	CLKT1: 8 KHz / T1 Clock Output The output on CLKT1 can be enabled or disabled, as determined by the CLKT1_EN bit (b1, CLKG). When the output is enabled, CLKT1 outputs an 8 KHz or 1.544 MHz clock, as selected by the CLKT1 bit (b0, CLKG). The output is locked to MCLK. When the output is disabled, CLKT1 is in High-Z state				
CLKE1	Output	AB16	When the output is disabled, CLKT1 is in High-Z state. CLKE1: 8 KHz / E1 Clock Output The output on CLKE1 can be enabled or disabled, as determined by the CLKE1_EN bit (b3, CLKG). When the output is enabled, CLKE1 outputs an 8 KHz or 2.048 MHz clock, as selected by the CLKE1 bit (b2, CLKG). The output is locked to MCLK. When the output is disabled, CLKE1 is in High-Z state.				

Name	I/O	Pin No.	Description
CLKA	Input	AA14	CLKA: External T1/E1 Clock Input A External T1/J1 (1.544 MHz) or E1 (2.048 MHz) clock is input on this pin. When not used, this pin should be connected to GNDD.
CLKB	Input	V14	CLKB: External T1/E1 Clock Input B External T1/J1 (1.544 MHz) or E1 (2.048 MHz) clock is input on this pin. When not used, this pin should be connected to GNDD.
			Common Control
VCOM[0] VCOM[1]	Output	M20 M19	VCOM: Voltage Common Mode [1:0] These pins are used only when the receive line interface is in Receive Differential mode and connected without a transformer (transformer-less). To enable these pins, the VCOMEN pin must be connected high. Refer to Figure-6 for the connection. When these pins are not used, they should be left open.
VCOMEN	Input (Pull-Down)	L19	VCOMEN: Voltage Common Mode Enable This pin should be connected high only when the receive line interface is in Receive Differen- tial mode and connected without a transformer (transformer-less). When not used, this pin should be left open.
REF	-	L20	REF: Reference Resistor An external resistor (10 K Ω , ±1%) is used to connect this pin to ground to provide a standard reference current for internal circuit. This resistor is required to ensure correct device operation.
RIM	Input (Pull-Down)	AB11	RIM: Receive Impedance Matching In Receive Differential mode, when RIM is low, all 16 receivers become High-Z and only external impedance matching is supported. In this case, the per-channel impedance matching configuration bits - the R_TERM[2:0] bits (b2~0, RCF0,) and the R120IN bit (b4, RCF0,) - are ignored. In Receive Differential mode, when RIM is high, impedance matching is configured on a per-channel basis by the R_TERM[2:0] bits (b2~0, RCF0,) and the R120IN bit (b4, RCF0,). This pin can be used to control the receive impedance state for Hitless Protection applications. Refer to Section 4.4 Hitless Protection Switching (HPS) Summary for details. In Receive Single Ended mode, this pin should be left open.
OE	Input	V11	OE: Output Enable OE enables or disables all Line Drivers globally. A high level on this pin enables all Line Drivers while a low level on this pin places all Line Drivers in High-Z state and independent from related register settings. Note that the functionality of the internal circuit is not affected by OE. If this pin is not used, it should be tied to VDDIO. This pin can be used to control the transmit impedance state for Hitless protection applica- tions. Refer to Section 4.4 Hitless Protection Switching (HPS) Summary for details.
TEHWE	Input (Pull-Up)	V12	TEHWE: Hardware T1/J1 or E1 Mode Selection Enable When this pin is open, the T1/J1 or E1 operation mode is selected by TEHW globally. When this pin is low, the T1/J1 or E1 operation mode is selected by the T1E1 bit (b0 CHCF,) on a per-channel basis.
TEHW	Input (Pull-Up)	AB10	TEHW: Hardware T1/J1 or E1 Mode Selection When TEHWE is open, this pin selects the T1/J1 or E1 operation mode globally: Low - E1 mode; Open - T1/J1 mode. When TEHWE is low, the input on this pin is ignored.



Name	I/O	Pin No.	Description
GPIO[0] GPIO[1]	Output / Input	V10 AA10	GPIO: General Purpose I/O [1:0] These two pins can be defined as input pins or output pins by the DIR[1:0] bits (b1~0, GPIO) respectively. When the pins are input, their polarities are indicated by the LEVEL[1:0] bits (b3~2, GPIO) respectively. When the pins are output, their polarities are controlled by the LEVEL[1:0] bits (b3~2, GPIO) respectively.
RST	Input	AA11	RST: Reset (Active Low) A low pulse on this pin resets the device. This hardware reset process completes in 2 µs maximum. Refer to Section 4.1 Reset for an overview on reset options.
			MCU Interface
INT	Output	AB13	INT: Interrupt Request This pin indicates interrupt requests for all unmasked interrupt sources. The output characteristics (open drain or push-pull internally) and the active level are deter- mined by the INT_PIN[1:0] bits (b3~2, GCF).
CS	Input	V13	\overline{CS} : Chip Select (Active Low) This pin must be asserted low to enable the microprocessor interface. A transition from high to low must occur on this pin for each Read/Write operation and \overline{CS} should remain low until the operation is over.
SCLK	Input	AB12	SCLK: Shift Clock In Serial microprocessor interface, this multiplex pin is used as SCLK. SCLK inputs the shift clock for the Serial microprocessor interface. Data on SDI is sampled by the device on the rising edge of SCLK. Data on SDO is updated on the falling edge of SCLK.
SDI	Input	W13	SDI: Serial Data Input In Serial microprocessor interface, this multiplex pin is used as SDI. Address and data on this pin are serially clocked into the device on the rising edge of SCLK.
SDO	Output	AA12	SDO: Serial Data Output In Serial microprocessor interface, this multiplex pin is used as SDO. Data on this pin is serially clocked out of the device on the falling edge of SCLK.
			JTAG (per IEEE 1149.1)
TRST	Input Pull-Down	AB8	TRST: JTAG Test Reset (Active Low)A low signal on this pin resets the JTAG test port. To ensure deterministic operation of the teslogic, TMS should be held high when the signal on TRST changes from low to high.This pin may be left unconnected when JTAG is not used.This pin has an internal pull-down resistor.
TMS	Input Pull-up	W11	TMS: JTAG Test Mode SelectThe signal on this pin controls the JTAG test performance and is sampled on the rising edgeof TCK. To ensure deterministic operation of the test logic, TMS should be held high when thesignal on TRST changes from low to high.This pin may be left unconnected when JTAG is not used.This pin has an internal pull-up resistor.
ТСК	Input	W12	TCK: JTAG Test Clock The clock for the JTAG test is input on this pin. TDI and TMS are sampled on the rising edge of TCK and TDO is updated on the falling edge of TCK. When TCK is idle at low state, all stored-state devices contained in the test logic shall retain their state indefinitely. This pin should be connected to GNDD when JTAG is not used.



Name	1/0	Pin No.	Description
TDI	Input Pull-up	AA9	TDI: JTAG Test Data Input The test data is input on this pin. It is clocked into the device on the rising edge of TCK. This pin has an internal pull-up resistor. This pin may be left unconnected when JTAG is not used.
TDO	Output	AB9	TDO: JTAG Test Data Output The test data is output on this pin. It is clocked out of the device on the falling edge of TCK. TDO is a High-Z output signal except during the process of data scanning.
	·	·	Power & Ground
VDDIO		D8, D13, D15, D17, E10, F12, P13, R10, R11, R16, T7	VDDIO: 3.3 V I/O Power Supply
VDDA		N21, M12, N12, M18	VDDA: 3.3 V Analog Core Power Supply
VDDD		F5, F8, F10, F13, F14, F15, F16, F17, F18, G5, G6, G11, R12, R14, R15, T8, T9, T16, U8, U9	VDDD: 1.8 V Digital Core Power Supply
VDDRn (N=0~15)		H6, J16, K8, K9, K13, K15, L5, L6, L17, M7, M9, M16, N7, N16, P8, P17	VDDRn: 3.3 V Power Supply for Receiver
VDDT		J5, J6, J17, J18, K5, K6, K17, K18, L7, L9, L13, L16, M5, M6, M10, M17, N5, N6, N13, N15, N17, N18, P5, P6, R5, R6, T18, T19	VDDT: 3.3 V Power Supply for Transmitter Driver
GNDA		E2, E3, E4, E19, E20, F3, F4, F19, F20, M13, M21, T3, T4, T20, U3, U4, U19, U20, V3, V20, V21	GNDA: GND for Analog Core / Receiver
GNDD		C6, C7, C8, C9, C10, C11, C12, C13, C14, C15, C16, C17, D10, E8, E13, E15, F6, F7, F11, G7, G8, G9, G10, G12, G13, G14, G15, G16, G17, H7, H8, H9, H10, H11, H12, H13, H14, H15, H16, H17, P9, P10, P12, P15, R9, R13, R17, T10, T11, T12, T13, T14, T17, U10, U11, U12, U13, U14, U15, U16, U17, Y20	GNDD: Digital GND
GNDT		J7, J8, J9, J10, J11, J12, J13, J14, J15, K7, K10, K11, K12, K14, K16, L8, L10, L11, L12, L14, L15, M8, M11, M14, M15, N8, N9, N10, N11, P7, P16	GNDT: Analog GND for Transmitter Driver



Name	I/O	Pin No.	Description										
	TEST												
NC	-	A1, A2, A3, A4, A19, A20, A21, A22, B1, B2, B3, B4, B19, B20, B21, B22, C2, C3, C4, C19, C20, C21, C22, D2, D3, D4, D19, D20, D21, D22, E5, E21, F2, F9, F21, G2, G18, G21, H2, H5, H18, H21, J2, J21, K2, K21, L2, L18, L21, M2, N2, N14, P2, P11, P14, P18, P21, R2, R7, R8, R18, R21, T2, T5, T6, T21, U2, U5, U6, U7, U18, U21, V2, W2, W21, W22, Y4, Y5, Y6, Y7, Y8, Y9, Y10, Y11, Y12, Y13, Y14, Y15, Y16, Y17, Y18, Y19, Y21, Y22, AA1, AA22, AB15, W14	NC: No Connected These pins should be left open.										
			Others										
IC		W20	IC: Internal Connected This pin is for IDT use only and should be connected to GNDD.										

3 FUNCTIONAL DESCRIPTION

3.1 T1 / E1 / J1 MODE SELECTION

The IDT82P20416 can be configured to T1/J1 mode or E1 mode globally or on a per-channel basis. The configuration is determined by the TEHWE pin, the TEHW pin and the T1E1 bit (b0, CHCF,...). Refer to Table-1 for details of the operation mode selection.

Table-1 Operation Mode Selection

	Global Prog	ramming	Per-Channel Programming					
TEHWE Pin	Оре	n	Low					
TEHW Pin	Open	Low	(The configuration of this pin is ignored)					
T1E1 Bit	(The configuration of	this bit is ignored).	0	1				
Operation Mode	T1/J1	E1	T1/J1 E1					

3.2 RECEIVE PATH

3.2.1 R_X TERMINATION

The receive line interface supports Receive Differential mode. In Receive Differential mode, both RTIPn and RRINGn are used to receive signal from the line side.

In Receive Differential mode, the line interface can be connected with T1 100 Ω , J1 110 Ω or E1 120 Ω twisted pair cable or E1 75 Ω coaxial cable.

The receive impedance matching is realized by using internal impedance matching or external impedance matching for each channel in different applications.

3.2.1.1 Receive Differential Mode

In Receive Differential mode, three kinds of impedance matching are supported: Fully Internal Impedance Matching, Partially Internal Impedance Matching and External Impedance Matching. Figure-3 shows an overview of how these Impedance Matching modes are switched.

Fully Internal Impedance Matching circuit uses an internal programmable resistor (IM) only and does not use an external resistor. This configuration saves external components and supports 1:1 Hitless Protection Switching (HPS) applications without relays. Refer to Section 4.4 Hitless Protection Switching (HPS) Summary.

Partially Internal Impedance Matching circuit consists of an internal programmable resistor (IM) and a value-fixed 120 Ω external resistor (Rr). Compared with Fully Internal Impedance Matching, this configuration provides considerable savings in power dissipation of the device. For example, In E1 120 Ω PRBS mode, the power savings would be 0.44 W. For power savings in other modes, please refer to Chapter 8 Physical And Electrical Specifications.

External Impedance Matching circuit uses an external resistor (Rr) only.

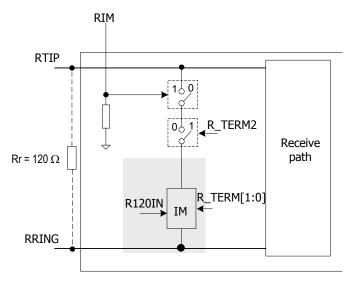


Figure-3 Switch between Impedance Matching Modes

To support some particular applications, such as hot-swap or Hitless Protection Switch (HPS) hot-switchover, RTIPn/RRINGn must be forced to enter high impedance state (i.e., External Impedance Matching). For hot-swap, RTIPn/RRINGn must be always held in high impedance state during /after power up; for HPS hot-switchover, RTIPn/RRINGn must enter high impedance state immediately after switchover. Though each channel can be individually configured to External Impedance Matching through register access, it is too slow for hitless switch. Therefore, a hardware pin - RIM - is provided to globally control the high impedance for all 16 receivers.

When RIM is low, only External Impedance Matching is supported for all 16 receivers and the per-channel impedance matching configuration bits - the R_TERM[2:0] bits (b2~0, RCF0,...) and the R120IN bit (b4, RCF0,...) - are ignored.

When RIM is high, impedance matching is configured on a perchannel basis. Three kinds of impedance matching are all supported and selected by the R_TERM[2:0] bits (b2~0, RCF0,...) and the R120IN bit (b4, RCF0,...). The R_TERM[2] bit (b2, RCF0,...) should be set to match internal or external impedance. If the R_TERM[2] bit (b2, RCF0,...) is '0', internal impedance matching is enabled. The R120IN bit (b4, RCF0,...) should be set to select Partially Internal Impedance Matching or Fully Internal Impedance Matching. The internal programmable resistor (IM) is determined by the R_TERM[1:0] bits (b1~0, RCF0,...). If the R_TERM[2] bit (b2, RCF0,...) is '1', external impedance matching is enabled. The configuration of the R120IN bit (b4, RCF0,...) and the R_TERM[1:0] bits (b1~0, RCF0,...) is ignored. A twisted pair cable can be connected with a 1:1 transformer or without a transformer (transformer-less), while a coaxial cable must be connected with a 1:1 transformer. Table 2 lists the recommended impedance matching value in different applications. Figure-4 to Figure-6 show the connection for one channel.

The transformer-less connection will offer a termination option with reduced cost and board space. However, the waveform amplitude is not standard compliant, and surge protection and common mode depression should be enhanced depending on equipment environment.

Cable Condition	Partially Internal Im (R120IN	• •	Fully Internal Imper (R120IN =	-	External Impedance Matching		
	R_TERM[2:0]	Rr	R_TERM[2:0]	Rr	R_TERM[2:0] ³	Rr	
T1 100 Ω twisted pair (with transformer)	000		000			100 Ω	
J1 110 Ω twisted pair (with transformer)	001	-	001	(open)	1XX	110 Ω	
E1 120 Ω twisted pair (with transformer)	010		010			120 Ω	
E1 75 Ω coaxial (with transformer)	011	120 Ω	011			75 Ω	
T1 100 Ω twisted pair (transformer-less ⁴)	000					100 Ω	
J1 110 Ω twisted pair (transformer-less)	001		(not supp	orted)		110 Ω	
E1 120 Ω twisted pair (transformer-less)	010					120 Ω	
Note:	•	1			1		

Note:

1. Partially Internal Impedance Matching and Fully Internal Impedance Matching are not supported when RIM is low.

2. Fully Internal Impedance Matching is not supported in transformer-less applications.

3. When RIM is low, the setting of the R_TERM[2:0] bits is ignored.

4. In transformer-less applications, the device should be protected against overvoltage. There are three important standards for overvoltage protection:

• UL1950 and FCC Part 68;

• Telcordia (Bellcore) GR-1089

• ITU-T K.20, K.21 and K.41

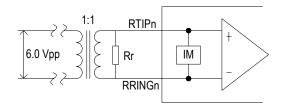


Figure-4 Receive Differential Line Interface with Twisted Pair Cable (with transformer)

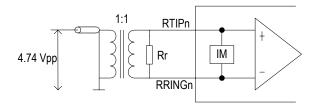
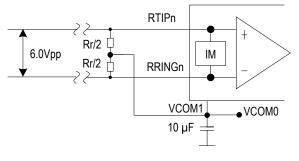


Figure-5 Receive Differential Line Interface with Coaxial Cable (with transformer)



Note: 1. Two Rr/2 resistors should be connected to VCOM[1:0] that are coupled to ground via a 10 μF capacitor, which provide 60 Ω common mode input resistance.

2. In this mode, lightning protection should be enhanced.

3. The maximum input dynamic range of RTIP/TRING pin is -0.3 V ~3.6 V (in line monitor mode it is -0.3 V ~ 2 V)

Figure-6 Receive Differential Line Interface with Twisted Pair Cable (transformer-less, non standard compliant)

3.2.2 EQUALIZER

The equalizer compensates high frequency attenuation to enhance receive sensitivity.

3.2.2.1 Line Monitor

In both T1/J1 and E1 short haul applications, the Protected Non-Intrusive Monitoring per T1.102 can be performed between two devices. The monitored channel of one device is in normal operation, and the monitoring channel of the other device taps the monitored one through a high impedance bridging circuit (refer to Figure-7 and Figure-8).

After the high resistance bridging circuit, the signal arriving at RTIPn/ RRINGn of the monitoring channel is dramatically attenuated. To compensate this bridge resistive attenuation, Monitor Gain can be used to boost the signal by 20 dB, 26 dB or 32 dB, as selected by the MG[1:0] bits (b1~0, RCF2,...). For normal operation, the Monitor Gain should be set to 0 dB, i.e., the Monitor Gain of the monitored channel should be 0 dB.

The monitoring channel can be configured to any of the External, Partially Internal or Fully Internal Impedance Matching mode. Here the external r or internal IM is used for voltage division, not for impedance matching. That is, the r (IM) and the two R make up of a resistance bridge. The resistive attenuation of this bridge is 20lg(r/(2R+r)) dB.

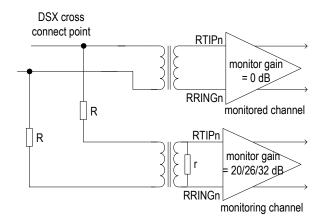
Note that line monitor is only available in differential line interface.

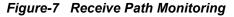
3.2.2.2 Receive Sensitivity

The receive sensitivity is the minimum range of receive signal level for which the receiver recovers data error-free with -18 dB interference signal added.

For Receive Differential line interface, the receive sensitivity is -15 dB.

For Receive Single Ended line interface, the receive sensitivity is -12 dB.





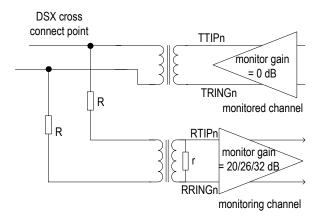


Figure-8 Transmit Path Monitoring

3.2.3 SLICER

The Slicer is used to generate a standard amplitude mark or a space according to the amplitude of the input signals. The input signal is sliced at 50% of the peak value.

3.2.4 R_x CLOCK & DATA RECOVERY

The Rx Clock & Data Recovery is used to recover the clock signal from the received data. It is accomplished by an integrated Digital Phase Locked Loop (DPLL). The recovered clock tracks the jitter in the data output from the Slicer and keeps the phase relationship between data and clock during the absence of the incoming pulse.

3.2.5 DECODER

The Decoder is used only when the receive system interface is in Single Rail NRZ Format mode. When the receive system interface is in other modes, the Decoder is bypassed automatically. (Refer to Section 3.2.6 Receive System Interface for the description of the receive system interface).

In T1/J1 mode, the received signal is decoded by AMI or B8ZS line code rule. In E1 mode, the received signal is decoded by AMI or HDB3 line code rule. The line code rule is selected by the R_CODE bit (b2, RCF1,...).

3.2.6 RECEIVE SYSTEM INTERFACE

The received data can be output to the system side in four modes: Single Rail NRZ Format mode, Dual Rail NRZ Format mode, Dual Rail RZ Format mode and Dual Rail Sliced mode, as selected by the R_MD[1:0] bits (b1~0, RCF1).

If data is output on RDn in NRZ format and the recovered clock is output on RCLKn, the receive system interface is in Single Rail NRZ Format mode. In this mode, the data is decoded and updated on the active edge of RCLKn. RCLKn outputs a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock.

If data is output on RDPn and RDNn in NRZ format and the recovered clock is output on RCLKn, the receive system interface is in Dual Rail NRZ Format mode. In this mode, the data is un-decoded and updated on the active edge of RCLKn. RCLKn outputs a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock.

If data is output on RDPn and RDNn in RZ format and the recovered clock is output on RCLKn, the receive system interface is in Dual Rail RZ Format mode. In this mode, the data is un-decoded and updated on the active edge of RCLKn. RCLKn outputs a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock.

If data is output on RDPn and RDNn in RZ format directly after passing through the Slicer, the receive system interface is in Dual Rail Sliced mode. In this mode, the data is raw sliced and un-decoded.

Table-3 summarizes the multiplex pin used in different receive system interface.

Table-3 Multiplex Pin Used in Receive System Interface

Receive System Interface	Multiplex P	Multiplex Pin Used On Receive System Interface									
interface	RDn / RDPn	RDNn	RCLKn								
Single Rail NRZ Format	RDn ¹		RCLKn ²								
Dual Rail NRZ Format	RDPn ¹	RDNn ¹	RCLKn ²								
Dual Rail RZ Format	RDPn ¹	RDNn ¹	RCLKn ²								
Dual Rail Sliced	RDPn ¹	RDNn ¹									
Note: 1. The active level on RDr	n, RDPn and RDNr	n is selected by the	e RD_INV bit (b3,								

 The active level on RDn, RDPn and RDNn is selected by the RD_INV bit (b3, RCF1,...).

2. The active edge of RCLKn is selected by the RCK_ES bit (b4, RCF1,...).

3.2.7 RECEIVER POWER DOWN

Set the R_OFF bit (b5, RCF0,...) to '1' will power down the corresponding receiver.

In this way, the corresponding receive circuit is turned off and the RTIPn/RRINGn pins are forced to High-Z state. The pins on receive system interface (including RDn/RDPn, RDNn, RCLKn) will be in High-Z state if the RHZ bit (b6, RCF0,...) is '1' or in low level if the RHZ bit (b6, RCF0,...) is '0'.

After clearing the R_OFF bit (b5, RCF0,...), it will take 1 ms for the receiver to achieve steady state, i.e., to return to the previous configuration and performance.

3.3 TRANSMIT PATH

3.3.1 TRANSMIT SYSTEM INTERFACE

The data from the system side is input to the device in three modes: Single Rail NRZ Format mode, Dual Rail NRZ Format mode and Dual Rail RZ Format mode, as selected by the T_MD[1:0] bits (b1~0, TCF1,...).

If data is input on TDn in NRZ format and a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock is input on TCLKn, the transmit system interface is in Single Rail NRZ Format mode. In this mode, the data is encoded and sampled on the active edge of TCLKn.

If data is input on TDPn and TDNn in NRZ format and a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock is input on TCLKn, the transmit system interface is in Dual Rail NRZ Format mode. In this mode, the data is pre-encoded and sampled on the active edge of TCLKn.

If data is input on TDPn and TDNn in RZ format and no transmit clock is input, the transmit system interface is in Dual Rail RZ Format mode. In this mode, the data is pre-encoded.

Table-4 summarizes the multiplex pin used in different transmit system interface.

Table-4 Multiplex Pin Used in Transmit System Interface

Transmit System Interface	Multiplex P	Multiplex Pin Used On Transmit System Interface									
interface	TDn / TDPn	TDNn	TCLKn / TDNn								
Single Rail NRZ Format	TDn ¹		TCLKn ²								
Dual Rail NRZ Format	TDPn ¹	TDNn ¹	TCLKn ²								
Dual Rail RZ Format	TDPn ¹		TDNn ¹								

Note:

1. The active level on TDn, TDPn and TDNn is selected by the TD_INV bit (b3, TCF1,...).

2. The active edge of TCLKn is selected by the TCK_ES bit (b4, TCF1,...). If TCLKn is missing, i.e., no transition for more than 64 T1/E1 clock cycles, the TCKLOS_S bit (b3, STAT0,...) will be set. A transition from '0' to '1' on the TCKLOS_S bit (b3, STAT0,...) or any transition (from '0' to '1' or from '1' to '0') on the TCKLOS_S bit (b3, STAT0,...) will set the TCKLOS_IS bit (b3, INTS0,...) to '1', as selected by the TCKLOS_IES bit (b3, INTES,...). When the TCKLOS_IS bit (b3, INTS0,...) is '1', an interrupt will be reported by INT if not masked by the TCKLOS_IM bit (b3, INTM0,...).

3.3.2 T_X CLOCK RECOVERY

The Tx Clock Recovery is used only when the transmit system interface is in Dual Rail RZ Format mode. When the transmit system interface is in other modes, the Tx Clock Recovery is bypassed automatically.

The Tx Clock Recovery is used to recover the clock signal from the data input on TDPn and TDNn.

3.3.3 ENCODER

The Encoder is used only when the transmit system interface is in Single Rail NRZ Format mode. When the transmit system interface is in other modes, the Encoder is bypassed automatically.

In T1/J1 mode, the data to be transmitted is encoded by AMI or B8ZS line code rule. In E1 mode, the data to be transmitted is encoded by AMI or HDB3 line code rule. The line code rule is selected by the T_CODE bit (b2, TCF1,...).

3.3.4 WAVEFORM SHAPER

The IDT82P20416 provides two ways to manipulate the pulse shape before data is transmitted:

- · Preset Waveform Template;
- · User-Programmable Arbitrary Waveform.

