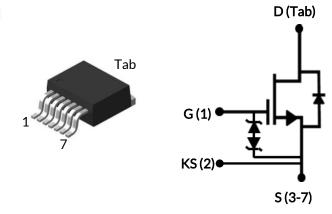


### $750V-33m\Omega$ SiC FET

Rev. B, March 2022

#### DATASHEET

# UJ4C075033B7S



Part NumberPackageMarkingUJ4C075033B7SD²PAK-7LUJ4C075033B7S



#### Description

The UJ4C075033B7S is a 750V,  $33m\Omega$  G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows for a true "drop-in replacement" to Si IGBTs, Si FETs, SiC MOSFETs or Si superjunction devices. Available in the D<sup>2</sup>PAK-7L package, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

#### Features

- On-resistance R<sub>DS(on)</sub>: 33mΩ (typ)
- Operating temperature: 175°C (max)
- Excellent reverse recovery: Q<sub>rr</sub> = 62nC
- Low body diode V<sub>FSD</sub>: 1.26V
- Low gate charge: Q<sub>G</sub> = 37.8nC
- Threshold voltage V<sub>G(th)</sub>: 4.8V (typ) allowing 0 to 15V drive
- Low intrinsic capacitance
- ESD protected: HBM class 2 and CDM class C3
- D<sup>2</sup>PAK-7L package for faster switching, clean gate waveforms

#### Typical applications

- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Motor drives
- Induction heating





#### **Maximum Ratings**

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	V <sub>DS</sub>		750	V
Cata aquiraquialtaga	V	DC	-20 to +20	V
Gate-source voltage	V <sub>GS</sub>	AC (f > 1Hz)	-25 to +25	V
<b>a</b>	1	T <sub>C</sub> = 25°C	44	А
Continuous drain current <sup>1</sup>	I <sub>D</sub>	T <sub>C</sub> =100°C	32.5	А
Pulsed drain current <sup>2</sup>	I <sub>DM</sub>	T <sub>C</sub> = 25°C	140	А
Single pulsed avalanche energy <sup>3</sup>	E <sub>AS</sub>	L=15mH, I <sub>AS</sub> =2.4A	43	mJ
SiC FET dv/dt ruggedness	dv/dt	$V_{DS} \le 500V$	200	V/ns
Power dissipation	P <sub>tot</sub>	T <sub>C</sub> = 25°C	197	W
Maximum junction temperature	T <sub>J,max</sub>		175	°C
Operating and storage temperature	T <sub>J</sub> , T <sub>STG</sub>		-55 to 175	°C
Reflow soldering temperature	$T_{solder}$	reflow MSL 1	245	°C

1. Limited by  $T_{\text{J},\text{max}}$ 

2. Pulse width  $t_p$  limited by  $T_{J,max}$ 

3. Starting  $T_J = 25^{\circ}C$ 

Thermal Characteristics

Parameter	Symbol	Test Conditions	Value			Units
			Min	Тур	Max	Onits
Thermal resistance, junction-to-case	$R_{ ext{ heta}JC}$			0.58	0.76	°C/W





#### Electrical Characteristics (T<sub>J</sub> = +25°C unless otherwise specified)

#### **Typical Performance - Static**

Parameter	Symbol	Test Conditions	Value			11.21.
			Min	Тур	Max	- Units
Drain-source breakdown voltage	BV <sub>DS</sub>	V <sub>GS</sub> =0V, I <sub>D</sub> =1mA	750			V
Total drain leakage current		V <sub>DS</sub> =750V, V <sub>GS</sub> =0V, T <sub>J</sub> =25°C		2	20	- μΑ
	I <sub>DSS</sub>	V <sub>DS</sub> =750V, V <sub>GS</sub> =0V, T <sub>J</sub> =175°C		20		
Total gate leakage current	I <sub>GSS</sub>	V <sub>DS</sub> =0V, T <sub>J</sub> =25°C, V <sub>GS</sub> =-20V / +20V		6	±20	μA
Drain-source on-resistance		V <sub>GS</sub> =12V, I <sub>D</sub> =30A, T <sub>J</sub> =25°C		33	41	
	R <sub>DS(on)</sub>	V <sub>GS</sub> =12V, I <sub>D</sub> =30A, T <sub>J</sub> =125°C		57		mΩ
		V <sub>GS</sub> =12V, I <sub>D</sub> =30A, T <sub>J</sub> =175°C		75		
Gate threshold voltage	V <sub>G(th)</sub>	$V_{DS}$ =5V, $I_{D}$ =10mA	4	4.8	6	V
Gate resistance	R <sub>G</sub>	f=1MHz, open drain		4.5		Ω