3.3.4.1 Preset Waveform Template

In T1/J1 applications, the waveform template meets T1.102. The T1 template is shown in Figure-9. It is measured in the far end, as shown in Figure-10. The J1 template is measured in the near end line side.

In T1 applications, to meet the template, five preset waveform templates are provided according to five grades of cable length. The selection is made by the PULS[3:0] bits (b3~0, PULS,...). In J1 applications, the PULS[3:0] bits (b3~0, PULS,...) should be set to '0111'. Refer to Table-5 for details.

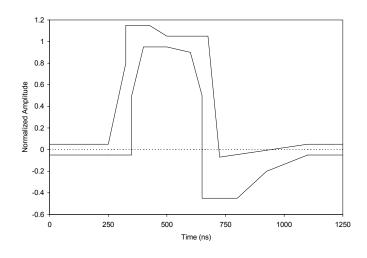


Figure-9 DSX-1 Waveform Template

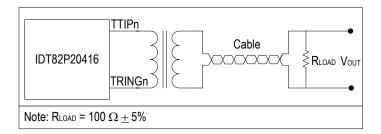


Figure-10 T1 Waveform Template Measurement Circuit

Table-5 PULS[3:0] Setting in T1/J1 Mode

Cable Conditions	PULS[3:0]
DSX1 - 0 ~ 133 ft	0010
DSX1 - 133 ~ 266 ft	0011
DSX1 - 266 ~ 399 ft	0100
DSX1 - 399 ~ 533 ft	0101
DSX1 - 533 ~ 655 ft	0110
J1 - 0 ~ 655 ft	0111

In E1 applications, the waveform template meets G.703, as shown in Figure-11. It is measured in the near end line side, as shown in Figure-12.

In E1 applications, the PULS[3:0] should be set to '0000' if differential signals (output from TTIP and TRING) are coupled to a 75 Ω coaxial cable using Internal Impedance matching mode; the PULS[3:0] should be set to '0001' for other E1 interfaces. Refer to Table-6 for details.

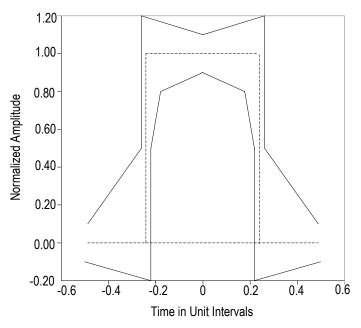


Figure-11 E1 Waveform Template

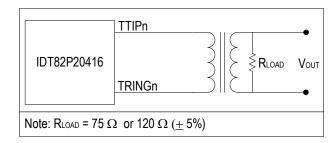


Figure-12 E1 Waveform Template Measurement Circuit

Table-6 F	PULS[3:0]	Setting	in E	1 Mode
-----------	-----------	---------	------	--------

Interface Conditions	PULS[3:0]
E1 75 Ω differential interface, Internal Impedance matching mode	0000
Other E1 interface	0001

After one of the preset waveform templates is selected, the preset waveform amplitude can be adjusted to get the desired waveform.

In T1 mode, the standard value of the SCAL[5:0] bits (b5~0, SCAL,...) is '110110' which is also the default value. The adjusting is made by increasing or decreasing by '1' from the standard value to scale up or down at a percentage ratio of 2% against the preset waveform amplitude.

In E1 mode, the SCAL[5:0] bits (b5~0, SCAL,...) should be set to '100001' to get the standard amplitude. The adjusting is made by increasing or decreasing by '1' from the standard value to scale up or down at a percentage ratio of 3%.

In summary, do the following step by step, the desired waveform will be got based on the preset waveform template:

- Select one preset waveform template by setting the PULS[3:0] bits (b3~0, PULS,...);
- Write '100001 to the SCAL[5:0] bits (b5~0, SCAL,...) if E1 mode is selected.
- Write the scaling value to the SCAL[5:0] bits (b5~0, SCAL,...) to scale the amplitude of the selected preset waveform template (this step is optional).

16-CHANNEL SHORT HAUL T1/E1/J1 LINE INTERFACE UNIT

3.3.4.2 User-Programmable Arbitrary Waveform

When the PULS[3:0] bits (b3~0, PULS,...) are set to '1XXX', userprogrammable arbitrary waveform will be used in the corresponding channel.

Each waveform shape can extend up to $1\frac{1}{4}$ UIs (Unit Interval), and is

divided into 20 sub-phases that are addressed by the SAMP[4:0] bits (b4~0, AWG0,...). The waveform amplitude of each phase is represented by a binary byte, within the range from +63 to -63, stored in the WDAT[6:0] bits (b6~0, AWG1,...) in signed magnitude form. The maximum number +63 (D) represents the maximum positive amplitude of the transmit pulse while the most negative number -63 (D) represents the maximum negative amplitude of the transmit pulse. Therefore, up to 20 bytes are used.

There are eight standard templates which are stored in a local ROM. One of them can be selected as reference and made some changes to get the desired waveform.

To do this, the first step is to choose a set of waveform value from the standard templates. The selected waveform value should be the most similar to the desired waveform shape. Table 7 to Table 14 list the sample data of each template.

Then modify the sample data to get the desired transmit waveform shape. By increasing or decreasing by '1' from the standard value in the SCAL[5:0] bits (b5~0, SCAL,...), the waveform amplitude will be scaled up or down.

In summary, do the following for the write operation:

- Modify the sample data in the AWG1 register;
- Write the AWG0 register to implement the write operation, including:
 - Write the sample address to the SAMP[4:0] bits (b4~0, AWG0,...);
 - Write '0' to the RW bit (b5, AWG0,...);
 - Write '1' to the DONE bit (b6, AWG0,...).

Do the following for the read operation:

- · Write the AWG0 register, including:
 - Write sample address to the SAMP[4:0] bits (b4~0, AWG0,...);
 - Write '1' to the RW bit (b5, AWG0,...);
 - Write '1' to the DONE bit (b6, AWG0,...);
- · Read the AWG1 register to get the sample data.

When the write operation is completed, write the scaling value to the SCAL[5:0] bits (b5~0, SCAL,...) to scale the amplitude of the selected standard waveform (- this step is optional).

When more than one UI is used to compose the waveform template and the waveform amplitude is not set properly, the overlap of the two consecutive waveforms will make the waveform amplitude overflow (i.e., exceed the maximum limitation). This overflow is captured by the DAC_IS bit (b7, INTS0,...) and will be reported by the INT pin if enabled by the DAC_IM bit (b7, INTM0,...).

Refer to application note AN-513 'User-Programmable Arbitrary Waveform for DSX1' for more details.

Table-7 Transmit Waveform Value for T1 0 ~ 133 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	17H	27H	27H	26H	25H	25H	25H	24H	23H	4AH	4AH	49H	47H	45H	44H	43H	42H	41H	00H	00H

Table-8 Transmit Waveform Value for T1 133 ~ 266 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	1BH	2EH	2CH	2AH	29H	28H	27H	26H	25H	50H	4FH	4DH	4AH	48H	46H	44H	43H	42H	41H	00H

Table-9 Transmit Waveform Value for T1 266 ~ 399 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	1FH	34H	2FH	2CH	2BH	2AH	29H	28H	25H	57H	53H	50H	4BH	48H	46H	44H	43H	42H	41H	00H

Table-10 Transmit Waveform Value for T1 399 ~ 533 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	20H	3BH	35H	2FH	2EH	2DH	2CH	2AH	28H	58H	58H	53H	4CH	48H	46H	44H	43H	42H	41H	00H

Table-11 Transmit Waveform Value for T1 533 ~ 655 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	20H	3FH	38H	33H	2FH	2EH	2DH	2CH	29H	5FH	5EH	57H	4FH	49H	47H	44H	43H	42H	41H	00H

Table-12 Transmit Waveform Value for E1 75 ohm

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	00H	00H	00H	0CH	30H	00H														

Table-13 Transmit Waveform Value for E1 120 ohm

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	00H	00H	00H	0FH	3CH	00H														

Table-14 Transmit Waveform Value for J1 0 ~ 655 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	17H	27H	27H	26H	25H	25H	25H	24H	23H	4AH	4AH	49H	47H	45H	44H	43H	42H	41H	00H	00H

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3.3.5 LINE DRIVER

The Line Driver can be set to High-Z for protection or in redundant applications.

The following two ways will set the Line Driver to High-Z:

- Setting the OE pin to low will globally set all the Line Drivers to High-Z;
- Setting the OE bit (b6, TCF0,...) to '0' will set the corresponding Line Driver to High-Z.

By these ways, the functionality of the internal circuit is not affected and TTIPn and TRINGn will enter High-Z state immediately.

3.3.5.1 Transmit Over Current Protection

The Line Driver monitors the Transmit Over Current (TOC) on the line interface. When TOC is detected, the driver's output (i.e., output on TTIPn/TRINGn) is determined by the THZ_OC bit (b4, TCF0,...). If the THZ_OC bit (b4, TCF0,...) is '0', the driver's output current (peak to peak) is limited to 100 mA; if the THZ_OC bit (b4, TCF0,...) is '1', the driver's output will enter High-Z. TOC is indicated by the TOC_S bit (b4, STAT0,...) or any transition (from '0' to '1' on the TOC_S bit (b4, STAT0,...) will set the TOC_IS bit (b4, INTS0,...) to '1', as selected by the TOC_IES bit (b4, INTES,...). When the TOC_IS bit (b4, INTS0,...) is '1', an interrupt will be reported by INT if not masked by the TOC_IM bit (b4, INTM0,...).

3.3.6 T_X TERMINATION

The transmit line interface supports Transmit Differential mode and Transmit Single Ended mode, as selected by the T_SING bit (b3, TCF0,...). In Transmit Differential mode, both TTIPn and TRINGn are used to transmit signals to the line side. In Transmit Single Ended mode, only TTIPn is used to transmit signal.

The line interface can be connected with T1 100 Ω , J1 110 Ω or E1 120 Ω twisted pair cable or E1 75 Ω coaxial cable.

The transmit impedance matching is realized by using internal impedance matching or external impedance matching for each channel in different applications.

3.3.6.1 Transmit Differential Mode

In Transmit Differential mode, different applications have different impedance matching. For T1/J1 applications, only Internal Impedance Matching is supported. For E1 applications, both Internal and External Impedance Matching are supported.

Internal Impedance Matching circuit uses an internal programmable resistor (IM) only.

External Impedance Matching circuit uses an external resistor (Rt) only.

A twisted pair cable can be connected with a 1:2 (step up) transformer or without a transformer (transformer-less), while a coaxial cable must be connected with a 1:2 transformer.

The T_TERM[2:0] bits (b2~0, TCF0,...) should be set according to different cable conditions, whether a transformer is used, and what kind of Impedance Matching is selected.

Table-15 lists the recommended impedance matching value in different applications. Figure-13 to Figure-15 show the connection for one channel in different applications.

The transformer-less connection will offer a termination option with reduced cost and board space. However, the waveform amplitude is not standard compliant, and surge protection and common mode depression should be enhanced depending on equipment environment..

Table-15 Impedance Matching Value in Transmit Differential Mode

Cable Condition	Interna	I Impedance Mat	ching	Extern	al Impedance Mat	ching
Cable Condition	T_TERM[2:0]	Rt	PULS[3:0]	T_TERM[2:0]	Rt	PULS[3:0]
T1 100 Ω twisted pair (with transformer)	000		Table-5			
J1 110 Ω twisted pair (with transformer)	001		0111		(not supported)	
E1 120 Ω twisted pair (with transformer)	010		0001	111	10 Ω	0000
E1 75 Ω coaxial (with transformer)	011	0	0000			0000
T1 100 Ω twisted pair (transformer-less)	100		Table-5		I	
J1 110 Ω twisted pair (transformer-less)	101		0111		(not supported)	
E1 120 Ω twisted pair (transformer-less)	110		0001			



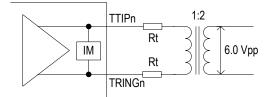


Figure-13 Transmit Differential Line Interface with Twisted Pair Cable (with Transformer)

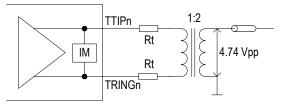


Figure-14 Transmit Differential Line Interface with Coaxial Cable (with transformer)

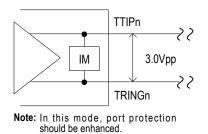


Figure-15 Transmit Differential Line Interface with Twisted Pair Cable (transformer-less, non standard compliant)

3.3.7 TRANSMITTER POWER DOWN

Set the T_OFF bit (b5, TCF0,...) to '1' will power down the corresponding transmitter.

In this way, the corresponding transmit circuit is turned off. The pins on the transmit line interface (including TTIPn and TRINGn) will be in High-Z state. The input on the transmit system interface (including TDn, TDPn, TDNn and TCLK) is ignored. The output on the transmit system interface will be in High-Z state.

After clearing the T_OFF bit (b5, TCF0,...), it will take 1 ms for the transmitter to achieve steady state, i.e., return to the previous configuration and performance.

3.3.8 OUTPUT HIGH-Z ON TTIP AND TRING

TTIPn and TRINGn can be set to High-Z state globally or on a perchannel basis.

The following three conditions will set TTIPn and TRINGn to High-Z state globally:

- · Connecting the OE pin to low;
- · Loss of MCLK (i.e., no transition on MCLK for more than 1 ms);
- Power on reset, hardware reset by pulling RST to low for more than 2 µs or global software reset by writing the RST register.

The following six conditions will set TTIPn and TRINGn to High-Z state on a per-channel basis:

- Writing '0' to the OE bit (b6, TCF0,...);
- Loss of TCLKn in Transmit Single Rail NRZ Format mode or Transmit Dual Rail NRZ Format mode (i.e., no transition on TCLKn for more than 64 XCLK¹ cycles) except that the channel is in Remote Loopback or transmit internal pattern with XCLK;
- Transmitter power down;
- Per-channel software reset by writing '1' to the CHRST bit (b1, CHCF,...);
- Setting the THZ_OC bit (b4, TCF0,...) to '1' when transmit driver over-current is detected.
 - 1. XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.

3.4 JITTER ATTENUATOR (RJA & TJA)

Two Jitter Attenuators are provided for each channel of receiver and transmitter. Each Jitter Attenuator can be enabled or disabled, as determined by the RJA_EN/TJA_EN bit (b3, RJA/TJA,...) respectively.

Each Jitter Attenuator consists of a FIFO and a DPLL, as shown in Figure-16.

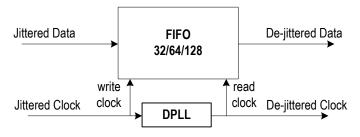


Figure-16 Jitter Attenuator

The FIFO is used as a pool to buffer the jittered input data, then the data is clocked out of the FIFO by a de-jittered clock. The depth of the FIFO can be 32 bits, 64 bits or 128 bits, as selected by the RJA_DP[1:0]/ TJA_DP[1:0] bits (b2~1, RJA/TJA,...). Accordingly, the typical delay produced by the Jitter Attenuator is 16 bits, 32 bits or 64 bits. The 128-bit FIFO is used when large jitter tolerance is expected, while the 32-bit FIFO is used in delay sensitive applications.

The DPLL is used to generate a de-jittered clock to clock out the data stored in the FIFO. The DPLL can only attenuate the incoming jitter whose frequency is above Corner Frequency (CF) by 20 dB per decade falling off. The jitter whose frequency is lower than the CF passes through the DPLL without any attenuation. In T1/J1 applications, the CF of the DPLL is 5 Hz or 1.26 Hz. In E1 applications, the CF of the DPLL is 6.77 Hz or 0.87 Hz. The CF is selected by the RJA_BW/TJA_BW bit (b0, RJA/TJA,...). The lower the CF is, the longer time is needed to achieve synchronization.

If the incoming data moves faster than the outgoing data, the FIFO will overflow. If the incoming data moves slower than the outgoing data, the FIFO will underflow. The overflow and underflow are both captured by the RJA_IS/TJA_IS bit (b5/6, INTS0,...). The occurrence of overflow or underflow will be reported by the INT pin if enabled by the RJA_IM/TJA_IM bit (b5/6, INTM0,...).

To avoid overflow or underflow, the JA-Limit function can be enabled by setting the RJA_LIMT/TJA_LIMT bit (b4, RJA/TJA,...). When the JA-Limit function is enabled, the speed of the outgoing data will be adjusted automatically if the FIFO is 2-bit close to its full or emptiness. Though the JA-Limit function can reduce the possibility of FIFO overflow and underflow, the quality of jitter attenuation is deteriorated.

The performance of the Jitter Attenuator meets ITUT I.431, G.703, G.736-739, G.823, G.824, ETSI 300011, ETSI TBR12/13, AT&T TR62411, TR43802, TR-TSY 009, TR-TSY 253 and TR-TRY 499. Refer to Section 8.12 Jitter Attenuation Characteristics for the jitter performance.

3.5 DIAGNOSTIC FACILITIES

The diagnostic facilities include:

- BPV (Bipolar Violation) / CV (Code Violation) detection and BPV insertion;
- · EXZ (Excessive Zero) detection;
- · LOS (Loss Of Signal) detection;
- AIS (Alarm Indication Signal) detection and generation;
- Pattern generation and detection, including PRBS (Pseudo Random Bit Sequence), ARB (Arbitrary Pattern) and IB (Inband Loopback).

The above defects, alarms or patterns can be counted by an internal Error Counter, indicated by the respective interrupt bit.

3.5.1 BIPOLAR VIOLATION (BPV) / CODE VIOLATION (CV) DETECTION AND BPV INSERTION

3.5.1.1 Bipolar Violation (BPV) / Code Violation (CV) Detection

BPV/CV is monitored in both the receive path and the transmit path. BPV is detected when the data is AMI coded and CV is detected when the data is B8ZS/HDB3 coded. If the transmit system interface is in Transmit Single Rail NRZ Format mode, the BPV/CV detection is disabled in the transmit path automatically.

A BPV is detected when two consecutive pulses of the same polarity are received.

A CV is detected when two consecutive BPVs of the same polarity that are not a part of the B8ZS/HDB3 zero substitution are received.

When BPV/CV is detected in the receive path, the Line Bipolar Violation LBPV_IS bit (b4, INTS2,...) will be set and an interrupt will be reported by INT if not masked by the LBPV_IM bit (b4, INTM2,...).

When BPV/CV is detected in the transmit path, the System Bipolar Violation SBPV_IS bit (b5, INTS2,...) will be set and an interrupt will be reported by INT if not masked by the SBPV_IM bit (b5, INTM2,...).

3.5.1.2 Bipolar Violation (BPV) Insertion

The BPV can only be inserted in the transmit path.

A BPV will be inserted on the next available mark in the data stream to be transmitted by writing a '1' to the BPV_INS bit (b6, ERR,...). This bit will be reset once BPV insertion is done.

3.5.2 EXCESSIVE ZEROES (EXZ) DETECTION

EXZ is monitored in both the receive path and the transmit path.

Different line code has different definition of the EXZ. The IDT82P20416 provides two standards of EXZ definition for each kind of line code rule. The standards are ANSI and FCC, as selected by the EXZ_DEF bit (b7, ERR,...). Refer to Table-16 for details.

Table-16 EXZ Definition

Line Code	Defir	nition
Rule	ANSI (EXZ_DEF = 0)	FCC (EXZ_DEF = 1)
AMI	An EXZ is detected when any string of more than 15 consecutive '0's are received.	T1/J1 - An EXZ is detected when any string of more than 80 consecutive '0's are received. E1 - An EXZ is detected when any string of more than 15 consecutive '0's are received.
B8ZS	An EXZ is detected when any string of more than 7 consecutive '0's are received.	An EXZ is detected when any string of more than 7 consecutive '0's are received.
HDB3	An EXZ is detected when any string of more than 3 consecutive '0's are received.	
,	tem interface is in Transmit Single R g to the standard of AMI.	Rail NRZ Format mode, the EXZ is

When EXZ is detected in the receive path, the LEXZ_IS bit (b2, INTS2,...) will be set and an interrupt will be reported by \overline{INT} if not masked by the LEXZ_IM bit (b2, INTM2,...).

When EXZ is detected in the transmit path, the SEXZ_IS bit (b3, INTS2,...) will be set and an interrupt will be reported by $\overline{\text{INT}}$ if not masked by the SEXZ_IM bit (b3, INTM2,...).

EXZ may be counted by an internal Error Counte. Refer to Chapter 3.5.6 Error Counter.

3.5.3 LOSS OF SIGNAL (LOS) DETECTION

The IDT82P20416 detects three kinds of LOS:

- · LLOS: Line LOS, detected in the receive path;
- · SLOS: System LOS, detected in the transmit system side;
- TLOS: Transmit LOS, detected in the transmit line side.

3.5.3.1 Line LOS (LLOS)

Table-17 LLOS Criteria

The amplitude and density of the data received from the line side are monitored. When the amplitude of the data is less than Q Vpp for N consecutive pulse intervals, LLOS is declared. When the amplitude of the data is more than P Vpp and the average density of marks is at least 12.5% for M consecutive pulse intervals starting with a mark, LLOS is cleared. Here Q is defined by the ALOS[2:0] bits (b6~4, LOS,...). P is the sum of Q and 250 mVpp. N and M are defined by the LAC bit (b7, LOS,...). Refer to Table-17 for details.

In T1/J1 mode, LLOS detection supports ANSI T1.231 and I.431. In E1 mode, LLOS detection supports G.775 and ETSI 300233/I.431. The criteria are selected by the LAC bit (b7, LOS,...).

When LLOS is detected, the LLOS_S bit (b0, STAT0,...) will be set. A transition from '0' to '1' on the LLOS_S bit (b0, STAT0,...) or any transition (from '0' to '1' or from '1' to '0') on the LLOS_S bit (b0, STAT0,...) will set the LLOS_IS bit (b0, INTS0,...) to '1', as selected by the LOS_IES bit (b1, INTES,...). When the LLOS_IS bit (b0, INTS0,...) is '1', an interrupt will be reported by INT if not masked by the LLOS_IM bit (b0, INTM0,...).

Two pins (LLOS0 and LLOS) are dedicated to LLOS indication. Whether LLOS is detected in channel 0 or not, LLOS0 is high for a CLKE1 clock cycle to indicate the start position on LLOS. LLOS indicates LLOS status of all 16 channels in a serial format and repeats every 17 cycles. Refer to Figure-17. LLOS0 and LLOS are updated on the rising edge of CLKE1. When the clock output on CLKE1 is disabled, LLOS0 and LLOS will be held in High-Z state. The output on CLKE1 is controlled by the CLKE1_EN bit (b3, CLKG) and the CLKE1 bit (b2, CLKG). Refer to section 8.11 on page 110 for CLKE1 timing characteristics.

LLOS may be counted by an internal Error Counte. Refer to Section 3.5.6 Error Counter.

During LLOS, in Receive Single Rail NRZ Format mode, Receive Dual Rail NRZ Format mode and Receive Dual Rail RZ Format mode, RDPn/RDNn output low level. In Receive Dual Rail Sliced mode RDPn/RDNn still output sliced data. RCLKn (if available) outputs high level or XCLK¹, as selected by the RCKH bit (b7, RCF0,...).

During LLOS, if any of AIS, pattern generation in the receive path or Digital Loopback is enabled or automatic digital loopback happens, RDPn/RDNn and RCLKn output corresponding data and clock, and the setting of the RCKH bit (b7, RCF0,...) is ignored. Refer to the corresponding chapters for details.

^{1.} XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.

Operation Mode	LAC	Criteria	LLOS Declaring	LLOS Clearing
T1/J1	0	ANSI T1.231	below Q Vpp, N = 175 bits	above P Vpp, 12.5% mark density with less than 100 consecutive zeros, M = 175 bits
11/01	1	ANSI I.431	below Q Vpp, N = 1544 bits	above P Vpp, 12.5% mark density with less than 100 consecutive zeros, M = 175 bits
	0	G.775	below Q Vpp, N = 32 bits	above P Vpp, 12.5% mark density with less than 16 consecutive zeros, M = 32 bits
E1	1	ETSI 300233/ I.431	below Q Vpp, N = 2048 bits	above P Vpp, 12.5% mark density with less than 16 consecutive zeros, M = 32 bits

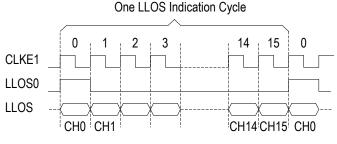


Figure-17 LLOS Indication on Pins

3.5.3.2 System LOS (SLOS)

SLOS can only be detected when the transmit system interface is in Dual Rail NRZ Format mode or in Dual Rail RZ Format mode.

The amplitude and density of the data input from the transmit system side are monitored. When the input '0's are equal to or more than N consecutive pulse intervals, SLOS is declared. When the average density of marks is at least 12.5% for M consecutive pulse intervals starting with a mark, SLOS is cleared. Here N and M are defined by the LAC bit (b7, LOS,...). Refer to Table-18 for details.

In T1/J1 mode, SLOS detection supports ANSI T1.231 and I.431. In E1 mode, SLOS detection supports G.775 and ETSI 300233/I.431. The criteria are selected by the LAC bit (b7, LOS,...).

When SLOS is detected, the SLOS_S bit (b1, STAT0,...) will be set. A transition from '0' to '1' on the SLOS_S bit (b1, STAT0,...) or any transition (from '0' to '1' or from '1' to '0') on the SLOS_S bit (b1, STAT0,...) will set the SLOS_IS bit (b1, INTS0,...) to '1', as selected by the LOS_IES bit (b1, INTES,...). When the SLOS_IS bit (b1, INTS0,...) is '1', an interrupt will be reported by INT if not masked by the SLOS_IM bit (b1, INTM0,...).

SLOS may be counted by an internal Error Counter. Refer to Section 3.5.6 Error Counter.

Operation Mode	LAC	Criteria	SLOS Declaring ¹	SLOS Clearing ¹
T1/J1	0	ANSI T1.231	no pulse detected for N consecutive pulse intervals, N = 175 bits	12.5% mark density with less than 100 consecutive zeros for M consecutive pulse intervals, M = 175 bits
11/31	1	ANSI I.431	no pulse detected for N consecutive pulse intervals, N = 1544 bits	12.5% mark density with less than 100 consecutive zeros for M consecutive pulse intervals, M = 175 bits
E1	0	G.775	no pulse detected for N consecutive pulse intervals, N = 32 bits	12.5% mark density with less than 16 consecutive zeros for M consecutive pulse intervals, M = 32 bits
	1	ETSI 300233/ I.431	no pulse detected for N consecutive pulse intervals, N = 2048 bits	12.5% mark density with less than 16 consecutive zeros for M consecutive pulse intervals, M = 32 bits

Table-18 SLOS Criteria

Note:

1. System input ports are schmitt-trigger inputs)

3.5.3.3 Transmit LOS (TLOS)

The amplitude and density of the data output on the transmit line side are monitored. When the amplitude of the data is less than a certain voltage for a certain period, TLOS is declared. The voltage is defined by the TALOS[1:0] bits (b3~2, LOS,...). The period is defined by the TDLOS[1:0] bits (b1~0, LOS,...). When a valid pulse is detected, i.e., the amplitude is above the setting in the TALOS[1:0] bits (b3~2, LOS,...), TLOS is cleared.

When TLOS is detected, the TLOS_S bit (b2, STAT0,...) will be set. A transition from '0' to '1' on the TLOS_S bit (b2, STAT0,...) or any transition (from '0' to '1' or from '1' to '0') on the TLOS_S bit (b2, STAT0,...) will set the TLOS_IS bit (b2, INTS0,...) to '1', as selected by the TLOS_IES bit (b2, INTES,...). When the TLOS_IS bit (b2, INTS0,...) is '1', an interrupt will be reported by \overline{INT} if not masked by the TLOS_IM bit (b2, INTM0,...).

TLOS may be counted by an internal Error Counter. Refer to Section 3.5.6 Error Counter.

TLOS can be used to monitor the LOS in the transmit line side between two channels. The connection between the two channels is shown in Figure-18. The two channels can be of the same device or different devices on the premises that the transmit line interfaces are in the same mode and at least the output of one channel is in High-Z state. Table-19 lists each results in this case. In the left two columns, the OE bit (b6, TCF0,...) of the two channels controls the output status in the transmit line side to ensure that at least one channel is in High-Z state.

Table-19 TLOS Detection Between Two Channels

The middle two columns list the internal operation status. In the right two columns, the TLOS_S bit (b2, STAT0,...) of the two channels indicates the TLOS status in the transmit line side.

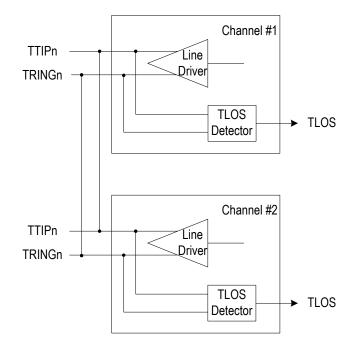


Figure-18 TLOS Detection Between Two Channels

Output Status ~ Con	ntrolled By the OE Bit	Internal Ope	ration Status	TLOS Status ~ Indicat	ed By the TLOS_S Bit
Channel #1	Channel #2	Channel #1	Channel #2	Channel #1	Channel #2
Normal ~ 1	High-Z ~ 0	Normal	(don't-care)	No TLOS ~ 0	No TLOS ~ 0
Normal ~ 1	High-Z ~ 0	Failure	Normal	TLOS Detected ~ 1 *	TLOS Detected ~ 1
High-Z ~ 0	Normal ~ 1	(don't-care)	Normal	No TLOS ~ 0	No TLOS ~ 0
High-Z ~ 0	Normal ~ 1	Normal	Failure	TLOS Detected ~ 1	TLOS Detected ~ 1 *
High-Z ~ 0	High-Z ~ 0	(don't-care)	(don't-care)	TLOS Detected ~ 1	TLOS Detected ~ 1

3.5.4 ALARM INDICATION SIGNAL (AIS) DETECTION AND GEN-ERATION

3.5.4.1 Alarm Indication Signal (AIS) Detection

AIS is monitored in both the receive path and the transmit path.

When the mark density in the received data or in the data input from the transmit system side meets certain criteria, AIS is declared or cleared. In T1/J1 mode, the criteria are in compliance with ANSI T1.231. In E1 mode, the criteria are in compliance with ITU G.775 or ETSI 300233, as selected by the LAC bit (b7, LOS,...). Refer to Table-20 for details.

Table-20 AIS Criteria

		ITU G.775 for E1 (LAC = 0)	ETSI 300233 for E1 (LAC = 1)	ANSI T1.231 for T1 (LAC = 0 or 1)
AIS	6 Declaring	Less than 3 zeros are received in each of two consecutive 512-bit data streams.	Less than 3 zeros are received in a 512-bit data stream.	Less than 9 zeros are received in a 8192-bit stream, i.e., less than 99.9% of marks in a period of 5.3 ms are received.
AIS	S Clearing	3 or more zeros are received in each of two consecutive 512-bit data streams.	3 or more zeros are received in a 512-bit data stream.	9 or more zeros are received in a 8192-bit data stream.

When AIS is detected in the receive path, the LAIS_S bit (b6, STAT1,...) will be set. A transition from '0' to '1' on the LAIS_S bit (b6, STAT1,...) or any transition (from '0' to '1' or from '1' to '0') on the LAIS_S bit (b6, STAT1,...) will set the LAIS_IS bit (b6, INTS1,...) to '1', as selected by the AIS_IES bit (b6, INTES,...). When the LAIS_IS bit (b6, INTS1,...) is '1', an interrupt will be reported by INT if not masked by the LAIS_IM bit (b6, INTM1,...).

When AIS is detected in the transmit path, the SAIS_S bit (b7, STAT1,...) will be set. A transition from '0' to '1' on the SAIS_S bit (b7, STAT1,...) or any transition (from '0' to '1' or from '1' to '0') on the SAIS_S bit (b7, STAT1,...) will set the SAIS_IS bit (b7, INTS1,...) to '1', as selected by the AIS_IES bit (b6, INTES,...). When the SAIS_IS bit (b7, INTS1,...) is '1', an interrupt will be reported by INT if not masked by the SAIS_IM bit (b7, INTM1,...).

AIS may be counted by an internal Error Counte. Refer to Section 3.5.6 Error Counter.

3.5.4.2 (Alarm Indication Signal) AIS Generation

AIS can be generated automatically in the receive path and the transmit path.

In the receive path, when the ASAIS_LLOS bit (b2, AISG,...) is set, AIS will be generated automatically once LLOS is detected. When the ASAIS_SLOS bit (b3, AISG,...) is set, AIS will be generated automatically once SLOS is detected. When AIS is generated, RDPn/RDNn output all '1's. RCLKn (if available) outputs XCLK.

In the transmit path, when the ALAIS_LLOS bit (b0, AISG,...) is set, AIS will be generated automatically once LLOS is detected. When the ALAIS_SLOS bit (b1, AISG,...) is set, AIS will be generated automatically once SLOS is detected. When AIS is generated, TTIPn/TRINGn output all '1's.

In the transmit path, the AIS transmission is controled by the TXAIS bit (b4, AISG,...). When the TXAIS bit (b4, AISG,...) is set to '1', all '1's pattern is transmitted at TTIPn/TRINGn.

AIS generation uses XCLK¹ as reference clock.

If pattern (including PRBS, ARB and IB) is generated in the same direction, the priority of pattern generation is higher. The generated pattern will overwrite automatic AIS. Refer to Section 3.5.5.1 Pattern Generation for the output data and clock.

^{1.} XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.

3.5.5 PRBS, QRSS, ARB AND IB PATTERN GENERATION AND DETECTION

The pattern includes: Pseudo Random Bit Sequence (PRBS), Quasi-Random Signal Source (QRSS), Arbitrary Pattern (ARB) and Inband Loopback (IB).

3.5.5.1 Pattern Generation

The pattern can be generated in the receive path or the transmit path, as selected by the PG_POS bit (b3, PG,...).

The pattern to be generated is selected by the PG_EN[1:0] bits (b5~4, PG,...).

If PRBS is selected, three kinds of PRBS patterns with maximum zero restriction according to ITU-T 0.151 and AT&T TR62411 are provided. They are: $(2^{2}0 - 1)$ QRSS per 0.150-4.5, $(2^{15} - 1)$ PRBS per 0.152 and $(2^{11} - 1)$ PRBS per 0.150, as selected by the PRBG_SEL[1:0] bits (b1~0, PG,...).

If ARB is selected, the content is programmed in the ARB[23:0] bits (b7~0, ARBH~ARBM~ARBL,...).

If IB is selected, the IB generation is in compliance with ANSI T1.403. The length of the IB code can be 3 to 8 bits, as determined by the IBGL[1:0] bits ($b5\sim4$, IBL,...). The content is programmed in the IBG[7:0] bits ($b7\sim0$, IBG,...).

The selected pattern is transmitted repeatedly until the PG_EN[1:0] bits (b5~4, PG,...) is set to '00'.

When pattern is generated in the receive path, the reference clock is XCLK or the recovered clock from the received signal, as selected by the PG_CK bit (b6, PG,...). The selected reference clock is also output on RCLKn (if available).

When pattern is generated in the transmit path, the reference clock is XCLK¹ or the transmit clock, as selected by the PG_CK bit (b6, PG,...). The transmit clock refers to the clock input on TCLKn (in Transmit Single Rail NRZ Format mode and in Transmit Dual Rail NRZ Format mode) or the clock recovered from the data input on TDPn and TDNn (in Transmit Dual Rail RZ Format mode).

In summary, do the followings step by step to generate pattern:

- Select the generation direction by the PG_POS bit (b3, PG,...);
- Select the reference clock by the PG_CK bit (b6, PG,...);
- Select the PRBS pattern by the PRBG_SEL[1:0] bits (b1~0, PG,...) when PRBS is to be generated; program the ARB pattern in the ARB[23:0] bits (b7~0, ARBH~ARBM~ARBL,...) when ARB is to be generated; or set the length and the content of the IB code in the IBGL[1:0] bits (b5~4, IBL,...) and in the IBG[7:0] bits (b7~0, IBG,...) respectively when IB is to be generated;

• Set the PG_EN[1:0] bits (b5~4, PG,...) to generate the pattern.

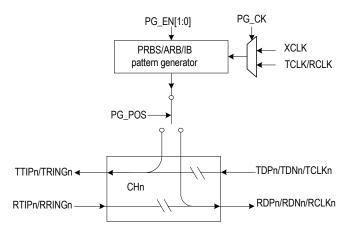
If PRBS or ARB is selected to be generated, the following two steps can be optionally implemented after the pattern is generated:

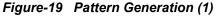
- Insert a single bit error by writing '1' to the ERR_INS bit (b5, ERR,...);
- Invert the generated pattern by setting the PAG_INV bit (b2, PG,...).

If pattern is generated in the receive path, the generated pattern should be encoded by using AMI or B8ZS (for T1/J1) / HDB3 (for E1) in Receive Dual Rail NRZ Format mode, Receive Dual Rail RZ Format mode and Receive Dual Rail Sliced mode. The encoding rule is selected by the R_CODE bit (b2, RCF1,...).

If pattern is generated in the transmit path, the generated pattern should be encoded by using AMI or B8ZS (for T1/J1) / HDB3 (for E1). The encoding rule is selected by the T_CODE bit (b2, TCF1,...).

The pattern generation is shown in Figure-19 and Figure-20.





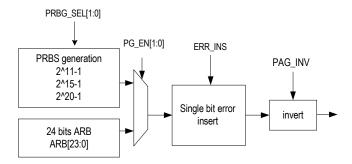


Figure-20 Pattern Generation (2)

The priority of pattern generation is higher than that of AIS generation. If they are generated in the same direction, the generated pattern will overwrite the generated AIS.

^{1.} XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.

3.5.5.2 Pattern Detection

Data received from the line side or data input from the transmit system side may be extracted for pattern detection. The direction of data extraction is determined by the PD_POS bit (b3, PD,...). One of PRBS or ARB pattern is selected for detection and IB detection is always active.

If data is extracted from the receive path, before pattern detection the data should be decoded by using AMI or B8ZS (for T1/J1) / HDB3 (for E1). The decoding rule is selected by the R_CODE bit (b2, RCF1,...).

If data is extracted from the transmit path, before pattern detection the data should be decoded by using AMI or B8ZS (for T1/J1) / HDB3 (for E1) in Transmit Dual Rail NRZ Format mode and Transmit Dual Rail RZ Format mode. The decoding rule is selected by the T_CODE bit (b2, TCF1,...).

<u>Pseudo Random Bit Sequence (PRBS) /Arbitrary Pattern (ARB)</u> <u>Detection</u>

The extracted data can be optionally inverted by the PAD_INV bit (b2, PD,...) before PRBS/ARB detection.

The extracted data is used to compare with the desired pattern. The desired pattern is re-generated from the extracted data if the desired pattern is (2^20 - 1) QRSS per O.150-4.5, (2^15 - 1) PRBS per O.152 or (2^11 - 1) PRBS per O.150; or the desired pattern is programmed in the ARB[23:0] bits (b7~0, ARBH~ARBM~ARBL,...) if the desired pattern is ARB. The desired pattern is selected by the PAD_SEL[1:0] bits (b1~0, PD,...).

In summary, do the followings step by step to detect PRBS/ARB:

- Select the detection direction by the PD_POS bit (b3, PD,...);
- Set the ARB[23:0] bits (b7~0, ARBH~ARBM~ARBL,...) if the ARB pattern is desired - this step is omitted if the PRBS pattern is desired;
- Select the desired PRBS/ARB pattern by the PAD_SEL[1:0] bits (b1~0, PD,...).

The priority of decoding, data inversion, pattern re-generation, bit programming and pattern comparison is shown in Figure-21.

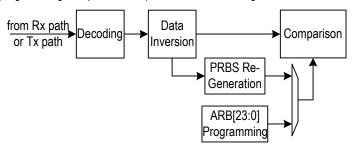


Figure-21 PRBS / ARB Detection

During comparison, if the extracted data coincides with the re-generated PRBS pattern or the programmed ARB pattern for more than 64-bit hopping window, the pattern is synchronized and the PA_S bit (b5, STAT1,...) will be set.

In synchronization state, if more than 6 PRBS/ARB errors are detected in a 64-bit hopping window, the pattern is out of synchronization and the PA_S bit (b5, STAT1,...) will be cleared.

In synchronization state, each mismatched bit will generate a PRBS/ ARB error. When a PRBS/ARB error is detected during the synchronization, the ERR_IS bit (b1, INTS2,...) will be set and an interrupt will be reported by INT if not masked by the ERR_IM bit (b1, INTM2,...). The PRBS/ARB error may be counted by an internal Error Counter. Refer to Section 3.5.6 Error Counter.

A transition from '0' to '1' on the PA_S bit (b5, STAT1,...) or any transition (from '0' to '1' or from '1' to '0') on the PA_S bit (b5, STAT1,...) will set the PA_IS bit (b5, INTS1,...) to '1', as selected by the PA_IES bit (b5, INTES,...). When the PA_IS bit (b5, INTS1,...) is '1', an interrupt will be reported by INT if not masked by the PA_IM bit (b5, INTM1,...).

Inband Loopback (IB) Detection

The IB detection is in compliance with ANSI T1.403.

The extracted data is used to compare with the target IB code. The length of the target activate/deactivate IB code can be 3 to 8 bits, as determined by the IBAL[1:0]/IBDL[1:0] bits (b3~2/b1~0, IBL,...). The content of the target activate/deactivate IB code is programmed in the IBA[7:0]/IBD[7:0] bits (b7~0, IBDA/IBDD,...). Refer to Figure-22.

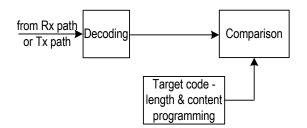


Figure-22 IB Detection

During comparison, if the extracted data coincides with the target activate/deactivate IB code with no more than 10^{-2} bit error rate for a certain period, the IB code is detected. The period depends on the setting of the AUTOLP bit (b3, LOOP,...).

If the AUTOLP bit (b3, LOOP,...) is '0', Automatic Digital/Remote Loopback is disabled. In this case, when the activate IB code is detected for more than 40 ms, the IBA_S bit (b1, STAT1,...) will be set to indicate the activate IB code detection; when the deactivate IB code is detected for more than 40 ms (T1/J1 mode) / 30 ms (E1 mode), the IBD_S bit (b0, STAT1,...) will be set to indicate the deactivate IB code detection.

If the AUTOLP bit (b3, LOOP,...) is '1', Automatic Digital/Remote Loopback is enabled. In this case, when the activate IB code is detected for more than 5.1 seconds, the IBA_S bit (b1, STAT1,...) will be set to indicate the activate IB code detection. The detection of the activate IB code in the receive path will activate Remote Loopback or the detection of the activate IB code in the transmit path will activate Digital Loopback (refer to Section 3.5.7.2 Remote Loopback & Section 3.5.7.3 Digital Loopback). When the deactivate IB code is detected for more than 5.1 seconds, the IBD_S bit (b0, STAT1,...) will be set to indicate the deactivate IB code detection. The detection of the deactivate IB code in the receive path will deactivate Remote Loopback or the detection of the deactivate IB code in the transmit path will be set to indicate the deactivate IB code in the transmit path will deactivate IB code in the receive path will deactivate Remote Loopback or the detection of the deactivate IB code in the transmit path will deactivate IB code in the receive path will deactivate Remote Loopback or the detection of the deactivate IB code in the transmit path will deactivate Digital Loopback (refer to Section 3.5.7.2 Remote Loopback & Section 3.5.7.3 Digital Loopback).

A transition from '0' to '1' on the IBA_S/IBD_S bit (b1/b0, STAT1,...) or any transition (from '0' to '1' or from '1' to '0') on the IBA_S/IBD_S bit (b1/b0, STAT1,...) will set the IBA_IS/IBD_IS bit (b1/b0, INTS1,...) to '1'

respectively, as selected by the IB_IES bit (b0, INTES,...). When the IBA_IS/IBD_IS bit (b1/b0, INTS1,...) is '1', an interrupt will be reported on INT if not masked by the IBA_IM/IBD_IM bit (b1/b0, INTM1,...).

3.5.6 ERROR COUNTER

An internal 16-bit Error Counter is used to count one of the following errors:

- LBPV: BPV/CV detected in the receive path (line side);
- LEXZ: EXZ detected in the receive path (line side);
- LBPV + LEXZ: BPV/CV and EXZ detected in the receive path (line side);
- SBPV: BPV/CV detected in the transmit path (system side) (disabled in Transmit Single Rail NRZ Format mode);
- SEXZ: EXZ detected in the transmit path (system side);
- SBPV + SEXZ: BPV/CV and EXZ detected in the transmit path (system side) (disabled in Transmit Single Rail NRZ Format mode);
- PRBS/ARB error.

The CNT_SEL[2:0] bits (b4~2, ERR,...) select one of the above errors to be counted.

The Error Counter is buffered. It is updated automatically or manually, as determined by the CNT_MD bit (b1, ERR,...).

The Error Counter is accessed by reading the ERRCH and ERRCL registers.

3.5.6.1 Automatic Error Counter Updating

When the CNT_MD bit (b1, ERR,...) is '1', the Error Counter is updated every one second automatically.

The one-second timer uses MCLK as clock reference. The expiration of each one second will set the TMOV_IS bit (b0, INTTM) and induce an interrupt reported by INT if not masked by the TMOV_IM bit (b0, GCF).

When each one second expires, the Error Counter transfers the accumulated error numbers to the ERRCH and ERRCL registers and the Error Counter will be cleared to start a new round counting. The ERRCH and ERRCL registers should be read in the next second, otherwise they will be overwritten.

When the ERRCH and ERRCL registers are all '1's and there is still error to be accumulated, the registers will be overflowed. The overflow is indicated by the CNTOV_IS bit (b0, INTS2,...) and will induce an interrupt reported by INT if not masked by the CNTOV_IM (b0, INTM2,...).

The process of automatic Error Counter updating is illustrated in Figure-23.

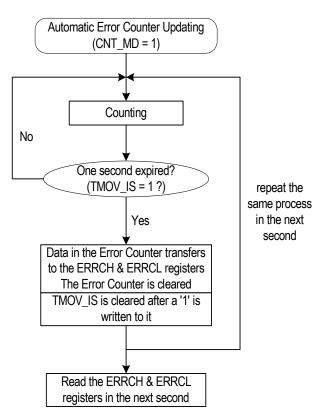


Figure-23 Automatic Error Counter Updating

3.5.6.2 Manual Error Counter Updating

When the CNT_MD bit (b1, ERR,...) is '0', the Error Counter is updated manually.

When there is a transition from '0' to '1' on the CNT_STOP bit (b0, ERR,...), the Error Counter transfers the accumulated error numbers to the ERRCH and ERRCL registers and the Error Counter will be cleared to start a new round counting. The ERRCH and ERRCL registers should be read in the next round of error counting, otherwise they will be overwritten.

When the ERRCH and ERRCL registers are all '1's and there is still error to be accumulated, the registers will be overflowed. The overflow is indicated by the CNTOV_IS bit (b0, INTS2,...) and will induce an interrupt reported by INT if not masked by the CNTOV_IM (b0, INTM2,...).

The process of manual Error Counter updating is illustrated in Figure-24.

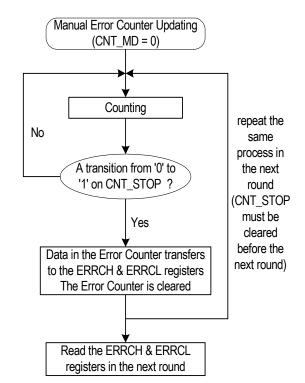


Figure-24 Manual Error Counter Updating



16-CHANNEL SHORT HAUL T1/E1/J1 LINE INTERFACE UNIT

3.5.7 LOOPBACK

There are four kinds of loopback:

- Analog Loopback
- Remote Loopback
- Digital Loopback

Refer to Figure-1 for loopback location.

3.5.7.1 Analog Loopback

Analog Loopback is enabled by the ALP bit (b0, LOOP,...). The data stream to be transmitted on the TTIPn/TRINGn pins is internally looped to the RTIPn/RRINGn pins.

In Analog Loopback mode, the data stream to be transmitted is still output to the line side, while the data stream received from the line side is covered by the Analog Loopback data.

Anytime when Analog Loopback is set, the other loopbacks (i.e., Digital Loopback and Remote Loopback) are disabled.

In Analog Loopback, the priority of the diagnostic facilities in the receive path is: pattern generation > looped data. AIS generation is disabled in both the receive path and the transmit path. Refer to Figure-25.

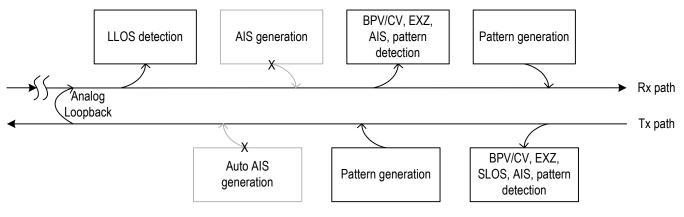


Figure-25 Priority Of Diagnostic Facilities During Analog Loopback

3.5.7.2 Remote Loopback

Remote Loopback can be configured manually or automatically. Either manual Remote Loopback configuration or automatic Remote Loopback configuration will enable Remote Loopback.

Manual Remote Loopback is enabled by the RLP bit (b1, LOOP,...).

Automatic Remote Loopback is enabled when the pattern detection is assigned in the receive path (i.e., the PD_POS bit (b3, PD,...) is '0') and the AUTOLP bit (b3, LOOP,...) is '1'. The corresponding channel will enter Remote Loopback when the activate IB code is detected in the receive path for more than 5.1 sec.; and will return from Remote Loopback when the deactivate IB code is detected in the receive path for more than 5.1 sec. Refer to section Inband Loopback (IB) Detection on page 40 for details. When automatic Remote Loopback is active, setting the AUTOLP bit (b3, LOOP,...) back to '0' will also stop automatic Remote Loopback. The setting of the PD_POS bit (b3, PD,...) should not be changed during automatic Remote Loopback. The AUTOLP_S bit (b7, STAT0,...) indicates the automatic Remote Loopback status.

In Remote Loopback mode, the data stream output from the RJA (if enabled) is internally looped to the Waveform Shaper. The data stream received from the line side is still output to the system side, while the data stream input from the system side is covered by the Remote Loopback data and the status on TCLKn does not affect the Remote Loopback. However, the BPV/CV, EXZ, SLOS, AIS and pattern detection in the transmit path still monitors the data stream input from the system side.

In Remote Loopback mode, the priority of the diagnostic facilities in the receive path is: pattern generation > AIS generation; the priority of the diagnostic facilities in the transmit path is: pattern generation > looped data. AIS generation is disabled in the transmit path. Refer to Figure-26.

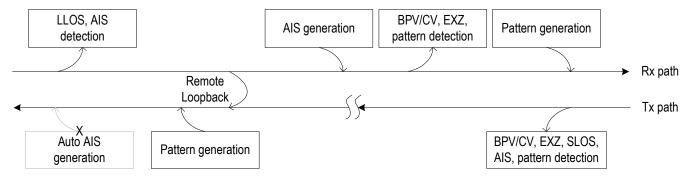


Figure-26 Priority Of Diagnostic Facilities During Manual Remote Loopback

DT82P20416

The Digital Loopback can be configured manually or automatically. Either manual Digital Loopback configuration or automatic Digital Loopback configuration will enable Digital Loopback.

Manual Digital Loopback is enabled by the DLP bit (b2, LOOP,...).