#### Typical Performance - Reverse Diode

Parameter	Symbol	Test Conditions		Units		
			Min	Тур	Max	Onits
Diode continuous forward current <sup>1</sup>	ا <sub>s</sub>	T <sub>C</sub> = 25°C			44	А
Diode pulse current <sup>2</sup>	I <sub>S,pulse</sub>	T <sub>C</sub> = 25°C			140	А
Forward voltage	V <sub>FSD</sub>	V <sub>GS</sub> =0V, I <sub>S</sub> =15A, T <sub>J</sub> =25°C		1.26	1.42	V
		V <sub>GS</sub> =0V, I <sub>S</sub> =15A, T <sub>J</sub> =175°C		1.59		- V
Reverse recovery charge	Q <sub>rr</sub>	$V_R$ =400V, I <sub>S</sub> =30A, $V_{GS}$ =0V, R <sub>G_EXT</sub> =50 $\Omega$		62		nC
Reverse recovery time	t <sub>rr</sub>	di/dt=1100A/μs, T_=25°C		11.2		ns
Reverse recovery charge	Q <sub>rr</sub>	$V_{R}$ =400V, I <sub>S</sub> =30A, $V_{GS}$ =0V, R <sub>G_EXT</sub> =50 $\Omega$		67		nC
Reverse recovery time	t <sub>rr</sub>	di/dt=1100A/µs, Tj=150°C		11.5		ns





#### Typical Performance - Dynamic

Parameter	Symbol	Test Conditions	Value			Units
			Min	Тур	Max	- Units
Input capacitance	C <sub>iss</sub>			1400		
Output capacitance	C <sub>oss</sub>	- V <sub>DS</sub> =400V, V <sub>GS</sub> =0V - f=100kHz -		68		pF
Reverse transfer capacitance	C <sub>rss</sub>	1-100KHZ		2.5		
Effective output capacitance, energy related	C <sub>oss(er)</sub>	$V_{DS}=0V$ to 400V, $V_{GS}=0V$		83		pF
Effective output capacitance, time related	C <sub>oss(tr)</sub>	V <sub>DS</sub> =0V to 400V, V <sub>GS</sub> =0V		162		pF
C <sub>OSS</sub> stored energy	E <sub>oss</sub>	V <sub>DS</sub> =400V, V <sub>GS</sub> =0V		6.6		μJ
Total gate charge	Q <sub>G</sub>	N/ 400X/1 20A		37.8		
Gate-drain charge	$Q_{GD}$	$V_{DS} = 400V, I_{D} = 30A,$ - $V_{GS} = 0V \text{ to } 15V$		8		nC
Gate-source charge	Q <sub>GS</sub>	$V_{GS} = 0V (0.15V)$		11.8		
Turn-on delay time	t <sub>d(on)</sub>	Notes 4,		11		
Rise time	t <sub>r</sub>	$V_{DS}$ =400V, $I_D$ =30A, Gate Driver =0V to +15V, Turn-on $R_{GEXT}$ =1 $\Omega$ ,		23		
Turn-off delay time	$t_{d(off)}$			157		  μJ
Fall time	t <sub>f</sub>	Turn-off $R_{G,EXT}$ =50 $\Omega$ ,		14		
Turn-on energy	E <sub>ON</sub>	inductive Load, FWD: same device with		146		
Turn-off energy	E <sub>OFF</sub>	$V_{GS} = 0V$ and $R_G = 50\Omega$ ,		96		
Total switching energy	E <sub>TOTAL</sub>	T <sub>J</sub> =25°C		242		
Turn-on delay time	t <sub>d(on)</sub>	Notes 4,		12		
Rise time	t <sub>r</sub>	V <sub>DS</sub> =400V, I <sub>D</sub> =30A, Gate		23		
Turn-off delay time	$t_{d(off)}$	- Driver =0V to +15V, Turn-on $R_{G,EXT}$ =1 $\Omega$ ,		165		ns
Fall time	t <sub>f</sub>	Turn-off $R_{G,EXT}$ =50 $\Omega$ ,		15		
Turn-on energy	E <sub>ON</sub>	inductive Load, FWD: same device with $V_{GS} = 0V$ and $R_G = 50\Omega$ , $T_J=150^{\circ}C$		160		
Turn-off energy	E <sub>OFF</sub>			116		μJ
Total switching energy	E <sub>TOTAL</sub>			276		