Automatic Digital Loopback is enabled when the pattern detection is assigned in the transmit path (i.e., the PD_POS bit (b3, PD,...) is '1') and the AUTOLP bit (b3, LOOP,...) is '1'. The corresponding channel will enter Digital Loopback when the activate IB code is detected in the transmit path for more than 5.1 sec.; and will return from Digital Loopback when the deactivate IB code is detected in the transmit path for more than 5.1 sec.; and will return from Digital Loopback when the deactivate IB code is detected in the transmit path for more than 5.1 sec. Refer to section Inband Loopback (IB) Detection on page 40 for details. When automatic Digital Loopback is active, setting the AUTOLP bit (b3, LOOP,...) back to '0' will also stop automatic Digital

Loopback. The setting of the PD_POS bit (b3, PD,...) should not be changed during automatic Digital Loopback. The AUTOLP_S bit (b7, STAT0,...) indicates the automatic Digital Loopback status.

In Digital Loopback mode, the data stream output from the TJA (if enabled) is internally looped to the Decoder (if enabled). The data stream to be transmitted is still output to the line side, while the data stream received from the line side is covered by the Digital Loopback data. However, LLOS and AIS detection in the receive path still monitors the data stream received from the line side.

In Digital Loopback mode, the priority of the diagnostic facilities in the receive path is: pattern generation > looped data; the priority of the diagnostic facilities in the transmit path is: pattern generation > looped data > AIS generation. AIS generation is disabled in the receive path.

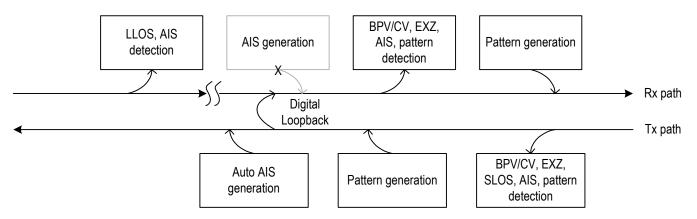


Figure-27 Priority Of Diagnostic Facilities During Digital Loopback

RENESAS

16-CHANNEL SHORT HAUL T1/E1/J1 LINE INTERFACE UNIT

3.5.8 CHANNEL 0 MONITORING

Channel 0 can be used as a monitoring channel when not used as a regular channel. Channel 0 support G.772 Monitoring and Jitter Measurements.

3.5.8.1 G.772 Monitoring

Selected by the MON[5:0] bits (b5~0, MON), any receiver or transmitter of the other 15 channels can be monitored by channel 0 (as shown in Figure-28).

When the G.772 Monitoring is implemented (the MON[5:0] bits (b5~0, MON) is not '0'), the registers of the receiver of channel 0 should be the same as those of the selected receiver /transmitter except the line interface related registers.

Once the G.772 Monitoring is implemented, the receiver of channel 0 switches to External Impedance Matching mode automatically, and the setting in the R_TERM[2:0] bits (b2~0, RCF0,...) of channel 0 is ignored.

During the G.772 Monitoring, channel 0 processes as normal after data is received from the selected path and the operation of the monitored path is not effected.

The signal which is monitored goes through the Clock & Data Recovery of monitoring channel (channel 0). The monitored clock can output on RCLK0. The monitored data can be observed digitally on the output pin of RCLK0, RD0/RDP0 and RDN0. LOS detector is still in use in channel 0 for the monitored signal.

In monitoring mode, channel 0 can be configured to Remote Loopback. The signal which is being monitored will output on TTIP0 and TRING0. The output signal can then be connected to a standard test equipment for non-intrusive monitoring.

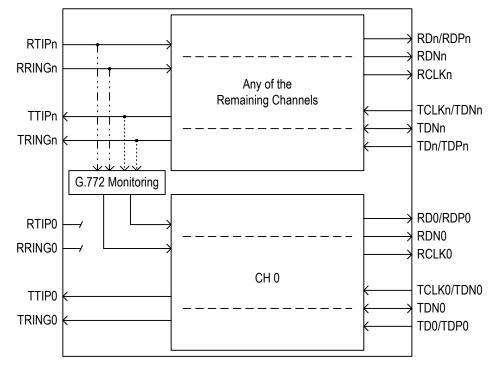


Figure-28 G.772 Monitoring

3.5.8.2 Jitter Measurement (JM)

DT82P20416

The RJA of channel 0 consists of a Jitter Measurement (JM) module. When the RJA is enabled in channel 0, the JM is used to measure the positive and negative peak value of the demodulated jitter signal of the received data stream. The bandwidth of the measured jitter is selected by the JM_BW bit (b0, JM).

The greatest positive peak value monitored in a certain period is indicated by the JIT_PH and JIT_PL registers, while the greatest negative peak value monitored in the same period is indicated by the JIT_NH and JIT_NL registers. The relationship between the greatest positive /negative peak value and the indication in the corresponding registers is:

Positive Peak = [JIT_PH, JIT_PL] / 16 (UIpp);

Negative Peak = [JIT_NH, JIT_NL] / 16 (Ulpp).

The period is determined by the JM_MD bit (b1, JM).

When the JM_MD bit (b1, JM) is '1', the period is one second automatically. The one-second timer uses MCLK as clock reference. The expiration of each one second will set the TMOV_IS bit (b0, INTTM) and induce an interrupt reported by INT if not masked by the TMOV_IM bit (b0, GCF). The TMOV_IS bit (b0, INTTM) is cleared after a '1' is written to this bit. When each one second expires, internal buffers transfer the greatest positive/negative peak value accumulated in this one second to the JIT_PH and JIT_PL / JIT_NH and JIT_NL registers respectively and the internal buffers will be cleared to start a new round measurement. The registers should be read in the next second, otherwise they will be overwritten. Refer to Figure-29 for the process.

When the JM_MD bit (b1, JM) is '0', the period is controlled by the JM_STOP bit (b2, JM) manually. When there is a transition from '0' to '1' on the JM_STOP bit (b2, JM), the internal buffers transfer the greatest positive/negative peak value accumulated in this period to the JIT_PH and JIT_PL / JIT_NH and JIT_NL registers respectively and the internal buffers will be cleared to start a new round measurement. The registers should be read in the next round of jitter measurement, otherwise they will be overwritten. Refer to Figure-30 for the process.

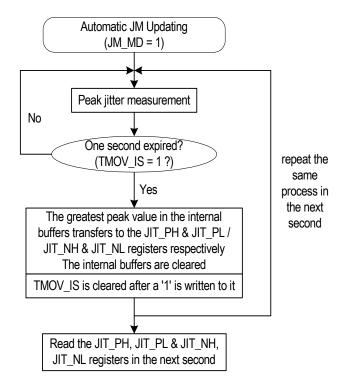


Figure-29 Automatic JM Updating

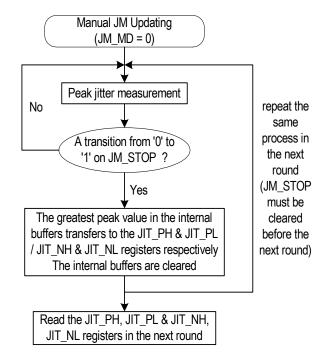


Figure-30 Manual JM Updating



3.6 CLOCK INPUTS AND OUTPUTS

The IDT82P20416 provides two kinds of clock outputs:

• Free running clock outputs on CLKT1 and CLKE1

The following Clock Input is provided:

• MCLK as programmable reference timing for the IDT82P20416.

3.6.1 FREE RUNNING CLOCK OUTPUTS ON CLKT1/CLKE1

An internal clock generator uses MCLK as reference to generate all the clocks required by internal circuits and CLKT1/CLKE1 outputs. MCLK should be a clock with +/-32 ppm (in T1/J1 mode) or +/-50 ppm (in E1 mode) accuracy. The clock frequency of MCLK is 1.544/2.048 X N MHz ($1 \le N \le 8$, N is an integer number), as determined by MCKSEL[3:0]. Refer to Chapter 2 Pin Description for details.

The outputs on CLKT1 and CLKE1 are free running (locking to MCLK). The output of CLKT1 is determined by the CLKT1_EN bit (b1, CLKG) and the CLKT1 bit (b0, CLKG). Refer to Table-21. The output of CLKE1 is determined by the CLKE1_EN bit (b3, CLKG) and the CLKE1 bit (b2, CLKG). Refer to Table-22.

Table-21 Clock Output on CLKT1

Contr	ol Bits	Clock Output On CLKT1
CLKT1_EN	CLKT1	
0	(don't-care)	High-Z
1	0	8 KHz
1	1	1.544 KHz

Table-22 Clock Output on CLKE1

Contro	ol Bits	Clock Output On CLKE1
CLKE1_EN	CLKE1	olock output on other
0	(don't-care)	High-Z
1	0	8 KHz
I	1	2.048 KHz

3.6.2 MCLK, MASTER CLOCK INPUT

MCLK provides a stable reference timing for the IDT82P20416. MCLK should be a clock with +/-32 ppm (in T1/J1 mode) or +/-50 ppm (in E1 mode) accuracy. The clock frequency of MCLK is set by pins MCKSEL[3:0] and can be N x 1.544 MHz or N x 2.048 MHz with $1 \le N \le 8$ (N is an integer number). Refer to MCKSEL[3:0] pin description for details.

If there is a loss of MCLK (duty cycle is less than 30% for 10 μ s), the device will enter power down. In this case, both the receive and transmit circuits are turned off. The pins on the line interface will be in High-Z state. The pins on receive system interface will be in High-Z state or in low level, as selected by the RHZ bit (b6, RCF0,...). The input on the

transmit system interface is ignored and the output on the transmit system interface will be in High-Z state. Refer to Section 3.2.7 Receiver Power Down and Section 3.3.7 Transmitter Power Down for details.

If MCLK recovers after loss of MCLK the device will be reset automatically.

3.6.3 XCLK, INTERNAL REFERENCE CLOCK INPUT

XCLK is derived from MCLK. For the respective channel, it is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode. XCLK is used as selectable reference clock for

- pattern /AIS generation
- RCLKn in LLOS
- · Loss of TCLKn to determine Transmit Output High-Z.

3.7 INTERRUPT SUMMARY

There are altogether 20 kinds of interrupt sources as listed in Table-23. Among them, No.1 to No.19 are per-channel interrupt sources, while No. 20 is a global interrupt source.

For interrupt sources from No.1 to No.10, the occurrence of the event will cause the corresponding Status bit to be set to '1'. And selected by the Interrupt Trigger Edges Select bit, either a transition from '0' to '1' or any transition from '0' to '1' or from '1' to '0' of the Status bit will cause the Interrupt Status bit to be set to '1', which indicates the occurrence of an interrupt event.

For interrupt sources from No.11 to No.20, the occurrence of the event will cause the corresponding Interrupt Status Bit to be set to '1'.

All the interrupt can be masked by the GLB_IM bit (b1, GCF) globally or by the corresponding interrupt mask bit individually. For all the interrupt sources, if not masked, the occurrence of the interrupt event will trigger an interrupt indicated by the INT pin. For per-channel interrupt sources, if not masked, the occurrence of the interrupt event will also cause the corresponding INT_CHn bit (INTCH1~3) to be set '1'.

An interrupt event is cleared by writing '1' to the corresponding Interrupt Status bit. The INT_CHn bit (INTCH1~3) will not be cleared until all the interrupts in the corresponding channel are acknowledged. The INT pin will be inactive until all the interrupts are acknowledged. Refer to Figure-31 for interrupt service flow.

Table-23 Interrupt Summary

No.	Interrupt Source	Status Bit	Interrupt Trigger Edges Select Bit	Interrupt Status Bit	Interrupt Mask Bit
1	TCLKn is missing.	TCKLOS_S (b3, STAT0,)	TCKLOS_IES (b3, INTES,)	TCKLOS_IS (b3, INTS0,)	TCKLOS_IM (b3, INTM0,)
2	LLOS is detected.	LLOS_S (b0, STAT0,)	LOS_IES (b1, INTES,)	LLOS_IS (b0, INTS0,)	LLOS_IM (b0, INTM0,)
3	SLOS is detected.	SLOS_S (b1, STAT0,)	LOS_IES (b1, INTES,)	SLOS _IS (b1, INTS0,)	SLOS_IM (b1, INTM0,)
4	TLOS is detected.	TLOS_S (b2, STAT0,)	TLOS_IES (b2, INTES,)	TLOS_IS (b2, INTS0,)	TLOS_IM (b2, INTM0,)
5	LAIS is detected.	LAIS_S (b6, STAT1,)	AIS_IES (b6, INTES,)	LAIS_IS (b6, INTS1,)	LAIS_IM (b6, INTM1,)
6	SAIS is detected.	SAIS_S (b7, STAT1,)	AIS_IES (b6, INTES,)	SAIS_IS (b7, INTS1,)	SAIS_IM (b7, INTM1,)
7	TOC is detected.	TOC_S (b4, STAT0,)	TOC_IES (b4, INTES,)	TOC_IS (b4, INTS0,)	TOC_IM (b4, INTM0,)
8	The PRBS/ARB pattern is detected syn- chronized.	PA_S (b5, STAT1,)	PA_IES (b5, INTES,)	PA_IS (b5, INTS1,)	PA_IM (b5, INTM1,)
9	Activate IB code is detected.	IBA_S (b1, STAT1,)	IB_IES (b0, INTES,)	IBA_IS (b1, INTS1,)	IBA_IM (b1, INTM1,)
10	Deactivate IB code is detected.	IBD_S (b0, STAT1,)	IB_IES (b0, INTES,)	IBD_IS (b0, INTS1,)	IBD_IM (b0, INTM1,)
11	The FIFO of the RJA is overflow or underflow.	-	-	RJA_IS (b5, INTS0,)	RJA_IM (b5, INTM0,)
12	The FIFO of the TJA is overflow or underflow.	-	-	TJA_IS (b6, INTS0,)	TJA_IM (b6, INTM0,)
13	Waveform amplitude is overflow.	-	-	DAC_IS (b7, INTS0,)	DAC_IM (b7, INTM0,)
14	SBPV is detected.	-	-	SBPV_IS (b5, INTS2,)	SBPV_IM (b5, INTM2,)
15	LBPV is detected.	-	-	LBPV_IS (b4, INTS2,)	LBPV_IM (b4, INTM2,)
16	SEXZ is detected.	-	-	SEXZ_IS (b3, INTS2,)	SEXZ_IM (b3, INTM2,)
17	LEXZ is detected.	-	-	LEXZ_IS (b2, INTS2,)	LEXZ_IM (b2, INTM2,)
18	PRBS/ARB error is detected.	-	-	ERR_IS (b1, INTS2,)	ERR_IM (b1, INTM2,)
19	The ERRCH and ERRCL registers are overflowed.	-	-	CNTOV_IS (b0, INTS2,)	CNTOV_IM (b0, INTM2,)
20	One second time is over.	-	-	TMOV_IS (b0, INTTM)	TMOV_IM (b0, GCF)



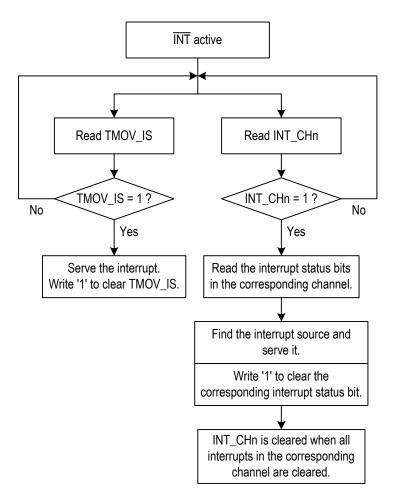


Figure-31 Interrupt Service Process

4 MISCELLANEOUS

4.1 RESET

The reset operation resets all registers, state machines as well as I/O pins to their default value or status.

The IDT82P20416 provides 4 kinds of reset:

- Power-on reset;
- · Hardware reset;
- · Global software reset;
- Per-channel software reset.

The Power-on, Hardware and Global software reset operations reset all the common blocks (including clock generator/synthesizer and microprocessor interface) and channel-related parts. The Per-channel software reset operation resets the channel-related parts. Figure-32 shows a general overview of the reset options.

During reset, all the line interface pins (i.e., TTIPn/TRINGn and RTIPn/RRINGn) are in High-Z state.

After reset, all the items listed in Table-24 are true.

Table-24 After Reset Effect Summary

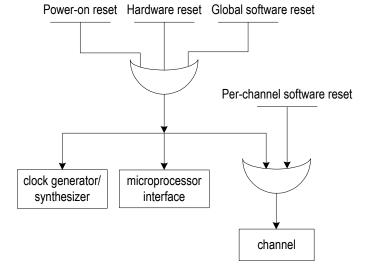


Figure-32 Reset

Effect On	Power-On Reset, Hardware Reset and Global Software Reset	Per-Channel Software Reset
TTIPn/TRINGn & RTIPn/ RRINGn	All TTIPn/TRINGn & RTIPn/RRINGn pins are in High-Z state.	Only TTIPn/TRINGn & RTIPn/RRINGn in the corresponding chan- nel are in High-Z.
Line Interface Mode	All channels are reset to T1/J1 mode.	Only the corresponding channel is reset to T1/J1 mode.
System interface	All channels are in Dual Rail NRZ Format.	Only the corresponding channel is in Dual Rail NRZ Format.
General I/O pins (i.e., D[7:0] and GPIO[1:0])	As input pins.	(No effect)
INT	Open drain output.	(No effect)
CLKT1, CLKE1	Output enable.	(No effect)
LLOS, LLOS0	Output enable.	(No effect)
TDO, SDO/ACK/RDY	High-Z.	(No effect)
state machines	All state machines are reset.	The state machines in the corresponding channel are reset.
Interrupt sources	All interrupt sources are masked.	The interrupt sources in the corresponding channel are masked.
Registers	All registers are reset to their default value.	The registers in the corresponding channel are reset to their default value except that there is no effect on the T1E1 bit.

4.1.1 POWER-ON RESET

Power-on reset is initiated during power-up. When all VDD inputs (1.8V and 3.3V) reach approximately 60% of the standard value of VDD, power-on reset begins. If MCLK is applied, power-on reset will complete within 1 ms maximum; if MCLK is not applied, the device remains in reset state.

4.1.2 HARDWARE RESET

Pulling the \overline{RST} pin to low will initiate hardware reset. The reset cycle should be more than 1 µs. If the \overline{RST} pin is held low continuously, the device remains in reset state.

4.1.3 GLOBAL SOFTWARE RESET

Writing the RST register will initiate global software reset. Once initiated, global software reset completes in 1 μs maximum.

4.1.4 PER-CHANNEL SOFTWARE RESET

Writing a '1' to the CHRST bit (b1, CHCF,...) will initiate per-channel software reset. Once initiated, per-channel software reset completes in 1 µs maximum and the CHRST bit (b1, CHCF,...) is self cleared.

This reset is different from other resets, for:

- It does not reset the T1E1 bit (b0, CHCF,...). That is, the operation mode of each channel is not changed;
- It does not reset the global registers, state machines and common pins (including the pins of clock generator, microprocessor interface and JTAG interface);
- · It does not reset the other channels.

4.2 MICROPROCESSOR INTERFACE

The microprocessor interface provides access to read and write the registers in the device. The interface consists of:

· Serial microprocessor interface;

4.3 POWER UP

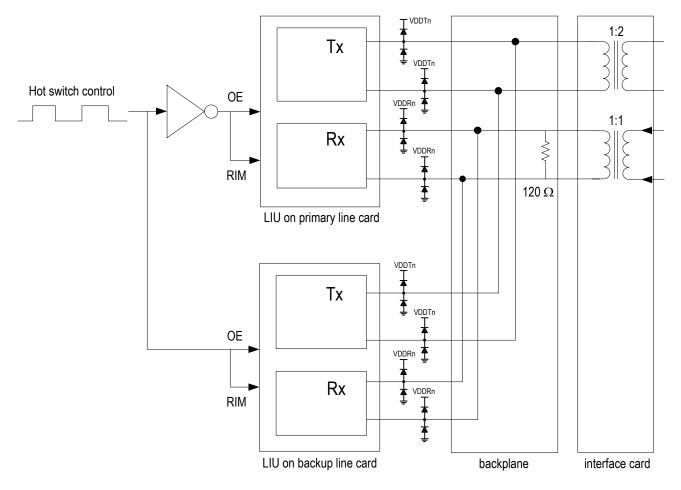
No power up sequencing for the VDD inputs (1.8 V and 3.3 V) has to be provided for the IDT82P20416. A Power-on reset will be initiated during power up. Refer to Section 4.1 Reset.

4.4 HITLESS PROTECTION SWITCHING (HPS) SUM-MARY

In today's telecommunication systems, ensuring no traffic loss is becoming increasingly important. To combat these problems, redundancy protection must be built into the systems carrying this traffic. There are many types of redundancy protection schemes, including 1+1 and 1:1 hardware protection without the use of external relays. Refer to Figure-33, Figure-34 and Figure-35 for different protection schemes. The IDT82P20416 provides an enhanced architecture to support both protection schemes.

IDT82P20416 highlights for HPS support:

- Independent programmable receive and transmit high impedance for Tip and Ring inputs and outputs to support 1+1 and 1:1 redundancy
- Fully integrated receive termination, required to support 1:1 redundancy
- Enhanced internal architecture to guarantee High Impedance for Tip and Ring Inputs and Outputs during Power Off or Power Failure
- Asynchronous hardware control (OE, RIM) for fast global high impedance of receiver and transmitter (hot switching between working and backup board)

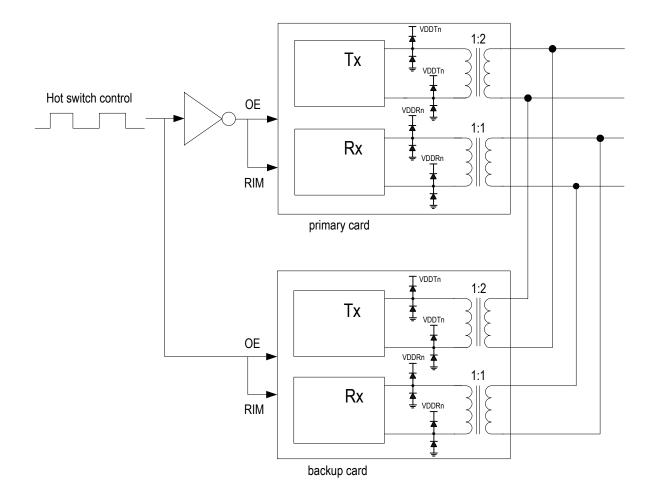


Rx: Partially Internal Impedance Matching mode. A fixed external 120 Ω resistor is placed on the backplane and provides a common termination for T1/J1/E1 applications. The R_TERM[2:0] bits (b2~0, RCF0,...) setting is as follows: '000' for T1 100 Ω twisted pair cable, '001' for J1 110 Ω twisted pair cable, '001' for E1 120 Ω twisted pair cable and '011' for E1 75 Ω coaxial cable.

Tx: Internal Impedance Matching mode. The T_TERM[2:0] bits (b2~0, TCF0,...) setting is as follows: '000' for T1 100 Ω twisted pair cable, '001' for J1 110 Ω twisted pair cable, '010' for E1 120 Ω twisted pair cable and '011' for E1 75 Ω coaxial cable.





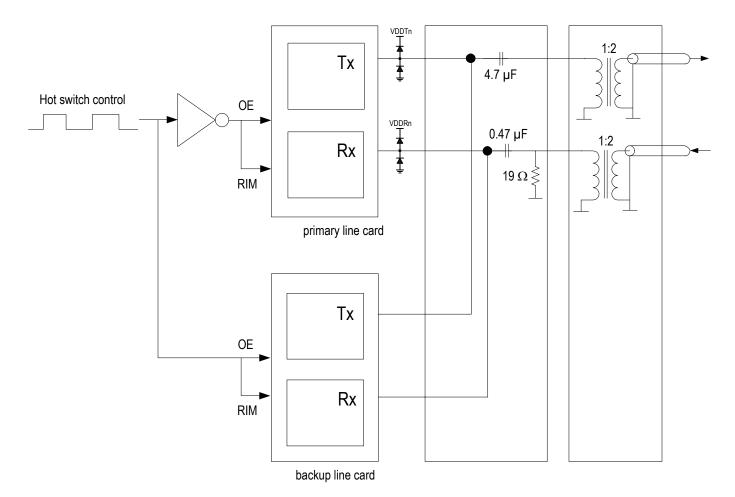


Rx: Fully Internal Impedance Matching mode. In this mode, there is no external resistor required. The R_TERM[2:0] bits (b2~0, RCF0,...) setting is as follows: '000' for T1 100 Ω twisted pair cable, '001' for J1 110 Ω twisted pair cable, '010' for E1 120 Ω twisted pair cable and '011' for E1 75 Ω coaxial cable.

Tx: Internal Impedance Matching mode. The T_TERM[2:0] bits (b2~0, TCF0,...) setting is as follows: '000' for T1 100 Ω twisted pair cable, '001' for J1 110 Ω twisted pair cable, '010' for E1 120 Ω twisted pair cable and '011' for E1 75 Ω coaxial cable.

Figure-34 1:1 HPS Scheme, Differential Interface (Individual Transformer)





Rx: 75 Ω External Impedance Matching mode. In this mode, there is no external resistor required. The RIM pin should be left open and the configuration of the R_TERM[2:0] bits (b2~0, RCF0,...) is ignored.

Tx: 75 Ω Internal Impedance Matching mode. The T_TERM[2:0] bits (b2~0, TCF0,...) should be set to '011'.

Figure-35 1+1 HPS Scheme, E1 75 ohm Single-Ended Interface (Shared Common Transformer)

5 PROGRAMMING INFORMATION

5.1 REGISTER MAP

5.1.1 GLOBAL REGISTER

Address (Hex)	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reference Page
Common	Control		I	1	1	1		1		
000	ID - Device ID Register	ID7	ID6	ID5	ID4	ID3	ID2	ID1	ID0	P 60
040	RST - Global Reset Register	RST7	RST6	RST5	RST4	RST3	RST2	RST1	RST0	P 60
080	GCF - Global Configuration Register	-	-	-	COPY	INT_PIN1	INT_PIN0	GLB_IM	TMOV_IM	P 61
0C0	MON - G.772 Monitor Configura- tion Register	-	-	MON5	MON4	MON3	MON2	MON1	MON0	P 61
100	GPIO - General Purpose I/O Pin Definition Register	-	-	-	-	LEVEL1	LEVEL0	DIR1	DIR0	P 62
Reference	e Clock Timing Option									
1C0	CLKG - CLKT1 & CLKE1 Gener- ation Control Register	-	-	-	-	CLKE1_EN	CLKE1	CLKT1_EN	CLKT1	P 63
Interrupt I	Indication		L			•		•		
2C0	INTCH1 - Interrupt Requisition Source Register 1	INT_CH8	INT_CH7	INT_CH6	INT_CH5	INT_CH4	INT_CH3	INT_CH2	INT_CH1	P 63
300	INTCH2 - Interrupt Requisition Source Register 2	-	INT_CH15	INT_CH14	INT_CH13	INT_CH12	INT_CH11	INT_CH10	INT_CH9	P 64
380	INTCH3 - Interrupt Requisition Source Register 3	INT_CH0	-	-	-	-	-	-	-	P 64
3C0	INTTM - One Second Timer Interrupt Status Register	-	-	-	-	-	-	-	TMOV_IS	P 64

5.1.2 PER-CHANNEL REGISTER

Only the address of channel 1 is listed in the 'Address (Hex)' column of the following table. For the addresses of the other channels, refer to the description of each register.

Address (Hex)	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reference Page
Channel	Control									
001	CHCF - Channel Configuration Register	-	-	-	-	-	-	CHRST	T1E1	P 65
JA Config	guration									
002	TJA - Transmit Jitter Attenuation Configuration Register	-	-	-	TJA_LIMT	TJA_EN	TJA_DP1	TJA_DP0	TJA_BW	P 65
003	RJA - Receive Jitter Attenuation Configuration Register	-	-	-	RJA_LIMT	RJA_EN	RJA_DP1	RJA_DP0	RJA_BW	P 66
Transmit	Path Configuration									
004	TCF0 - Transmit Configuration Register 0	-	OE	T_OFF	THZ_OC	T_SING	T_TERM2	T_TERM1	T_TERM0	P 67
005	TCF1 - Transmit Configuration Register 1	-	-	-	TCK_ES	TD_INV	T_CODE	T_MD1	T_MD0	P 68
006	PULS - Transmit Pulse Configu- ration Register	-	-	-	-	PULS3	PULS2	PULS1	PULS0	P 69
007	SCAL - Amplitude Scaling Con- trol Register	-	-	SCAL5	SCAL4	SCAL3	SCAL2	SCAL1	SCAL0	P 70
008	AWG0 - Arbitrary Waveform Generation Control Register 0	-	DONE	RW	SAMP4	SAMP3	SAMP2	SAMP1	SAMP0	P 70
009	AWG1 - Arbitrary Waveform Generation Control Register 1	-	WDAT6	WDAT5	WDAT4	WDAT3	WDAT2	WDAT1	WDAT0	P 71
Receive I	Path Configuration									
00A	RCF0 - Receive Configuration Register 0	RCKH	RHZ	R_OFF	R120IN	R_SING	R_TERM2	R_TERM1	R_TERM0	P 72
00B	RCF1 - Receive Configuration Register 1	-	-	-	RCK_ES	RD_INV	R_CODE	R_MD1	R_MD0	P 73
00C	RCF2 - Receive Configuration Register 2	-	-	-	-	-	-	MG1	MG0	P 73
Diagnosti	ics			1		1			1	
00D	LOS - LOS Configuration Regis- ter	LAC	ALOS2	ALOS1	ALOS0	TALOS1	TALOS0	TDLOS1	TDLOS0	P 74
00E	ERR - Error Detection & Inser- tion Control Register	EXZ_DEF	BPV_INS	ERR_INS	CNT_SEL2	CNT_SEL1	CNT_SEL0	CNT_MD	CNT_STOP	P 75
00F	AISG - AIS Generation Control Register	-	-	-	TXAIS	ASAIS_SL OS	ASAIS_LLO S	ALAIS_SLO S	ALAIS_LLO S	P 76
010	PG - Pattern Generation Control Register	-	PG_CK	PG_EN1	PG_EN0	PG_POS	PAG_INV	PRBG_SEL 1	PRBG_SEL 0	P 77
	•			•		•			•	

16-CHANNEL SHORT HAUL T1/E1/J1 LINE INTERFACE UNIT

Address (Hex)	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reference Page
011	PD - Pattern Detection Control Register	-	-	-	-	PD_POS	PAD_INV	PAD_SEL1	PAD_SEL0	P 78
012	ARBL - Arbitrary Pattern Gener- ation / Detection Low-Byte Reg- ister	ARB7	ARB6	ARB5	ARB4	ARB3	ARB2	ARB1	ARB0	P 79
013	ARBM - Arbitrary Pattern Gen- eration / Detection Middle-Byte Register	ARB15	ARB14	ARB13	ARB12	ARB11	ARB10	ARB9	ARB8	P 79
014	ARBH - Arbitrary Pattern Gener- ation / Detection High-Byte Reg- ister	ARB23	ARB22	ARB21	ARB20	ARB19	ARB18	ARB17	ARB16	P 79
015	IBL - Inband Loopback Control Register	-	-	IBGL1	IBGL0	IBAL1	IBAL0	IBDL1	IBDL0	P 80
016	IBG - Inband Loopback Genera- tion Code Definition Register	IBG7	IBG6	IBG5	IBG4	IBG3	IBG2	IBG1	IBG0	P 80
017	IBDA - Inband Loopback Detec- tion Target Activate Code Defini- tion Register	IBA7	IBA6	IBA5	IBA4	IBA3	IBA2	IBA1	IBA0	P 81
018	IBDD - Inband Loopback Detec- tion Target Deactivate Code Definition Register	IBD7	IBD6	IBD5	IBD4	IBD3	IBD2	IBD1	IBD0	P 81
019	LOOP - Loopback Control Reg- ister	-	-	-	-	AUTOLP	DLP	RLP	ALP	P 82
Interrupt I	Edge Selection									
01A	INTES - Interrupt Trigger Edges Select Register	-	AIS_IES	PA_IES	TOC_IES	TCKLOS_I ES	TLOS_IES	LOS_IES	IB_IES	P 83
Interrupt I	Mask									
01B	INTM0 - Interrupt Mask Register 0	DAC_IM	TJA_IM	RJA_IM	TOC_IM	TCKLOS_I M	TLOS_IM	SLOS_IM	LLOS_IM	P 84
01C	INTM1 - Interrupt Mask Register 1	SAIS_IM	LAIS_IM	PA_IM	-	-	-	IBA_IM	IBD_IM	P 85
01D	INTM2 - Interrupt Mask Register 2	-	-	SBPV_IM	LBPV_IM	SEXZ_IM	LEXZ_IM	ERR_IM	CNTOV_IM	P 86
Status Ind	dication									
01E	STAT0 - Status Register 0	AUTOLP_S	-	-	TOC_S	TCKLOS_S	TLOS_S	SLOS_S	LLOS_S	P 87
01F	STAT1 - Status Register 1	SAIS_S	LAIS_S	PA_S	-	-	-	IBA_S	IBD_S	P 88
Interrupt	Status Indication			•			-		•	
020	INTS0 - Interrupt Status Regis- ter 0	DAC_IS	TJA_IS	RJA_IS	TOC_IS	TCKLOS_I S	TLOS_IS	SLOS_IS	LLOS_IS	P 89
021	INTS1 - Interrupt Status Regis- ter 1	SAIS_IS	LAIS_IS	PA_IS	-	-	-	IBA_IS	IBD_IS	P 90
022	INTS2 - Interrupt Status Regis- ter 2	-	-	SBPV_IS	LBPV_IS	SEXZ_IS	LEXZ_IS	ERR_IS	CNTOV_IS	P 91

RENESAS

IDT82P20416

RENESAS

Address (Hex)	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reference Page
Counter										
023	ERRCL - Error Counter Low- Byte Register	ERRC7	ERRC6	ERRC5	ERRC4	ERRC3	ERRC2	ERRC1	ERRC0	P 92
024	ERRCH - Error Counter High- Byte Register	ERRC15	ERRC14	ERRC13	ERRC12	ERRC11	ERRC10	ERRC9	ERRC8	P 92
Jitter Mea	asurement (channel 0 Only)						•			
7E5	JM - Jitter Measurement Config- uration For Channel 0 Register	-	-	-	-	-	JM_STOP	JM_MD	JM_BW	P 93
7E6	JIT_PL - Positive Peak Jitter Measurement Low-Byte Regis- ter	JIT_P7	JIT_P6	JIT_P5	JIT_P4	JIT_P3	JIT_P2	JIT_P1	JIT_P0	P 93
7E7	JIT_PH - Positive Peak Jitter Measurement High-Byte Regis- ter	-	-	-	-	JIT_P11	JIT_P10	JIT_P9	JIT_P8	P 93
7E8	JIT_NL - Negative Peak Jitter Measurement Low-Byte Regis- ter	JIT_N7	JIT_N6	JIT_N5	JIT_N4	JIT_N3	JIT_N2	JIT_N1	JIT_N0	P 94
7E9	JIT_NH - Negative Peak Jitter Measurement High-Byte Regis- ter	-	-	-	-	JIT_N11	JIT_N10	JIT_N9	JIT_N8	P 94

5.2 REGISTER DESCRIPTION

5.2.1 GLOBAL REGISTER

ID - Device ID Register

Address: 000H Type: Read Default Value: 7×	Н							
7	6	5	4	3	2	1	0	
ID7	ID6	ID5	ID4	ID3	ID2	ID1	ID0	1
Bit	Name			Descri	ption			
7 - 0	ID[7:0]	The ID[7:0] bits are pre-s	set, where the ID[3:0]] bits 'x' represent th	e current version nu	ımber ('0000' is for t	he first version).	

RST - Global Reset Register

Type:	ess: 040H : Write ult Value: 00	Н						
	7	6	5	4	3	2	1	0
	RST7	RST6	RST5	RST4	RST3	RST2	RST1	RST0
	Bit	Name			Descri	ption		
	7 - 0	RST[7:0]	Writing this register will in	itiate global softwa	re reset. This reset co	ompletes in 1 µs ma	ximum.	

GCF - Global Configuration Register

T_PIN0	GLB_IM	TMOV_IM
es whether th	he written value is	s copied to the sam
es whether th	he written value is	s copied to the sam
es whether th	he written value is	s copied to the sam
		prresponding interru
	e generated.	rupt mask bit is '0' and the co e generated. (default) is over. This one second time

MON - G.772 Monitor Configuration Register

7	6	5	4	3	2	1	0
-	•	MON5	MON4	MON3	MON2	MON1	MON0
Bit	Name			Descrip	tion		
7 - 6	-	Reserved.					
5 - 0	MON[5:0]	These bits determine wh transmitter or receiver to 000000: No transmitter of 000001: The receiver of 000010: The receiver of 010000: The receiver of 010000: The receiver of 0100001 ~ 011111: Resen 100000: No transmitter of 100001: The transmitter 100010: The transmitter	be monitored by cha r receiver is monitore channel 1 is monitore channel 2 is monitor channel 15 is monitor channel 16 is monitor ved. r receiver is monitore of channel 1 is monitor	nnel 0. ed. (default) ed. ed. red. ed. ored.	When the G.772 Mor	ntor is implemented	I, these bits selec

101111: The transmitter of channel 15 is monitored.

110000 ~ 111111: Reserved.

GPIO - General Purpose I/O Pin Definition Register

- Bit	· ·				2	I	0
Rit		•	-	LEVEL1	LEVEL0	DIR1	DIR0
Dit	Name			Descri	ption		
7 - 4	-	Reserved.					
		0: Output low level. 1: Output high level. (def When the GPIO1 pin is d 0: Input low level. 1: Input high level. (defau	efined as input, this Ilt)			·	
2	LEVEL0	When the GPIO0 pin is d 0: Output low level. 1: Output high level. When the GPIO0 pin is d 0: Input low level. 1: Input high level. (defau	efined as input, this				
1	DIR1	This bit determines wheth 0: Output. 1: Input. (default)	ner the GPIO1 pin i	is used as output or in	nput.		

CLKG - CLKT1 & CLKE1 Generation Control Register

7	6	5	4	3	2	1	0
•	-	-	-	CLKE1_EN	CLKE1	CLKT1_EN	CLKT1
Bit	Name			Descript	tion		
7 - 4	-	Reserved.					
3	CLKE1_EN	This bit controls whether the output is disabled. 1: The output is enabled.	CLKE1 is in High-Z	state.	y the CLKE1 bit (b	2, CLKG). (default)	
2	CLKE1	This bit is valid only when 0: 8 KHz. 1: 2.048 MHz. (default)	the CLKE1_EN bit	(b3, CLKG) is '1'. Thi	s bit selects the clo	ock frequency output c	on the CLKE1 p
1	CLKT1_EN	This bit controls whether the output is disabled. 1: The output is enabled.	CLKT1 is in High-Z	state.	y the CLKT1 bit (b), CLKG). (default)	
0	CLKT1	This bit is valid only when 0: 8 KHz.					on the CLKT1 pi

INTCH1 - Interrupt Requisition Source Register 1

Address: 2 Type: Rea Default Va	ad	4						
	7	6	5	4	3	2	1	0
INT	T_CH8	INT_CH	I7 INT_CH6	INT_CH5	INT_CH4	INT_CH3	INT_CH2	INT_CH1
Bit	t	Name			Descri	ption		
7 - 0	0	INT_CH[8:1]	These bits indicate wheth channel 8 to 1 respectivel 0: No interrupt is generate 1: At least one interrupt is	ly. ed or all the interrupts	are cleared in the	corresponding chanr		[8:1] bits correspond to

INTCH2 - Interrupt Requisition Source Register 2

Address: 300H Type: Read Default Value: 0	ОH						
7	6	5	4	3	2	1	0
-	INT_CH	15 INT_CH14	INT_CH13	INT_CH12	INT_CH11	INT_CH10	INT_CH9
Bit	Name			Descri	ption		
7	-	Reserved.					
6 - 0	INT_CH[15:9]	These bits indicate wheth channel 16 to 9 respectiv 0: No interrupt is generate 1: At least one interrupt is	ely. ed or all the interrupt	s are cleared in the	corresponding chan	-	6:9] bits correspond to

INTCH3 - Interrupt Requisition Source Register 3

Address: 380H Type: Read Default Value: 0	0Н						
7	6	5	4	3	2	1	0
INT_CHO) -	· ·				-	
Bit	Name			Descri	ption		
7	INT_CH0	This bit indicates whether 0: No interrupt is generate 1: At least one interrupt is	ed or all the interrupts	are cleared in cha			
6 - 0	-	Reserved.					

INTTM - One Second Timer Interrupt Status Register

ddress: 3C0H /pe: Read / W efault Value: 0	rite							
7	6		5	4	3	2	1	0
-	•		•	•	·	•	•	TMOV_IS
Bit	Name				Descri	ption		
7 - 1	-	Reserved.						
0	TMOV_IS	0: No one se	econd time ov	ver interrupt is gene	b0, GCF) is '0'. This rated; or a '1' is writte ed and is reported by	en to this bit. (defa		e second time over.

5.2.2 PER-CHANNEL REGISTER

CHCF - Channel Configuration Register

	241H, 281H, 2C1F <i>(CH0)</i> ite	I, 101H, 141H, 181H, 1C1 I, 301H, 341H, 381H, <i>(CH</i>)					
7	6	5	4	3	2	1	0
-	•	•	-	-	•	CHRST	T1E1
Bit	Name			Desc	ription		
7 - 2	-	Reserved.					
1	CHRST	Writing a '1' to this bit wil mum. This bit is self cleared.	l initiate per-channe	l software reset. O	nce initiated, per-cha	nnel software reset o	completes in 1 µs maxi-
0	T1E1	This bit is valid only when 0: T1/J1. (default) 1: E1. This bit can not be reset			s T1/J1 or E1 operatio	on mode.	

TJA - Transmit Jitter Attenuation Configuration Register

Addres	ss: 002H. 0	42H. 082H. 0C2H	I, 102H, 142H, 182H, 1C	2H. (CH1~CH8)				
			I, 302H, 342H, 382H, <i>(Cl</i>					
	7C2H (, , , ,					
Type:	Read / Wri							
	It Value: 00							
						_		
	7	6	5	4	3	2	1	0
	_			TJA_LIMT	TJA_EN	TJA_DP1	TJA_DP0	TJA_BW
	-						IJA_DI U	137_0
	Bit	Name			Descri	ption		
Ī	7 - 5	-	Reserved.					
	4	TJA_LIMT	This bit determines whe	ther the JA-Limit funct	ion is enabled in the	e TJA.		
			0: Disable. (default)					
				the TJA outgoing data	a will be adjusted au	utomatically if the FIF	O in the TJA is 2-bi	t close to its full or emp-
			tiness.					
	3	TJA_EN	This bit controls whethe	r the TJA is enabled to) use.			
			0: Disable. (default)					
			1: Enable.					
4	2 - 1	TJA_DP[1:0]	These bits select the de	pth of the TJA FIFO.				
			00: 128-bit. (default)					
			01: 64-bit.					
			1X: 32-bit.					
	0	TJA_BW	This bit selects the Corr	ner Frequency for the	ГJА.			
			0: 5 Hz (in T1/J1 mode)	/ 6.77 Hz (in E1 mode	e). (default)			
			1: 1.26 Hz (in T1/J1 mo	de) / 0.87 Hz (in E1 m	ode).			

RJA - Receive Jitter Attenuation Configuration Register

7	6	5	4	3	2	1	0
-	•	· ·	RJA_LIMT	RJA_EN	RJA_DP1	RJA_DP0	RJA_BW
Bit	Name			Descri	ption		
7 - 5	-	Reserved.					
4	RJA_LIMT	This bit determines whe 0: Disable. (default) 1: Enable. The speed o emptiness.				FIFO in the RJA is	2-bit close to its f
3	RJA_EN	This bit controls whether 0: Disable. (default) 1: Enable.	r the RJA is enabled to	o use.			
2 - 1	RJA_DP[1:0]	These bits select the de 00: 128-bit. (default) 01: 64-bit. 1X: 32-bit.	oth of the RJA FIFO.				
	RJA BW	This bit selects the Corn	an Engennen vielen tier				

TCF0 - Transmit Configuration Register 0

7	6	5	4	3	2	1	0	
-	OE	T_OFF	THZ_OC	T_SING	T_TERM2	T_TERM1	T_TERM0	
Bit	Name			Descrip	otion			
7	-	Reserved.						
6	OE	This bit determines the o 0: High-Z. (default) 1: Normal operation.	utput of the Line Driv	er, i.e., the output or	n the TTIPn and TRI	NGn pins.		
5	T_OFF		his bit determines whether the transmitter is powered down. : Normal operation. (default) : Power down.					
4	THZ_OC	This bit determines the o 0: The output current is li 1: The output current is li when the TOC is detecte	mited to 100 mAp-p. mited to 100 mAp-p	(default) within the first 1 ms a		·		
3	T_SING	This bit determines the tr 0: Transmit Differential lin 1: Transmit Single Endeo	ne interface. Both TT					
2 - 0	T_TERM[2:0]	These bits select the imp 000: The 100 Ω internal 001: The 110 Ω internal i 010: The 120 Ω internal 011: The 75 Ω internal in 100: The 100 Ω internal 101: The 110 Ω internal 110: The 120 Ω internal	impedance matching mpedance matching impedance matching ipedance matching impedance matching mpedance matching	is selected for T1 10 is selected for J1 11 is selected for E1 12 s selected for E1 75 is selected for T1 10 is selected for J1 11	00 Ω twisted pair ca 10 Ω twisted pair cal 20 Ω twisted pair ca Ω coaxial cable (wit 00 Ω twisted pair ca 10 Ω twisted pair cal	ble (with transforme ole (with transforme ble (with transforme th transformer). ble (transformer-les ole (transformer-less	r). er). s). s).	

TCF1 - Transmit Configuration Register 1

It Value: 0	rite 1H 6	5	4	3	2	1	0		
-	-	T_MD1	T_MD0						
Bit	Name			Descrip	ption				
7 - 5	-	Reserved.							
4	TCK_ES	This bit selects the active 0: Falling edge. (default) 1: Rising edge.							
3	TD_INV	This bit determines the a 0: Active high. (default) 1: Active low.	ctive level on the TD	n, TDPn and TDNn	pins.				
2	T_CODE	This bit selects the line c 0: B8ZS (in T1/J1 mode) 1: AMI.							
1-0	T_MD[1:0]	These bits determines th 00: Transmit Single Rail or 2.048 MHz (in E1 moc 01: Transmit Dual Rail N T1/J1 mode) or 2.048 MI 10: Transmit Dual Rail R 11: Reserved.	NRZ Format system le) clock is input on 1 RZ Format system ir Hz (in E1 mode) cloc	interface. The data is CLKn. nterface. The data is k is input on TCLKn.	s input on TDPn and (default)	TDNn in NRZ forma	·		

PULS - Transmit Pulse Configuration Register

7	6	5	4	4 3	2	1	0
-	-	-		- PUL		PULS1	PULS0
Bit	Name	1	_		Description		
7 - 4	-	Reserved.					
3 - 0	PULS[3:0]	waveform.	elect one of the eight preset waveform templates for short haul applicat				
		PULS[3:0]	Mode	Clock	Cable Impedance	Cable Range	Cable Loss
		0000	E1	2.048 MHz	E1 75 Ω	-	0 ~ 12 dB
		0001	E1	2.048 MHz	E1 120 Ω	-	0 ~ 12 dB
		0010 (default)	DSX1	1.544 MHz	100 Ω	0 ~ 133 ft	0 ~ 0.6 dB
		0011	DSX1	1.544 MHz	100 Ω	133 ~ 266 ft	0.6 ~ 1.2 dB
		0100	DSX1	1.544 MHz	100 Ω	266 ~ 399 ft	1.2 ~ 1.8 dB
		0101	DSX1	1.544 MHz	100 Ω	399 ~ 533 ft	1.8 ~ 2.4 dB
		0110	DSX1	1.544 MHz	100 Ω	533 ~ 655 ft	2.4 ~ 3.0 dB
					110 0		
		0111	J1	1.544 MHz	110 Ω	-	0 ~ 12 dB

SCAL - Amplitude Scaling Control Register

	247H, 287H, 2C7⊢ <i>(CH0)</i> ite	I, 107Н, 147Н, 187Н, 1С7I I, 307Н, 347Н, 387Н, <i>(СН</i> §	,				
7	6	5	4	3	2	1	0
-	-	SCAL5	SCAL4	SCAL3	SCAL2	SCAL1	SCAL0
Bit	Name			Descrip	otion		
7 - 6 5 - 0	- SCAL[5:0]	Reserved. These bits specify a scali In T1/J1 mode, the stand standard value will result In E1 mode, the standard standard value will result Note: The default value should be written to these	ard value is '110110 in 2% scaling up or d value is '100001' f in 3% scaling up or for SCAL[5:0] is '11	' for the waveform a down against the wa for the waveform an down against the wa 0110' which is the T	mplitude. If necessa aveform amplitude. 1 aplitude. If necessar aveform amplitude. 1	ry, increasing or de The scale range is fr y, increasing or deo The scale range is fr	rom +20% to -100%. creasing by '1' from the rom +100% to -100%.

AWG0 - Arbitrary Waveform Generation Control Register 0

A	04011 00011 00011							
	248H, 288H, 2C8H <i>(CH0)</i> rite	I, 108H, 148H, 188H, 1Ci I, 308H, 348H, 388H, <i>(Ch</i>						
7	6	5	4	3	2	1	0	
-	DONE	RW	SAMP4	SAMP3	SAMP2	SAMP1	SAMP0	
Bit	Name			Dosoriu	ntion			
-	indille	Deserved		Descrip	ption			
7	- DONE	Reserved. This bit is valid only when the user-programmable arbitrary waveform is enabled (i.e., the PULS[3:0] bits (b3~0, PULS,) are set						
o	DONE	to '1XXX'). This bit dete 0: Disable. (default) 1: Enable.		•		the Pols(3:0) bits (D3~U, PULS,) are set	
5	 RW This bit is valid only when the user-programmable arbitrary waveform is enabled (i.e., the PULS[3:0] bits (b3~0, PULS to '1XXX'). This bit determines read/write direction. 0: Write data to RAM. (default) 1: Read data from RAM. 						b3~0, PULS,) are set	
4 - 0	SAMP[4:0]	These bits are valid onl are set to '1XXX'). Thes 00000: The RAM sampl 00001: The RAM sampl 00010: The RAM sampl 10001: The RAM sampl 10010: The RAM sampl 10011 ~ 11111: The RAM	e bits specify the RAI e address is 0. (defau e address is 1. e address is 2. e address is 17. e address is 18.	M sample address. ult)	waveform is enabled	d (i.e., the PULS[3:(0] bits (b3~0, PULS,)	

AWG1 - Arbitrary Waveform Generation Control Register 1

,	249H, 289H, 2C9H (<i>CH0)</i> ite	I, 109H, 149H, 189H, 1C9 I, 309H, 349H, 389H, <i>(CH</i>	// /					
7	6	5	4	3	2	1	0	
	WDAT	WDAT5	WDAT4	WDAT3	WDAT2	WDAT1	WDAT0	
Bit	Name			Descri	ption			
7	-	Reserved.						
6 - 0	WDAT[6:0]	These bits are valid only when the user-programmable arbitrary waveform is enabled (i.e., the PULS[3:0] bits (b3~0, PULS,) are set to '1XXX'). These bits contain the template sample data to be stored in RAM which address is specified by the SAMP[4:0] bits (b4~0, AWG0,). They are not updated until new template sample data is written.						

RCF0 - Receive Configuration Register 0

	I <i>(CH0)</i> /rite	H, 30AH, 34AH, 38AH, <i>(Cl</i>	19~CH15)						
7	6	5	4	3	2	1	0		
RCKH	RHZ	R_OFF	R120IN	R_SING	R_TERM2	R_TERM1	R_TERM0		
Bit	Name			Descri	ption				
7	RCKH								
6	RHZ	This bit determines the output of all receive system interfaced pins (including RDPn, RDNn and RCLKn) when the corresponding receiver is powered down. 0: Low level. 1: High-Z. (default)							
5	R_OFF	This bit determines whether the receiver is powered down. 0: Normal operation. (default) 1: Power down.							
4	R120IN	This bit is valid only when the receive line interface is in Receive Differential mode and per-channel internal impedance matchin configuration is enabled. This bit selects the internal impedance matching mode. 0: Partially Internal Impedance Matching mode. An internal programmable resistor (IM) and a value-fixed external resistor (Ri are used. (default) 1: Fully Internal Impedance Matching mode. Only an internal programmable resistor (IM) is used.							
3	R_SING	This bit determines the receive line interface. 0: Receive Differential line interface. Both RTIPn and RRINGn are used to receive signal from the line side. (default)							
2 - 0	R_TERM[2:0]	These bits are valid only matching mode of the red In Receive Differential mode of the red 000: The 100 Ω internal i 001: The 110 Ω internal i 010: The 120 Ω internal in 011: The 75 Ω internal in 1XX: External impedance In Receive Single Endec (default)	eive path to match to ode: mpedance matching mpedance matching mpedance matching pedance matching matching is selecte	the cable impedance g is selected for T1 1 g is selected for J1 1 g is selected for E1 1 is selected for E1 75 ed for T1 100 Ω , J1 1	Ω Ω twisted pair can be calculated to Ω coaxial cable. 110 Ω , E1 120 Ω tw	able. ble. able. isted pair cable and	E1 75 Ω coaxial ca		

RCF1 - Receive Configuration Register 1

20BH, 7CBH Type: Read / W	, 24BH, 28BH, 2C <i>(CH0)</i> /rite	BH, 10BH, 14BH, 18BH, BH, 30BH, 34BH, 38BH,					
Default Value: 0 7	6	5	4	3	2	1	0
·	•	· ·	RCK_ES	RD_INV	R_CODE	R_MD1	R_MD0
Bit	Name			Descri	ption		
7 - 5	-	Reserved.					
4	RCK_ES	This bit selects the ac 0: Rising edge. (defau 1: Falling edge.	tive edge of the RCLKn lt)	ı pin.			
3	RD_INV	This bit determines th 0: Active high. (defaul 1: Active low.	e active level on the RE t)	DPn and RDNn pins.			
2	R_CODE		e code rule for the rece de) / HDB3 (in E1 mode				
1 - 0	R_MD[1:0]	01: Receive Dual Rail T1/J1 mode) or 2.048 10: Receive Dual Rail J1 mode) or 2.048 MH	MHz (in E1 mode) reco RZ Format system inte Iz (in E1 mode) recover	terface. The data is overed clock is outpuerface. The data is outpuerface. The data is output o	ut on RCLKn. (defaul utput on RDPn and R n RCLKn.	t) RDNn in RZ format a	nat and a 1.544 MHz (in and a 1.544 MHz (in T1/ ly after passing through

RCF2 - Receive Configuration Register 2

	24CH, 28CH, 2CC <i>(CH0)</i> ite	:н, 10СН, 14СН, 18СН, 1С :н, 30СН, 34СН, 38СН, <i>(Сł</i>					
7	6	5	4	3	2	1	0
·	•	-	-	·	-	MG1	MG0
Bit	Name			Descript	ion		
7 - 2	-	Reserved.					
1 - 0	MG[1:0]	These bits select the Moni 00: 0 dB. (default) 01: 20 dB. 10: 26 dB. 11: 32 dB.	tor Gain.				

LOS - LOS Configuration Register

7	6		5	4	3	2	1	0
LAC	ALOS	62 A	LOS1	ALOS0	TALOS1	TALOS0	TDLOS1	TDLOS0
Bit	Name				Des	cription		
7	LAC	0: T1.231 (in	T1/J1 mode) /	LOS and AIS c G.775 (in E1 m TSI 300233 & I.				
6 - 4	ALOS[2:0]	vals, LLOS is	declared. The 0] settings for I	consecutive pu Normal Receive	lse intervals (N) ar	itude of the data is less e determined by the LA pnitor mode are differer	C bit (b7, LOS,).	
				OS[2:0]	Q (Vpp)	vs. 6.0 Vpp (dB)	vs. 4.74 Vpp (dB)	
				000	0.5	21.58	19.54	
			001	(default)	0.7	18.66	16.61	
				010	0.9	16.48	14.43	
				011	1.2	13.98	11.93	
				100	1.4	12.64	10.59	
				101	1.6	11.48	9.43	
				110	1.8	10.46	8.41	
				111	2.0	9.54	7.49	
			ALOS[2	2:0] Setting ir	Line Monitor N	lode		
			AL	OS[2:0]	Q (Vpp)	vs. 6.0 Vpp (dB)	vs. 4.74 Vpp (dB)	
				000	1.0	15.56	13.52	
			001	(default)	1.4	12.64	10.59	
				010	1.8	10.46	8.41	
				011	2.2	8.71	6.67	
				1xx		reserved.		
3 - 2	TALOS[1:0]	declared. The TLOS is clear For Differentia 00: 1.2 Vp. 01: 0.9 Vp. (d 10: 0.6 Vp. 11: 0.4 Vp.	e period is dete red. al line interface lefault) nded line interfa	ermined by the		e of the data is less than , LOS,). When the ar		



1 - 0	TDLOS[1:0]	These bits select the period. When the amplitude of the data is less than a certain voltage for the period, TLOS is declared. The
		voltage is determined by the TALOS bits (b3~2, LOS,).
		00: 16-pulse.
		01: 32-pulse. (default)
		1X: 64-pulse.

ERR - Error Detection & Insertion Control Register

7CEH Read / Wi It Value: 0										
7	6	5	4	3	2	1	0			
EXZ_DE	F BPV_IN	BPV_INS ERR_INS CNT_SEL2 CNT_SEL1 CNT_SEL0 CNT_MD								
Bit	Name	Description								
7	EXZ_DEF	This bit selects the EXZ c 0: ANSI. (default) 1: FCC.	lefinition standard.							
6	BPV_INS	This bit controls whether Writing '1' to this bit will ir This bit is cleared once th	isert a BPV on the n	ext available mark i		be transmitted.				
5	ERR_INS	A transition from '0' to '1'	This bit controls whether to insert a single bit error to the generated PRBS/ARB pattern. A transition from '0' to '1' on this bit will insert a single bit error to the generated PRBS/ARB pattern. This bit is cleared once the single bit error insertion is completed.							
4 - 2	CNT_SEL[2:0]	These bits select what kir 000: Disable. (default) 001: LBPV. 010: LEXZ. 011: LBPV + LEXZ. 100: SBPV. 101: SEXZ. 110: SBPV + SEXZ. 111: PRBS/ARB error.	nd of error to be cour	nted by the internal	Error Counter.					
1	CNT_MD	This bit determines wheth 0: Manually by setting the 1: Every-one second auto	CNT_STOP bit (b0		updated automaticall	y or manually.				
0	CNT_STOP	This bit is valid only wher A transition from '0' to '1' This bit must be cleared t	on this bit updates the	he ERRCH & ERRC	CL registers.					

AISG - AIS Generation Control Register

	24FH, 28FH, 2CFF <i>(CH0)</i> ite	H, 10FH, 14FH, 18FH, 1Cl H, 30FH, 34FH, 38FH, <i>(CF</i> 5		3	2	1	0
·		· ·	TXAIS	ASAIS_SLOS	ASAIS_LLOS	ALAIS_SLOS	ALAIS_LLOS
Bit	Name			Descri	ption		
7 - 5	-	Reserved.					
4	TXAIS	This bit controls the trans 0: Disable. (default) 1: Transmit all ones patte In remote loopback, this	ern at TTIPn/TRINGn	·			
3	ASAIS_SLOS	This bit controls the AIS 0: Disable. (default) 1: Enable.	-				
2	ASAIS_LLOS	This bit controls the AIS 0: Disable. (default) 1: Enable.	-				
1	ALAIS_SLOS	This bit controls the AIS (0: Disable. (default) 1: Enable.	generation in the trar	smit path once SLC)S is detected.		
0	ALAIS_LLOS	This bit controls the AIS (0: Disable. (default) 1: Enable.	generation in the trar	smit path once LLC)S is detected.		

PG - Pattern Generation Control Register

7	6	5	4	3	2	1	0
-	PG_CK	PG_EN1	PG_EN0	PG_POS	PAG_INV	PRBG_SEL1	PRBG_SEL0
Bit	Name			Descrip	otion		
7	-	Reserved.					
	PG_CK	This bit selects the referer When the pattern is gener 0: XCLK. (default) 1: Recovered clock from the When the pattern is gener 0: XCLK. (default) 1: Transmit clock, i.e., the mat mode) or the clock reference to the the the mat mode) or the clock reference to the	ated in the receive ne received signal. ated in the transmit clock input on TCL covered from the da	path: t path: .Kn (in Transmit Sing ata input on TDPn an	le Rail NRZ Forma	t mode and in Trans	
5 - 4	PG_EN[1:0]	These bits select the patte 00: Disable. (default) 01: PRBS. 10: ARB. 11: IB.	rn to be generated				
3	PG_POS	This bit selects the patterr 0: Transmit path. (default) 1: Receive path.	(including PRBS, /	ARB & IB) generation	n direction.		
2	PAG_INV	This bit controls whether to 0: Normal. (default) 1: Invert.	o invert the generat	ted PRBS/ARB patte	m.		
1 - 0	PRBG_SEL[1:0]	These bits are valid only v 00: 2 ²⁰ - 1 QRSS. (default 01: 2 ¹⁵ - 1 PRBS. 1X: 2 ¹¹ - 1 PRBS.		tern is generated. Th	ey select the PRBS	pattern.	

PD - Pattern Detection Control Register

	251H, 291H, 2D1H <i>CH0)</i> te	, 111H, 151H, 191H, 1D1H , 311H, 351H, 391H, <i>(CH9</i> -					
7	6	5	4	3	2	1	0
-	·	•	-	PD_POS	PAD_INV	PAD_SEL1	PAD_SEL0
Bit	Name			Descrip	otion		
7 - 4	-	Reserved.					
3	PD_POS	This bit selects the pattern 0: Receive path. (default) 1: Transmit path.	(including PRBS, A	RB & IB) detection	direction.		
2	PAD_INV	This bit controls whether to 0: Normal. (default) 1: Invert.	o invert the data bef	ore PRBS/ARB dete	ection.		
1 - 0		These bits select the desir 00: 2 ²⁰ - 1 QRSS. 01: 2 ¹⁵ - 1 PRBS. 10: 2 ¹¹ - 1 PRBS. 11: ARB. (default)	ed PRBS/ARB patte	ern to be detected.			

ARBL - Arbitrary Pattern Generation / Detection Low-Byte Register

	252H, 292H, 2D2H <i>(CH0)</i> ite	, 112H, 152H, 192H, 1D2I , 312H, 352H, 392H, <i>(CH</i>					
7	6	5	4	3	2	1	0
ARB7	ARB6	ARB5	ARB4	ARB3	ARB2	ARB1	ARB0
Bit	Name			Descrip	tion		
7 - 0	ARB[7:0]	These bits, together with be generated or detected	·		•	r detected. The AR	B23 bit is the first bit to

ARBM - Arbitrary Pattern Generation / Detection Middle-Byte Register

	253H, 293H, 2D3H C <i>H0)</i> te	I, 113H, 153H, 193H, 1D3 , 313H, 353H, 393H, <i>(CH</i>					
7	6	5	4	3	2	1	0
ARB15	ARB14	ARB13	ARB12	ARB11	ARB10	ARB9	ARB8
Bit 7 - 0	Name ARB[15:8]	(Refer to the description	of the ARBL register.	Descr	iption		

ARBH - Arbitrary Pattern Generation / Detection High-Byte Register

214	H, 254 H <i>(CH</i> Write	H, 294H, 2D4H	I, 114H, 154H, 194H, 1 I, 314H, 354H, 394H, (ι					
7		6	5	4	3	2	1	0
ARE	23	ARB22	2 ARB21	ARB20	ARB19	ARB18	ARB17	ARB16
Bit		Name			Descri	ption		
7 - 0		ARB[23:16]	(Refer to the description	on of the ARBL register.	.)			

IBL - Inband Loopback Control Register

7	6	5	4 IBGL0	3	2 IBAL0	1	0 IBDL0
-	-	IBGL1		IBAL1		IBDL1	
Bit	Name			Descrip	tion		
7 - 6	-	Reserved.					
5 - 4	IBGL[1:0]	These bits define the ler 00: 5-bit long in the IBG 01: 6-bit long in the IBG 10: 7-bit long in the IBG 11: 8-bit long in the IBG	[4:0] bits (b4~0, IBG, [5:0] bits (b5~0, IBG, [6:0] bits (b6~0, IBG,	.). (default) .). .).	mmed in the IBG[7:0] bits (b7~0, IBĜ,).	
3 - 2	IBAL[1:0]	These bits define the ler 00: 5-bit long in the IBA 01: 6-bit long in the IBA 10: 7-bit long in the IBA 11: 8-bit long in the IBA	4:0] bits (b4~0, IBDA, 5:0] bits (b5~0, IBDA, 6:0] bits (b6~0, IBDA,). (default)).).	o code programmed	in the IBA[7:0] bits (b	.7~0, IBDA,).

IBG - Inband Loopback Generation Code Definition Register

	256H, 296H, 2D6H C <i>H0)</i> te	I, 116H, 156H, 196H, 1D6I I, 316H, 356H, 396H, <i>(CH</i> s	,				
7	6	5	4	3	2	1	0
IBG7	IBG6	IBG5	IBG4	IBG3	IBG2	IBG1	IBG0
Bit	Name			Descript	tion		
7 - 0	IBG[7:0]	The IBG[X:0] bits define IBG0 bit is the last bit to b					

IBDA - Inband Loopback Detection Target Activate Code Definition Register

	257H, 297H, 2D7H (<i>CH0)</i> ite	, 117H, 157H, 197H, 1D7 , 317H, 357H, 397H, <i>(CH</i>					
7	6	5	4	3	2	1	0
IBA7	IBA6	IBA5	IBA4	IBA3	IBA2	IBA1	IBA0
Bit	Name			Descrip	otion		
7 - 0	IBA[7:0]	The IBA[X:0] bits define IBL,). The IBA0 bit is the IBA0 bi		•	ection code. The 'X'	is determined by th	ne IBAL[1:0] bits (b3~2,

IBDD - Inband Loopback Detection Target Deactivate Code Definition Register

	258H, 298H, 2D8H C <i>H0)</i> te	i, 118H, 158H, 198H, 1D8i , 318H, 358H, 398H, <i>(CH</i> :					
7	6	5	4	3	2	1	0
IBD7	IBD6	IBD5	IBD4	IBD3	IBD2	IBD1	IBD0
Bit	Name			Descrip	tion		
7 - 0	IBD[7:0]	The IBD[X:0] bits define to IBL,). The IBD0 bit is the IBD0		•	ection code. The 'X	is determined by th	ne IBDL[1:0] bits (b1~0,

LOOP - Loopback Control Register

	OH						
7	6	5	4	4 3	2	1	0
-	-	•	-	AUTOLP	DLP	RLP	ALP
Bit	Name			Descript	tion		
7 - 4	- F	Reserved.					
3		Fhis bit determines wheth D: Automatic Digital/Remo 1: Automatic Digital/Remo vate IB code is detected in he deactivate IB code is d	te Loopback is dis te Loopback is ena n the transmit/rece	abled. (default) abled. The correspond ive path for more than	ing channel will ente 5.1 sec.; and will r		
2	(This bit controls whether [): Disable. (default) 1: Enable.	Digital Loopback is	enabled.			
1	(This bit controls whether F): Disable. (default) 1: Enable.	Remote Loopback i	is enabled.			
0	(This bit controls whether A): Disable. (default) 1: Enable.	Analog Loopback is	enabled.			

INTES - Interrupt Trigger Edges Select Register

7	6	5	4	3	2	1	0		
	AIS_IE	S PA_IES	LOS_IES	IB_IES					
Bit	Name			Descrip	iption				
7	-	Reserved.							
6	AIS_IES	This bit selects the transi 0: A transition from '0' t INTS1,) / the SAIS_IS I 1: Any transition from '0 LAIS_IS bit (b6, INTS1,	o '1' on the LAIS_S bit (b7, INTS1,) to ' to '1' or from '1' to	bit (b6, STAT1,) '1' respectively. (defa '0' on the LAIS_S k	/ the SAIS_S bit (b nult) bit (b6, STAT1,) /	o7, STAT1,) will se			
5	PA_IES	_AIS_IS bit (b6, INTS1,) / the SAIS_IS bit (b7, INTS1,) to '1' respectively. This bit selects the transition edge of the PA_S bit (b5, STAT1,). D: A transition from '0' to '1' on the PA_S bit (b5, STAT1,) will set the PA_IS bit (b5, INTS1,) to '1'. (default) I: Any transition from '0' to '1' or from '1' to '0' on the PA_S bit (b5, STAT1,) will set the PA_IS bit (b5, INTS1,) to '1'.							
4	TOC_IES	0: A transition from '0' to	1: Any transition from '0' to '1' or from '1' to '0' on the PA_S bit (b5, STAT1,) will set the PA_IS bit (b5, INTS1,) to ' This bit selects the transition edge of the TOC_S bit (b4, STAT0,). 0: A transition from '0' to '1' on the TOC_S bit (b4, STAT0,) will set the TOC_IS bit (b4, INTS0,) to '1'. (default) 1: Any transition from '0' to '1' or from '1' to '0' on the TOC_S bit (b4, STAT0,) will set the TOC_IS bit (b4, INTS0,)						
3	TCKLOS_IES	This bit selects the transi 0: A transition from '0' to 1: Any transition from '0' '1'.	'1' on the TCKLOS_	S bit (b3, STAT0,)	will set the TCKLOS				
2	TLOS_IES	This bit selects the transi 0: A transition from '0' to 1: Any transition from '0'	'1' on the TLOS_S b	oit (b2, STAT0,) will	set the TLOS_IS bi				
1	LOS_IES	This bit selects the transi 0: A transition from '0' to INTS0,) / the SLOS_IS 1: Any transition from '0' LLOS_IS bit (b0, INTS0,.	o '1' on the LLOS_S bit (b1, INTS0,) to to '1' or from '1' to	bit (b0, STAT0,) / 1' respectively. (def '0' on the LLOS_S b	, the SLOS_S bit (b ault) bit (b0, STAT0,) /	o1, STAT0,) will se			
0	IB_IES	This bit selects the transi 0: A transition from '0' to the IBD_IS bit (b0, INTS' 1: Any transition from '0' bit (b1, INTS1,) / the IB	'1' on the IBA_S bit I,) to '1' respective to '1' or from '1' to '0	(b1, STAT1,) / the I ly. (default))' on the IBA_S bit (b	BD_S bit (b0, STAT o1, STAT1,) / the I	1,) will set the IBA	v		

INTM0 - Interrupt Mask Register 0

Ilt Value: F	6	5	4	3	2	1	0
DAC_IM	TJA_IM	1 RJA_IM	TOC_IM	TCKLOS_IM	TLOS_IM	SLOS_IM	LLOS_IM
Bit	Name			Descrip	tion		
7	DAC_IM	This bit is the waveform a 0: Interrupt is enabled. 1: Interrupt is masked. (d		•			
6	TJA_IM	This bit is the TJA FIFO c 0: Interrupt is enabled. 1: Interrupt is masked. (d		ow interrupt mask.			
5	RJA_IM	This bit is the RJA FIFO of 0: Interrupt is enabled. 1: Interrupt is masked. (d		ow interrupt mask.			
4	TOC_IM	This bit is the Line Driver 0: Interrupt is enabled. 1: Interrupt is masked. (d		ζ.			
3	TCKLOS_IM	This bit is the TCLKn mis 0: Interrupt is enabled. 1: Interrupt is masked. (d	•				
2	TLOS_IM	This bit is the TLOS intern 0: Interrupt is enabled. 1: Interrupt is masked. (d					
1	SLOS_IM	This bit is the SLOS inter 0: Interrupt is enabled. 1: Interrupt is masked. (d					
0	LLOS_IM	This bit is the LLOS interr 0: Interrupt is enabled.					

INTM1 - Interrupt Mask Register 1

7	6	5	4	3	2	1	0
SAIS_IM	LAIS_IM	I PA_IM	•	•	IBA_IM	IBD_IM	
Bit	Name			Descrip	otion		
7		This bit is the SAIS interru 0: Interrupt is enabled. 1: Interrupt is masked. (de					
6		This bit is the LAIS interru 0: Interrupt is enabled. 1: Interrupt is masked. (de	-				
5		This bit is the PRBS/ARB 0: Interrupt is enabled. 1: Interrupt is masked. (de		tion interrupt mask.			
4 - 2	-	Reserved.					

INTM2 - Interrupt Mask Register 2

e: Read / Wi ault Value: 3							
7	6	5	4	3	2	1	0
-	•	SBPV_IM	LBPV_IM	SEXZ_IM	LEXZ_IM	ERR_IM	CNTOV_IM
Bit	Name			Descri	otion		
7 - 6	-	Reserved.					
5	SBPV_IM	This bit is the SBPV inter 0: Interrupt is enabled. 1: Interrupt is masked. (c					
4	LBPV_IM	This bit is the LBPV inter 0: Interrupt is enabled. 1: Interrupt is masked. (c					
3	SEXZ_IM	This bit is the SEXZ inter 0: Interrupt is enabled. 1: Interrupt is masked. (c					
2	LEXZ_IM	This bit is the LEXZ inter 0: Interrupt is enabled. 1: Interrupt is masked. (c					
1	ERR_IM	This bit is the PRBS/ARE 0: Interrupt is enabled. 1: Interrupt is masked. (c		Κ.			
0	CNTOV_IM	This bit is the ERRCH ar 0: Interrupt is enabled. 1: Interrupt is masked. (c	d ERRCL registers of	overflow interrupt ma	ask.		

STAT0 - Status Register 0

	25EH, 29EH, 2DEI <i>CH0)</i>	Н, 11ЕН, 15ЕН, 19ЕН, 11 Н, 31ЕН, 35ЕН, 39ЕН, (0					
7	6	5	4	3	2	1	0
AUTOLP_S	- ·	•	TOC_S	TCKLOS_S	TLOS_S	SLOS_S	LLOS_S
Bit	Name			Descrip	tion		
7	AUTOLP_S	This bit indicates the au 0: Out of automatic Digit 1: In automatic Digital/R	al/Remote Loopback	e Loopback status.			
6 - 5	-	Reserved.					
4	TOC_S	This bit indicates the TC 0: No TOC is detected. 1: TOC is detected.					
3	TCKLOS_S	This bit indicates the TC 0: TCLKn is not missing 1: TCLKn is missing.					
2	TLOS_S	This bit indicates the TL 0: No TLOS is detected. 1: TLOS is detected.					
1	SLOS_S	This bit indicates the SL 0: No SLOS is detected. 1: SLOS is detected.					
0	LLOS_S	This bit indicates the LL 0: No LLOS is detected. 1: LLOS is detected.					

STAT1 - Status Register 1

	25FH, 29FH, 2DFH CH0)	I, 11FH, 15FH, 19FH, 1DFI I, 31FH, 35FH, 39FH, <i>(CH</i> S	., ,				
7	6	5	4	3	2	1	0
SAIS_S	LAIS_S	PA_S		-		IBA_S	IBD_S
Bit	Name			Descrip	tion		
7	SAIS_S	This bit indicates the SAIS 0: No SAIS is detected. (d 1: SAIS is detected.					
6	LAIS_S	This bit indicates the LAIS 0: No LAIS is detected. (d 1: LAIS is detected.					
5	PA_S	This bit indicates the PRB 0: The PRBS/ARB pattern 1: The PRBS/ARB pattern	is out of synchroniz	zation. (default)			
4 - 2	-	Reserved.					
1	IBA_S	This bit indicates the activ 0: No activate IB code is of 1: Activate IB code is dete more than 5.1 sec. when t	etected. (default) cted for more than		TOLP bit (b3, LOOF	2,) is '0' or activate I	B code is detected fo
0	IBD_S	This bit indicates the deac 0: No deactivate IB code is 1: Deactivate IB code is de is '0' or deactivate IB code	s detected. (default) etected for more that	n 40 ms (in T1/J1 mc			OLP bit (b3, LOOP,

INTS0 - Interrupt Status Register 0

7	6	5	1	0					
DAC_IS	TJA_I	S RJA_IS	RJA_IS TOC_IS TCKLOS_IS TLOS_IS S						
Bit	Name			Descrip	tion				
7	DAC_IS	This bit indicates the inter 0: No waveform amplitude 1: Waveform amplitude ov	e overflow interrupt	is generated; or a '1'	is written to this bit.				
6	TJA_IS	This bit indicates the inter 0: No TJA FIFO overflow	This bit indicates the interrupt status of the TJA FIFO overflow or underflow. 0: No TJA FIFO overflow or underflow interrupt is generated; or a '1' is written to this bit. (default) 1: TJA FIFO overflow or underflow interrupt is generated and is reported by the INT pin. This bit indicates the interrupt status of the RJA FIFO overflow or underflow. 0: No RJA FIFO overflow or underflow interrupt is generated; or a '1' is written to this bit. (default) 1: RJA FIFO overflow or underflow interrupt is generated and is reported by the INT pin. This bit indicates the interrupt status of the RJA FIFO overflow or underflow. 1: RJA FIFO overflow or underflow interrupt is generated and is reported by the INT pin.						
5	RJA_IS	0: No RJA FIFO overflow							
4	TOC_IS	This bit indicates the interrupt status of the Line Driver TOC. 0: No TOC interrupt is generated; or a '1' is written to this bit. (default) 1: TOC interrupt is generated and is reported by the INT pin. When the TOC_IES bit (b4, INTES,) is '0', a transition fro '1' on the TOC_S bit (b4, STAT0,) set this bit to '1'; when the TOC_IES bit (b4, INTES,) is '1', any transition (from '0' from '1' to '0') on the TOC_S bit (b4, STAT0,) set this bit to '1'.							
3	TCKLOS_IS	0: No TCLKn missing inter 1: TCLKn missing interrup tion from '0' to '1' on the T	This bit indicates the interrupt status of the TCLKn missing. 0: No TCLKn missing interrupt is generated; or a '1' is written to this bit. (default) 1: TCLKn missing interrupt is generated and is reported by the INT pin. When the TCKLOS_IES bit (b3, INTES,) is '0', tion from '0' to '1' on the TCKLOS_S bit (b3, STAT0,) set this bit to '1'; when the TCKLOS_IES bit (b3, INTES,) is '1', a sition (from '0' to '1' or from '1' to '0') on the TCKLOS_S bit (b3, STAT0,) set this bit to '1'.						
2	TLOS_IS	This bit indicates the inter 0: No TLOS interrupt is gene 1: TLOS interrupt is gene to '1' on the TLOS_S bit ('1' or from '1' to '0') on the	rupt status of TLOS enerated; or a '1' is rated and is reporte b2, STAT0,) set t TLOS_S bit (b2, S	S. written to this bit. (de ed by the INT pin. Wh his bit to '1'; when the TATO,) set this bit to	fault) ien the TLOS_IES to e TLOS_IES bit (b2	bit (b2, INTES,) is			
1	SLOS_IS		enerated; or a '1' is rated and is reporte , STAT0,) set this	written to this bit. (de d by the INT pin. Whe bit to '1'; when the LC	en the LOS_IES bit				
0	LLOS_IS	'1' on the SLOS_S bit (b1, STAT0,) set this bit to '1'; when the LOS_IES bit (b1, INTES,) is '1', any transitio from '1' to '0') on the SLOS_S bit (b1, STAT0,) set this bit to '1'.							

INTS1 - Interrupt Status Register 1

7	6	5 4 3 2			1	0	
SAIS_IS	LAIS_	IS PA_IS	-	IBA_IS	IBD_IS		
Bit	Name			Descrip	otion		
7	SAIS_IS	This bit indicates the inte 0: No SAIS interrupt is g 1: SAIS interrupt is gene on the SAIS_S bit (b7, S '1' to '0') on the SAIS_S	enerated; or a '1' is w rated and is reported TAT1,) set this bit to	vritten to this bit. (def by the INT pin. When o '1'; when the AIS_II	n the AIS_IES bit (
6	LAIS_IS	This bit indicates the inte 0: No LAIS interrupt is generated 1: LAIS interrupt is generated on the LAIS_S bit (b6, S '1' to '0') on the LAIS_S	enerated; or a '1' is w rated and is reported TAT1,) set this bit to	ritten to this bit. (defa by the INT pin. When o '1'; when the AIS_II	n the AIS_IES bit (
5	PA_IS	This bit indicates the inte 0: No PRBS/ARB pattern 1: PRBS/ARB pattern sy is '0', a transition from '0 transition (from '0' to '1' o	n synchronization inten nchronization interru)' to '1' on the PA_S	errupt is generated; o pt is generated and i bit (b5, STAT1,) se	or a '1' is written to s reported by the Ī et this bit to '1'; wh	NT pin. When the PA_ ien the PA_IES bit (b	
4 - 2	-	Reserved.					
1	IBA_IS	This bit indicates the inte 0: No activate IB code in 1: Activate IB code inter from '0' to '1' on the IBA '1' or from '1' to '0') on th	terrupt is generated; rupt is generated and _S bit (b1, STAT1,)	or a '1' is written to the d is reported by the set this bit to '1'; whe	INT pin. When the en the IB_IES bit (b		
0	IBD_IS	This bit indicates the inte 0: No deactivate IB code 1: Deactivate IB code int from '0' to '1' on the IBD	interrupt is generate errupt is generated a	d; or a '1' is written to nd is reported by the	INT pin. When the		

INTS2 - Interrupt Status Register 2

: Read / W ult Value: 0										
7	6	5	4 LBPV_IS	3 SEXZ_IS	2 LEXZ_IS	1	0 CNTOV_IS			
-	•	SBPV_IS				ERR_IS				
Bit	Name			Descrip	ption					
7 - 6	-	Reserved.								
5	SBPV_IS	0: No SBPV interrupt is g	This bit indicates the interrupt status of the SBPV.): No SBPV interrupt is generated; or a '1' is written to this bit. (default) I: SBPV interrupt is generated and is reported by the INT pin.							
4	LBPV_IS	This bit indicates the inter 0: No LBPV interrupt is go 1: LBPV interrupt is gene	enerated; or a '1' is v	written to this bit. (de	efault)					
3	SEXZ_IS	This bit indicates the inter 0: No SEXZ interrupt is g 1: SEXZ interrupt is gene	enerated; or a '1' is v	written to this bit. (de	efault)					
2	LEXZ_IS	This bit indicates the inter 0: No LEXZ interrupt is generated as the inter- 1: LEXZ interrupt is generated as the interrupt is the interrupt is generated a	enerated; or a '1' is v	vritten to this bit. (de	efault)					
1	ERR_IS	This bit indicates the inter 0: No PRBS/ARB error in 1: PRBS/ARB error interr	terrupt is generated;	or a '1' is written to						
0	CNTOV_IS	This bit indicates the inter 0: No ERRCH or ERRCL 1: ERRCH and ERRCL re	register overflow int	errupt is generated;	or a '1' is written to th					

ERRCL - Error Counter Low-Byte Register

	263H, 2A3H, 2E3H	, 123H, 163H, 1A3H, 1E3 , 323H, 363H, 3A3H, <i>(CH</i>						
Type: Read Default Value: 00)H							
7	6	5	4	3	2	1	0	
ERRC7	ERRC6	ERRC5	ERRC5 ERRC4 ERRC3 ERRC2 ERRC1					
	Γ							
Bit	Name			Descrip	otion			
7 - 0		These bits, together with updated automatically or error counting; otherwise,	manually, as determ	nined by the CNT_N			•	

ERRCH - Error Counter High-Byte Register

	264H, 2A4H, 2E4H CH0)	I, 124H, 164H, 1A4H, 1E4 I, 324H, 364H, 3A4H, <i>(CH</i>										
7	6	5	5 4 3 2 1 0									
ERRC15	ERRC1	4 ERRC13	ERRC12	ERRC11	ERRC10	ERRC9	ERRC8					
Bit	Name		Description									
7 - 0	ERRC[15:8]	(Refer to the description	of the ERRCL registe	er.)								

JM - Jitter Measurement Configuration For Channel 0 Register

7	6	5	4	3	2	1	0				
-	·	·	·	-	JM_STOP	JM_MD	JM_BW				
Bit	Name		Description								
7 - 3	-	Reserved.	Reserved.								
2	JM_STOP	This bit is valid only whe A transition from '0' to '1 This bit must be cleared	on this bit updates t	he JIT_PH, JIT_PL	. and JIT_NH, JIT_NL	registers.					
1	JM_MD	0: The period is determine	This bit selects the jitter measurement period. 0: The period is determined manually by setting the JM_STOP bit (b2, JM). (default) 1: The period is one second automatically.								
0	JM_BW	This bit selects the band 0: 10 Hz ~ 40 KHz (in T1 1: 8 KHz ~ 40 KHz (in T ²	/J1 mode) / 20 Hz ~	100 KHz (in E1 mo							

JIT_PL - Positive Peak Jitter Measurement Low-Byte Register

Address: 7E6H Type: Read Default Value: 00)H									
7	6	5	4	3	2	1	0			
JIT_P7	JIT_P6) JIT_P5	JIT_P4	JIT_P3	JIT_P2	JIT_P1	JIT_P0			
Bit	Name			Descrip	otion					
7 - 0	JIT_P[7:0]	measured by channel 0. T read in the next round of j The relationship between	Description These bits, together with the JIT_P[11:8] bits, reflect the greatest positive peak value of the demodulated jitter signal which neasured by channel 0. They are updated automatically or manually, as determined by the JM_MD bit (b1, JM). They should ead in the next round of jitter measurement; otherwise, they will be overwritten. The relationship between the greatest positive peak value and the indication in these bits is: Positive Peak = [JIT_PH, JIT_PL] / 16 (UIpp)							

JIT_PH - Positive Peak Jitter Measurement High-Byte Register

Address: 7E7H Type: Read Default Value: 0							
7	6	5	4	3	2	1	0
·	·		·	JIT_P11	JIT_P10	JIT_P9	JIT_P8
Bit	Name			Descr	ription		
7 - 4	-	Reserved.					
3 - 0	JIT_P[11:8]	(Refer to the description	on of the JIT_PL re	egister.)			

JIT_NL - Negative Peak Jitter Measurement Low-Byte Register

Address: 7E8H Type: Read Default Value: 0	0Н									
7	6	5	4	3	2	1	0			
JIT_N7	JIT_N6	JIT_N5	JIT_N4	JIT_N3	JIT_N2	JIT_N1	JIT_N0			
Bit	Name			Descri	ption					
7 - 0	JIT_N[7:0]	measured by channel 0. read in the next round of The relationship between	Description nese bits, together with the JIT_N[11:8] bits, reflect the greatest negative peak value of the demodulated jitter signal which easured by channel 0. They are updated automatically or manually, as determined by the JM_MD bit (b1, JM). They should b ad in the next round of jitter measurement; otherwise, they will be overwritten. ne relationship between the greatest negative peak value and the indication in these bits is: egative Peak = [JIT_NH, JIT_NL] / 16 (UIpp)							

JIT_NH - Negative Peak Jitter Measurement High-Byte Register

Address: 7E9 Type: Read Default Value									
7		6		5	4	3	2	1	0
-		-		•	-	JIT_N11	JIT_N10	JIT_N9	JIT_N8
Bit	N	ame				Descri	iption		
7 - 4		-	Reserved.						
3 - 0	JIT_	N[11:8]	(Refer to the d	escription of	the JIT_NL reg	jister.)			

16-CHANNEL SHORT HAUL T1/E1/J1 LINE INTERFACE UNIT



6 JTAG

The IDT82P20416 supports the digital Boundary Scan Specification as described in the IEEE 1149.1 standards.

The boundary scan architecture consists of data and instruction registers plus a Test Access Port (TAP) controller. The control of the TAP is achieved through signals applied to the Test Mode Select (TMS) and Test Clock (TCK) input pins. Data is shifted into the registers via the Test

Data Input (TDI) pin, and shifted out of the registers via the Test Data Output (TDO) pin. Both TDI and TDO are clocked at a rate determined by TCK.

The JTAG boundary scan registers include BSR (Boundary Scan Register), DIR (Device Identification Register), BR (Bypass Register) and IR (Instruction Register). These will be described in the following pages. Refer to Figure-36 for architecture.

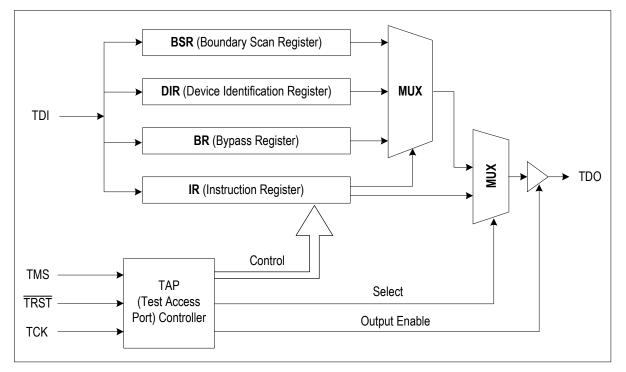


Figure-36 JTAG Architecture

6.1 JTAG INSTRUCTION REGISTER (IR)

The IR with instruction decode block is used to select the test to be executed or the data register to be accessed or both.

The instructions include: EXTEST, SAMPLE/PRELOAD, IDCODE, BYPASS, CLAMP and HIGHZ.

6.2 JTAG DATA REGISTER

6.2.1 DEVICE IDENTIFICATION REGISTER (IDR)

The IDR can be set to define the Version, the Part Number, the Manufacturer Identity and a fixed bit.

6.2.2 BYPASS REGISTER (BYP)

The BYP consists of a single bit. It can provide a serial path between the TDI input and the TDO output. Bypassing the BYR will reduce test access times.

6.2.3 BOUNDARY SCAN REGISTER (BSR)

The bidirectional ports interface to 2 boundary scan cells:

- In cell: The input cell is observable only.
- Out cell: The output cell is controllable and observable.

6.3 TEST ACCESS PORT (TAP) CONTROLLER

The TAP controller is a 16-state synchronous state machine. The states include: Test Logic Reset, Run-Test/Idle, Select-DR-Scan, Capture-DR, Shift-DR, Exit1-DR, Pause-DR, Exit2-DR, Update-DR, Select-IR-Scan, Capture-IR, Shift-IR, Exit1-IR, Pause-IR, Exit2-IR and Update-IR.

Figure-37 shows the state diagram. Note that the figure contains two main branches to access either the data or instruction registers. The value shown next to each state transition in this figure states the value present at TMS at each rising edge of TCK.

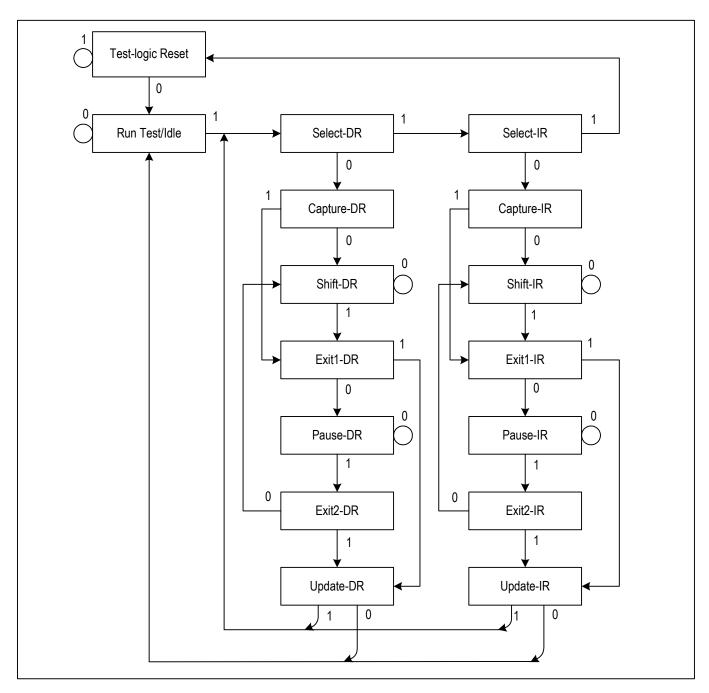


Figure-37 JTAG State Diagram

7 THERMAL MANAGEMENT

The device is designed to operate over the industry temperature range -40°C ~ +85°C. To ensure the functionality and reliability of the device, the maximum junction temperature, T_{jmax} , should not exceed 125°C. In some applications, the device will consume more power and a thermal solution should be provided to ensure the junction temperature T_j does not exceed T_{jmax} . Below is a table listing thermal data for the IDT82P20416.

Package	θ _{JC} (°C/W) ¹	θ_{JB} (°C/W) ²	θ _{JA} (°C/W) ³	Airflow (m/s)
			23.7	0
			18.7	1
484-pin BF	4.2	5.3	17.0	2
			16.1	3
			15.5	4
			15.0	5

Note:

1. Junction-to-Case Thermal Resistance

2. Junction-to-Board Thermal Resistance

3. Junction-to-Ambient Thermal Resistance

7.1 JUNCTION TEMPERATURE

Junction temperature T_j is the temperature of package typically at the geographical center of the chip where the device's electrical circuits are. It can be calculated as follows:

Equation 1: $T_i = T_A + P * \theta_{JA}$

Where:

 θ_{JA} = Junction-to-Ambient Thermal Resistance of the package

T_i = Junction Temperature

T_A = Ambient Temperature

P = Device Power Consumption

For the IDT82P20416, the above values are:

 $\theta_{\rm JA}$ = 23.7 °C/W (when airflow rate is 0 m/s. See the above table)

T_{jmax} = 125 °C

$$T_A = -40 \,^{\circ}\text{C} \sim 85 \,^{\circ}\text{C}$$

P = Refer to Section 8.3 Device Power Consumption and Dissipation (Typical) 1

7.2 EXAMPLE OF JUNCTION TEMPERATURE CAL-CULATION

Assume:

 θ_{JA} = 23.7 °C/W (airflow: 0 m/s)

 $P = 1.46 W (E1 120 \Omega, 100\% \text{ ones}, External Impedance matching})$

The junction temperature T_i can be calculated as follows:

 $T_i = T_A + P * \theta_{JA} = 85 \text{ °C} + 1.46 W X 23.7 \text{ °C/W} = 119.6 \text{ °C}$

The junction temperature of **119.6** °C is below the maximum junction temperature of 125 °C, so no extra heat enhancement is required.

In some operation environments, the calculated junction temperature might exceed the maximum junction temperature of 125 °C and an external thermal solution such as a heatsink is required.

7.3 HEATSINK EVALUATION

A heatsink is expanding the surface area of the device to which it is attached. θ_{JA} is now a combination of device case and heatsink thermal resistance, as the heat flowing from the die junction to ambient goes through the package and the heatsink. θ_{JA} can be calculated as follows:

Equation 2:
$$\theta_{JA} = \theta_{JC} + \theta_{HA}$$

Where:

 θ_{JC} = Junction-to-Case (heatsink) Thermal Resistance

 θ_{HA} = Heatsink-to-Ambient Thermal Resistance

For the IDT82P20416, 0, IC is 4.2 °C/W.

 θ_{HA} determines which heatsink can be selected to ensure the junction temperature does not exceed T_{jmax} . According to Equation 1 and 2, the heatsink-to-ambient thermal resistance θ_{HA} can be calculated as follows:

Equation 3: $\theta_{HA} = (T_j - T_A) / P - \theta_{JC}$

Assume:

$$T_j = 125 \text{ °C} (T_{jmax})$$

 $T_A = 85 \text{ °C}$
 $P = 2.72 W (E1 75 \Omega, 100\% \text{ ones}, Fully Internal Impedance matching})$
 $\theta_{JC} = 4.2 \text{ °C/W}$

The Heatsink-to-Ambient thermal resistance θ_{HA} can be calculated as follows:

θ_{HA} = (125 °C - 85 °C) / 2.72 W - 4.2 °C/W = 10.51 °C/W

That is, if a heatsink whose heatsink-to-ambient thermal resistance θ_{HA} is below or equal to 10.51 °C/W is used in such operation environment, the junction temperature will not exceed the maximum junction temperature.

8 PHYSICAL AND ELECTRICAL SPECIFICATIONS

8.1 ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Min	Мах	Unit
VDDD	Digital Core Power Supply	-0.5	2.2	V
VDDA	Analog Core Power Supply	-0.5	4.6	V
VDDIO	I/O Power Supply	-0.5	4.6	V
VDDT0~15	Power Supply for Transmitter Driver	-0.5	4.6	V
VDDR0~15	Power Supply for Receiver	-0.5	4.6	V
	Input Voltage, Any Digital Pin	GND - 0.5	6	V
V _{in}	Input Voltage, Any RTIP and RRING pin ¹	GND - 0.5	VDDR + 0.5	V
	ESD Voltage, Any Pin ²	2000		V
	Transient Latch-up Current, Any Pin		100	mA
l _{in}	Input Current, Any Digital Pin ³	-10	10	mA
	DC Input Current, Any Analog Pin ³		±100	mA
Pd	Maximum Power Dissipation in Package		1.68 ⁴	W
Тj	Junction Temperature		125	°C
Τ _s	Storage Temperature	-65	+150	°C

Note:

1. Reference to ground.

2. Human body model.

3. Constant input current.

4. If device power consumption exceeds this value, a heatsink or a fan must be used. Refer to Chapter 7 Thermal Management.

Caution:

Exceeding the above values may cause permanent damage. Functional operation under these conditions is not implied. Exposure to absolute maximum rating conditions for extended period may affect device reliability.

8.2 RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter	Min	Тур.	Max	Unit	
T _{op}	Operating Temperature Range	-40		85 ¹	٥°	
VDDIO	Digital I/O Power Supply	3.13	3.3	3.47	V	
VDDA	Analog Core Power Supply	3.13	3.3	3.47	V	
VDDD	Digital Core Power Supply	1.71	1.8	1.89	V	
VDDT	Power Supply for Transmitter Driver	3.13	3.3	3.47	V	
VDDR	Power Supply for Receiver	3.13	3.3	3.47	V	
V _{IL}	Input Low Voltage	-0.5		0.8	V	
V _{IH}	Input High Voltage	2.0		VDDIO+0.5	V	
ote:			ļ	-		

1. An external thermal solution such as heatsink may be required depending on the mode of operation. Refer to Chapter 7 Thermal Management.

8.3 DEVICE POWER CONSUMPTION AND DISSIPATION (TYPICAL)¹

		Total Consumption (W)				Total Device Power Dissipation (for Thermal Consideration, W)			Per-Channel Power Down Saving (mW) ²		
Mode	Parameter	1.8 V	3.3 V	Total	Fully Internal R120IN=1 ³	Partially Internal R120IN=0 ⁴	External ⁵	Fully Internal R120IN=1 ³	Partially Internal R120IN=0 ⁴	External ⁵	
E1/120 Ω	PRBS	0.18	1.73	1.91	1.91	1.48	1.20	80	60	40	
	100% ones	0.18	2.31	2.49	2.49	1.87	1.46	130	90	70	
E1/75 Ω	PRBS	0.18	1.86	2.04	2.04	1.77	0.98	90	60	50	
	100% ones	0.18	2.54	2.72	2.72	2.33	1.32	150	120	80	
T1/100 Ω	QRSS	0.14	1.93	2.07	2.07	1.66	1.57	100	80	60	
	100% ones	0.14	2.79	2.93	2.93	2.32	2.24	160	120	90	
J1/110 Ω	QRSS	0.14	1.87	2.01	2.01	1.59	1.56	100	70	60	
	100% ones	0.14	2.64	2.78	2.78	2.17	2.17	150	110	90	

Note:

1. Test conditions: VDDx (typical) at 25 °C operating temperature (ambient).

2. The R_OFF bit (b5, RCF0,...) and T_OFF bit (b5, TCF0,...) are set to '1' to enable per-channel power down.

3. The transmitter is in Internal Impedance Matching mode and the receiver is in Fully Internal Impedance Matching mode. That is, the R120IN bit (b4, RCF0,...) is set to '1'. And the T_TERM[2:0] bits (b2~0, TCF0,...) and R_TERM[2:0] bits (b2~0, RCF0,...) are set according to different cable conditions.

4. The transmitter is in Internal Impedance Matching mode and the receiver is in Partially Internal Impedance Matching mode. That is, the R120IN bit (b4, RCF0,...) is set to '0'. And the T_TERM[2:0] bits (b2~0, TCF0,...) and R_TERM[2:0] bits (b2~0, RCF0,...) are set according to different cable conditions.

5. For E1 mode, both the transmitter and the receiver are in External Impedance Matching mode. That is, the T_TERM[2:0] bits (b2~0, TCF0,...) are set to '111' and the R_TERM[2:0] bits (b2~0, RCF0,...) are set to '1xx'. For T1/J1 mode, as the transmitter External Impedance Matching mode is not supported, the transmitter is in Internal Impedance Matching mode and the receiver is in Internal Impedance Matching mode. That is, the T_TERM[2:0] bits (b2~0, TCF0,...) are set to '1xx'. For T1/J1 mode, as the transmitter External Impedance Matching mode is not supported, the transmitter is in Internal Impedance Matching mode and the receiver is in Internal Impedance Matching mode. That is, the T_TERM[2:0] bits (b2~0, TCF0,...) are set according to different cable conditions and the R_TERM[2:0] bits (b2~0, RCF0,...) are set to '1xx'.

8.4 DEVICE POWER CONSUMPTION AND DISSIPATION (MAXIMUM)¹

Mode		Total Consumption (W)			Total Device Power Dissipation (for Thermal Consideration, W)			
	Parameter	1.89 V	3.47 V	Total	Fully Internal R120IN=1 ²	Partially Internal R120IN=0 ³	External ⁴	
E1/120 Ω	PRBS	0.22	1.82	2.04	2.04	1.68	1.40	
	100% ones	0.22	2.43	2.65	2.65	2.08	1.67	
E1/75 Ω	PRBS	0.22	1.95	2.17	2.17	1.98	0.98	
	100% ones	0.22	2.67	2.89	2.89	2.53	1.37	
T1/100 Ω	QRSS	0.17	2.03	2.20	2.20	1.84	1.65	
	100% ones	0.17	2.93	3.10	3.10	2.53	2.34	
J1/110 Ω	QRSS	0.17	1.96	2.13	2.13	1.78	1.63	
	100% ones	0.17	2.78	2.95	2.95	2.38	2.28	

Note:

1. Test conditions: VDDx (maximum) at 85 °C operating temperature (ambient).

2. The transmitter is in Internal Impedance Matching mode and the receiver is in Fully Internal Impedance Matching mode. That is, the R120IN bit (b4, RCF0,...) is set to '1'. And the T_TERM[2:0] bits (b2~0, TCF0,...) and R_TERM[2:0] bits (b2~0, RCF0,...) are set according to different cable conditions.

3. The transmitter is in Internal Impedance Matching mode and the receiver is in Partially Internal Impedance Matching mode. That is, the R120IN bit (b4, RCF0,...) is set to '0'. And the T_TERM[2:0] bits (b2~0, TCF0,...) and R_TERM[2:0] bits (b2~0, RCF0,...) are set according to different cable conditions.

4. For E1 mode, both the transmitter and the receiver are in External Impedance Matching mode. That is, the T_TERM[2:0] bits (b2~0, TCF0,...) are set to '111' and the R_TERM[2:0] bits (b2~0, RCF0,...) are set to '1xx'. For T1/J1 mode, as the transmitter External Impedance Matching mode is not supported, the transmitter is in Internal Impedance Matching mode and the receiver is in Internal Impedance Matching mode. That is, the T_TERM[2:0] bits (b2~0, TCF0,...) are set to '1xx'. For T1/J1 mode, as the T_TERM[2:0] bits (b2~0, TCF0,...) are set according to different cable conditions and the R_TERM[2:0] bits (b2~0, RCF0,...) are set to '1xx'.

8.5 D.C. CHARACTERISTICS

@ TA = -40 to +85 °C, VDDIO = 3.3 V ± 5%, VDDD = 1.8 V ± 5%

Symbol	Parameter	Min	Тур.	Max	Unit	Test Conditions
V _{OL}	Output Low Voltage			0.40	V	VDDIO = 3.13 V, I _{OL} = 4 mA, 8 mA
V _{OH}	Output High Voltage	2.4		VDDIO	V	VDDIO = 3.13 V, I _{OH} = 4 mA, 8 mA
V _{T+}	Schmitt Trigger Input Low to High Threshold	1.47			V	
V _{T-}	Schmitt Trigger Input High to Low Threshold			0.89	V	
R _{pu}	Internal Pull-up /Pull-down Resistor	50	70	115	KΩ	
IIL	Input Low Current	-1	0	+1	μA	V _{IL} = GNDD
I _{IH}	Input High Current	-1	0	+1	μA	V _{IH} = VDDIO
C _{in}	Input Digital Pin Capacitance			10	pF	
C _{out}	Output Load Capacitance			50	pF	
C _{out}	Output Load Capacitance (bus pins)			100	pF	
I _{ZL}	Leakage Current of Digital Output in High-Z mode	-10		10	μA	GNDIO < V _O < VDDIO
Z _{OH}	Output High-Z on TTIPn, TRINGn pins	10			KΩ	



8.6 E1 RECEIVER ELECTRICAL CHARACTERISTICS

Paran	neter	Min	Тур.	Мах	Unit	Test Conditions
Receiver Sensitivity of Receive Differen- tial mode with Cable Loss @ 1024 KHz			15		dB	with Nominal Pulse Amplitude of 3.0 V for 120 Ω
Receiver Sensitivity of Receive Single Ended mode with Cable Loss @ 1024 kHz			12		dB	and 2.37 V for 75 $Ω$ termination, adding -18 dB interference signal.
Signal to Noise Interfe	erence Margin	-16			dB	@cable loss 0-6 dB
Analog LOS Level (Normal Mode)	ALOS[2:0] 000 001 (default) 010 011 100 101 110 111		0.5 0.7 0.9 1.2 1.4 1.6 1.8 2.0		V _{pp}	In Differential mode, measured between RTIP and RRING pins. In Singled Ended mode, measured between RTIP and GNDA pins Refer to Table-17 for LLOS Criteria Declare and Clear.
	LOS hysteresis		0.25			
Analog LOS Level (Line Monitor Mode)	ALOS[2:0] 000 001 (default) 010 011 1xx (reserved)		1.0 1.4 1.8 2.2		V _{pp}	Measured on the line with the monitor gain set by the MG[1:0] bits (b1~0, RCF2,) equal to the resistive attenuation. Refer to Table-17 for LLOS Criteria Declare and Clear.
	LOS hysteresis		0.41			
Allowable Consecutiv G.775 I.431 / ETSI300233			32 2048			
LOS Reset		12.5			% ones	G.775, ETSI 300233
Receive Intrinsic Jitte	r			0.05	U.I.	JA disabled; wide band
Input Jitter Tolerance: 1 Hz ~ 20 Hz 20 Hz ~ 2.4 KHz 18 KHz ~ 100 KHz		80 10 0.6 Refer to Fig- ure-41			U.I. U.I. U.I.	G.823, with 6 dB Cable Attenuation
Receiver Differential	Input Impedance		2.6		KΩ	
Receiver Common ance to GND	Mode Input Imped-		1.6		KΩ	— @1024 KHz; Rx port is high-Z

Parameter	Min	Тур.	Max	Unit	Test Conditions
Receiver Single Ended mode Input Impedance to GND		3.1		KΩ	The RRINGn pins are open.
Receive Return Loss: 51 KHz ~ 102 KHz 102 KHz ~ 2.048 MHz 2.048 MHz ~ 3.072 MHz	12 18 14			dB dB dB	G.703
Receive Path Delay: Single Rail Dual Rail NRZ Dual Rail RZ			6.6 1.8 1.5	U.I. U.I. U.I.	JA Disabled

8.7 T1/J1 RECEIVER ELECTRICAL CHARACTERISTICS

Param	neter	Min	Тур.	Max	Unit	Test Conditions
Receiver Sensitivity c tial mode with Cable L			15		dB	with Nominal Pulse Amplitude of 3.0 V for 100 Ω ter-
Receiver Sensitivity of Receive Single Ended mode with Cable Loss @ 772 KHz Signal to Noise Interference Margin			12		dB	mination, adding -18 dB interference signal.
Signal to Noise Interfe	erence Margin	-16			dB	
Analog LOS Level (Normal Mode)	ALOS[2:0] 000 001 (default) 010 011 100 101 110 111 LOS hysteresis		0.5 0.7 0.9 1.2 1.4 1.6 1.8 2.0 0.25		V _{pp}	In Differential mode, measured between RTIP and RRING pins. In Singled Ended mode, measured between RTIP and GNDA pins Refer to Table-17 for LLOS Criteria Declare and Clear.
Analog LOS Level (Line Monitor Mode)	ALOS[2:0] 000 001 (default) 010 011 1xx (reserved)		1.0 1.4 1.8 2.2		V _{pp}	Measured on the line with the monitor gain set by the MG[1:0] bits (b1~0, RCF2,) equal to the resistive attenuation. Refer to Table-17 for LLOS Criteria Declare and Clear.
	LOS hysteresis		0.41		-	
Allowable Consecutive T1.231 - 1993 I.431	e Zeros before LOS:		175 1544			
LOS Reset		12.5			% ones	G.775, ETSI 300233
Receive Intrinsic Jitter				0.05	U.I.	JA disabled; Wide band
Input Jitter Tolerance: 0.1 Hz ~ 1 Hz 4.9 Hz ~ 300 Hz 10 KHz ~ 100 KHz		138.0 28.0 0.4			U.I. U.I. U.I.	AT&T62411
Receiver Differential In	nput Impedance		3.1		KΩ	
Receiver Common M ance to GND	Mode Input Imped-		2.2		KΩ	→ @772 KHz; Rx port is high-Z
Receiver Single Er Impedance to GND	nded mode Input		4		KΩ	The RRINGn pins are open.
Receive Return Loss: 39 KHz ~ 77 KHz 77 KHz ~ 1.544 MH 1.544 MHz ~ 2.316	lz	20 20 20			dB dB dB	G.703
Receive Path Delay: Single Rail Dual Rail NRZ Dual Rail RZ			6.5 2.5 1.4		U.I. U.I. U.I.	JA Disabled

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8.8 E1 TRANSMITTER ELECTRICAL CHARACTERISTICS

Parameter		Min	Тур.	Max	Unit	Test Conditions
Output Pulse Amplitude: E1, 75 Ω load E1, 120 Ω load		2.14 2.7	2.37 3.0	2.60 3.3	V V	Differential Line Inter- face mode
Zero (Space) Level: E1, 75 Ω load E1, 120 Ω load		-0.237 -0.3		+0.237 0.3	V V	Differential Line Inter- face mode
Transmit Amplitude Variation with Supply		-1		+1	%	
Difference between Pulse Sequences for 17 consecutive pulses (T1.102)				200	mV	
Output Pulse Width at 50% of Nominal	Amplitude	232	244	256	ns	
Ratio of the Amplitudes of Positive and Negative Pulses at the Center of the Pulse Interval (G.703)		0.95		1.05		
Ratio of the Width of Positive and Negative Pulses at the Center of the Pulse Interval (G.703)		0.95		1.05		
Transmit Analog LOS Level (TALOS) (Differential line interface)	TALOS[1:0] 00 01 (default) 10 11		1.2 0.9 0.6 0.4		V _p	Measured on the TTIP and TRING pins.
Transmit Analog LOS Level (TALOS) (Single Ended line interface)	TALOS hysteresis TALOS[1:0] 00		0.08		V _p Meas pin.	Measured on the TTIP
	01 (default) 10 11		0.48 0.32 0.24			pin.
	TALOS hysteresis		0.04			
Transmit Return Loss (G.703): 51 KHz ~ 102 KHz 102 KHz ~ 2.048 MHz 2.048 MHz ~ 3.072 MHz		8 14 10			dB dB dB	Internal Impedance Matching
Intrinsic Transmit Jitter 20 Hz ~ 100 KHz				0.050	U.I.	TCLK is jitter free
Transmit Path Delay: Single Rail Dual Rail NRZ Dual Rail RZ			8.5 4.5 4.4		U.I. U.I. U.I.	JA is disabled
Line Short Circuit Current			100	1	mAp	Measured on pin

8.9 T1/J1 TRANSMITTER ELECTRICAL CHARACTERISTICS

Parameter		Min	Тур.	Max	Unit	Test Conditions
Output Pulse Amplitude		2.4	3.0	3.6	V	Differential Line Interface mode
Zero (Space) Level		-0.15		0.15	V	
Transmit Amplitude Variation with Supp	ly	-1		+1	%	
Difference between Pulse Sequences fr (T1.102)	or 17 consecutive pulses			200	mV	
Output Pulse Width at 50% of Nominal	Amplitude	338	350	362	ns	
Pulse Width Variation at the Half Amplitude (T1.102)				20	ns	
Imbalance between Positive and Negative Pulses Amplitude (T1.102)		0.95		1.05		
Output Power Levels (T1.102-1993): @772 KHz @1544 KHz (Referenced to Power at 772 KHz)		12.6 -29		17.9	dBm dBm	
Transmit Analog LOS Level (TALOS) (Differential line interface)	TALOS[1:0] 00 01 (default) 10 11 TALOS hysteresis		1.2 0.9 0.6 0.4 0.08		Vp	Measured on the TTIP and TRING pins.
Transmit Analog LOS Level (TALOS) (Single Ended line interface)	TALOS[1:0] 00 01 (default) 10 11 TALOS hysteresis		0.61 0.48 0.32 0.24 0.04		Vp	Measured on the TTIP pin.
Transmit Return Loss (G.703): 39 KHz ~ 77 KHz 77 KHz ~ 1.544 MHz 1.544 MHz ~ 2.316 MHz	TALOS Hysteresis	8 14 10	0.04		dB dB dB	Internal Impedance Matching
Intrinsic Transmit Jitter: 10 Hz ~ 8 KHz 8 KHz ~ 40 KHz 10 Hz ~ 40 KHz Wide Band				0.020 0.025 0.025 0.050	U.I.p-p U.I.p-p U.I.p-p U.I.p-p	TCLK is jitter free
Transmit Path Delay (JA is disabled): Single Rail Dual Rail NRZ Dual Rail RZ			8.2 4.1 4.3		U.I. U.I. U.I.	JA is disabled
Line Short Circuit Current			100		mAp	Measure on pin

8.10 TRANSMITTER AND RECEIVER TIMING CHARACTERISTICS

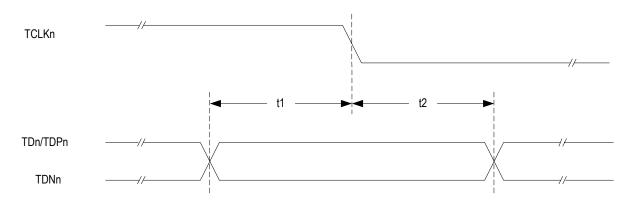
Symbol	Parameter	Min	Тур.	Max	Unit
	MCLK Frequency: E1 T1/J1		2.048 X n 1.544 X n (n = 1 ~ 8)		MHz MHz
	MCLK Tolerance	-100		100	ppm
	MCLK Duty Cycle	30		70	%
ransmit Path			·		
	TCLK Frequency: E1 T1/J1		2.048 1.544		MHz MHz
	TCLK Tolerance	-50		+50	ppm
	TCLK Duty Cycle	10		90	%
t1	Transmit Data Setup Time	40			ns
t2	Transmit Data Hold Time	40			ns
	Delay Time of OE low to Driver High-Z			1	μs
Receive Path			· · ·		
	Clock Recovery Capture Range ¹ : E1 T1/J1		+80 / -80 +180 / -180		ppm ppm
	RCLK Duty Cycle ²	40	50	60	%
t4	RCLK Pulse Width ² : E1 T1/J1	457 607	488 648	519 689	ns ns
t5	RCLK Pulse Width Low Time: E1 T1/J1	203 259	244 324	285 389	ns ns
t6	RCLK Pulse Width High Time: E1 T1/J1	203 259	244 324	285 389	ns ns
	Rise/Fall Time ³	20			ns
t7	Receive Data Setup Time: E1 T1/J1	200 200	244 324		ns ns
t8	Receive Data Hold Time: E1 T1/J1	200 200	244 324		ns ns

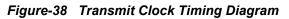
Note:

1. Relative to nominal frequency, MCLK = +100 or -100 ppm.

2. RCLK duty cycle width will vary depending on extent of the received pulse jitter displacement. Maximum and minimum RCLK duty cycles are for worst case jitter conditions (0.2 UI displacement for E1 per ITU G.823).

3. For all digital outputs. C_{load} = 15 pF.





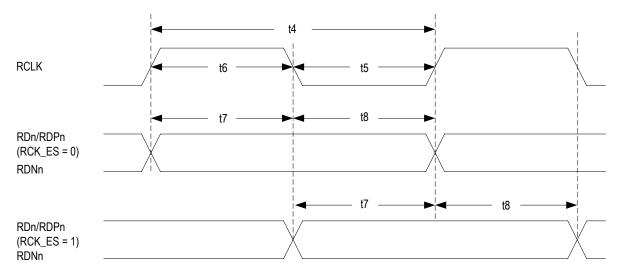


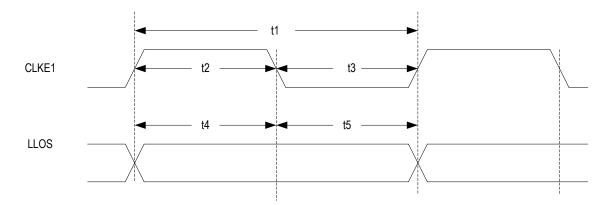
Figure-39 Receive Clock Timing Diagram

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8.11 CLKE1 TIMING CHARACTERISTICS

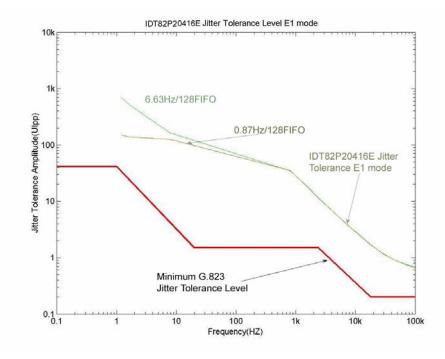
Symbol	Parameter	Min	Тур.	Max	Unit
CLKE1 outputs 2.04	8 MHz clock				
t1	CLKE1 Pulse Width		488		ns
t2	CLKE1 Pulse Width High Time	232	244	256	ns
t3	CLKE1 Pulse Width Low Time	232	244	256	ns
t4	LLOS Data Setup Time	217	244	271	ns
t5	LLOS Data Hold Time	217	244	271	ns
CLKE1 outputs 8kH	z clock				
t1	CLKE1 Pulse Width		125		μs
t2	CLKE1 Pulse Width High Time	62.4	62.5	62.6	μs
t3	CLKE1 Pulse Width Low Time	62.4	62.5	62.6	μs
t4	LLOS Data Setup Time	62.38	62.5	62.62	μs
t5	LLOS Data Hold Time	62.38	62.5	62.62	μs





8.12 JITTER ATTENUATION CHARACTERISTICS

Par	ameter	Min	Тур.	Мах	Unit
Jitter Transfer Function Corner (-3 dB) F	requency:				
E1, 32/64/128-bit FIFO	JA_BW = 0 JA_BW = 1 JA_BW = 0 JA_BW = 1		6.63 0.87 5 1.28		Hz Hz Hz Hz
Jitter Attenuator:					
E1 (G.736)	@ 3 Hz @ 40 Hz @ 400 Hz @ 100 KHz	-0.5 -0.5 +19.5 +19.5			dB dB dB dB
T1/J1 (AT&T pub.62411)	@ 1 Hz @ 20 Hz @ 1 KHz @ 1.4 KHz @ 70 KHz	0 0 +33.3 40 40			dB dB dB dB dB
Jitter Attenuator Latency Delay: 32-bit FIFO 64-bit FIFO 128-bit FIFO			16 32 64		U.I. U.I. U.I.
Input Jitter Tolerance before FIFO O Underflow: 32-bit FIFO 64-bit FIFO 128-bit FIFO	verflow or		28 56 120		U.I. U.I. U.I.





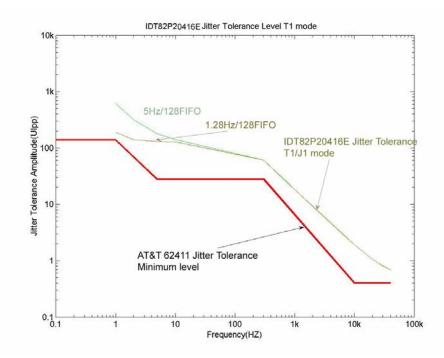
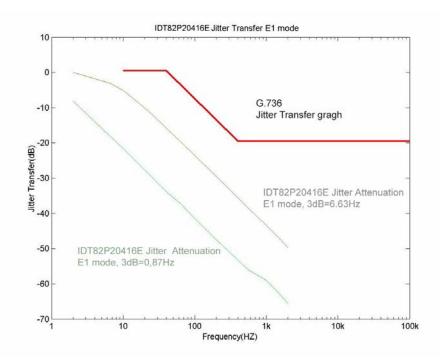


Figure-42 T1/J1 Jitter Tolerance Performance





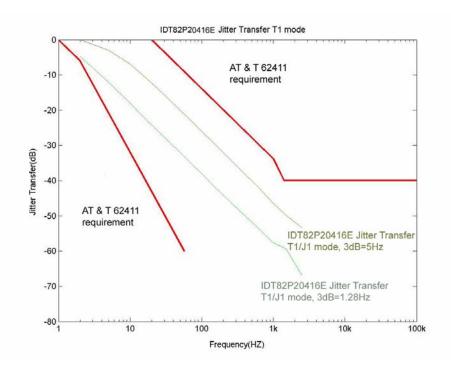


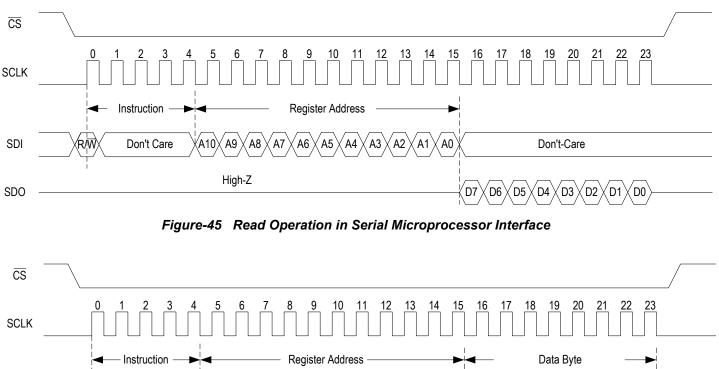
Figure-44 T1/J1 Jitter Transfer Performance

8.13 MICROPROCESSOR INTERFACE TIMING

8.13.1 SERIAL MICROPROCESSOR INTERFACE

A falling transition on CS indicates the start of a read/write operation, and a rising transition indicates the end of the operation. After \overline{CS} is set to low, a 5-bit instruction on SDI is input to the device on the rising edge of SCLK. If the MSB is '1', it is a read operation. If the MSB is '0', it is a

write operation. Following the instruction, an 11-bit address is clocked in on SDI to specify the register. If the device is in a read operation, the data read from the specified register is output on SDO on the falling edge of SCLK (refer to Figure-45). If the device is in a write operation, the data written to the specified register is input on SDI following the address byte (refer to Figure-46).



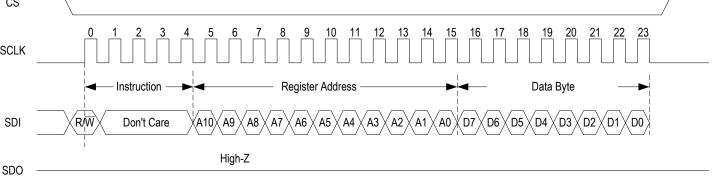


Figure-46 Write Operation in Serial Microprocessor Interface

Symbol	Description	Min.	Max.	Units
f _{OP}	SCLK Frequency		2.0	MHz
t _{CSH}	Minimum CS High Time	100		ns
t _{CSS}	CS Setup Time	50		ns
t _{CSD}	CS Hold Time	100		ns
t _{CLD}	Clock Disable Time	50		ns
t _{CLH}	Clock High Time	205		ns
t _{CLL}	Clock Low Time	205		ns
t _{DIS}	Data Setup Time	50		ns
t _{DIH}	Data Hold Time	150		ns
t _{PD}	Output Delay		150	ns
t _{DF}	Output Disable Time		50	ns

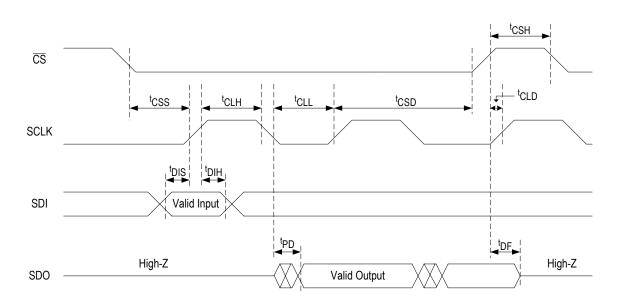
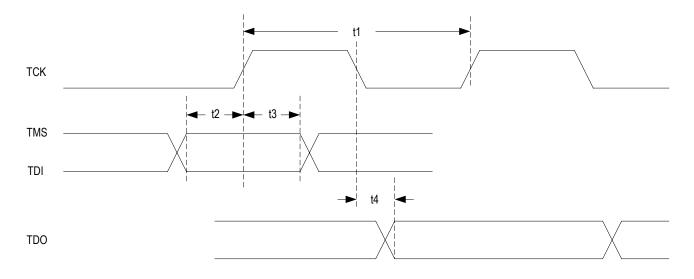


Figure-47 Timing Diagram

8.14 JTAG TIMING CHARACTERISTICS

Symbol	Parameter	Min	Тур.	Мах	Unit
t1	TCK Period	100			ns
t2	TMS to TCK Setup Time; TDI to TCK Setup Time	25			ns
t3	TCK to TMS Hold Time; TCK to TDI Hold Time	25			ns
t4	TCK to TDO Delay Time			50	ns





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Glossary

AIS	_	Alarm Indication Signal
AMI	_	Alternate Mark Inversion
ARB	_	Arbitrary Pattern
B8ZS	_	Binary 8 Zero Substitution
BPV	_	Bipolar Violation
CF	_	Corner Frequency
cv	_	Code Violation
DPLL	_	Digital Phase Locked Loop
EXZ	_	Excessive Zeroes
FIFO	_	First In First Out
HDB3	_	High Density Bipolar 3
HPS	_	Hitless Protection Switching
IB	_	Inband Loopback
LAIS	_	Line Alarm Indication Signal
LBPV	_	Line Bipolar Violation
LEXZ	_	Line Excessive Zeroes
LLOS	_	Line Loss of Signal
LOS	_	Loss Of Signal
NRZ	_	Non-Return to Zero
РВХ	_	Private Branch Exchange
PRBS	_	Pseudo Random Bit Sequence
QRSS	_	Quasi-Random Signal Source
RJA	_	Receive Jitter Attenuator
RZ	_	Return to Zero
SAIS	_	System Alarm Indication Signal
SBPV	_	System Bipolar Violation
SDH	_	Synchronous Digital Hierarchy

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SEXZ	_	System Excessive Zeroes
SLOS	_	System LOS
SONET	_	Synchronous Optical Network
TEPBGA	_	Thermally Enhanced Plastic Ball Grid Array
TJA	_	Transmit Jitter Attenuator
TLOS	_	Transmit Loss of Signal
TOC	_	Transmit Over Current

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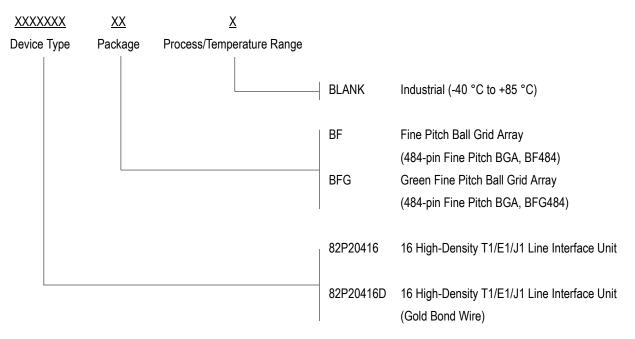
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