4. Measured with the switching test circuit in Figure 23





#### Typical Performance - Dynamic (continued)

Parameter	Symbol	Test Conditions -	Value			Linte
			Min	Тур	Max	- Units
Turn-on delay time	t <sub>d(on)</sub>	Notes 5 and 6.		11		- ns
Rise time	t <sub>r</sub>	V <sub>DS</sub> =400V, I <sub>D</sub> =30A, Gate		23		
Turn-off delay time	t <sub>d(off)</sub>	Driver =0V to +15V, Turn-on $R_{G,EXT}$ =1 $\Omega$ ,		42		
Fall time	t <sub>f</sub>	Turn-off $R_{G,EXT} = 5\Omega$ ,		7.2		
Turn-on energy including R <sub>s</sub> energy	E <sub>ON</sub>	inductive Load, FWD: same device with $V_{GS} = 0V$ and $R_G = 5\Omega$ ,		144		μ
Turn-off energy including R <sub>s</sub> energy	E <sub>OFF</sub>			25		
Total switching energy	E <sub>TOTAL</sub>	RC snubber: $R_s = 10\Omega$ and		169		
Snubber R <sub>s</sub> energy during turn-on	E <sub>RS_ON</sub>	С <sub>s</sub> =100pF, Т_=25°С		0.6		
Snubber R <sub>s</sub> energy during turn-off	E <sub>RS_OFF</sub>			1.4		
Turn-on delay time	t <sub>d(on)</sub>	$\begin{tabular}{ c c c c } \hline Notes 5 and 6, & & & \\ \hline V_{DS}=400V, I_D=30A, Gate & \\ \hline Driver =0V to +15V, & \\ \hline Turn-on R_{G,EXT}=1\Omega, & \\ \hline Turn-off R_{G,EXT}=5\Omega, & \\ \hline inductive Load, & \\ \hline FWD: same device with & \\ \hline V_{GS}=0V and R_G=5\Omega, & \\ \hline RC snubber: R_S=10\Omega and & \\ \hline C_S=100pF, & \\ \hline T_J=150^\circ C & \\ \hline \end{tabular}$		14		– ns
Rise time	t <sub>r</sub>			23		
Turn-off delay time	t <sub>d(off)</sub>			45		
Fall time	t <sub>f</sub>			9		-
Turn-on energy including R <sub>s</sub> energy	E <sub>ON</sub>			157		
Turn-off energy including R <sub>s</sub> energy	E <sub>OFF</sub>			29		
Total switching energy	E <sub>TOTAL</sub>			186		μJ
Snubber R <sub>s</sub> energy during turn-on	E <sub>RS_ON</sub>			0.6		
Snubber R <sub>s</sub> energy during turn-off	E <sub>RS_OFF</sub>			1.4		

5. Measured with the switching test circuit in Figure 24.

6. In this datasheet, all the switching energies (turn-on energy, turn-off energy and total energy) presented in the tables and Figures include the device RC snubber energy losses.



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#### Typical Performance Diagrams

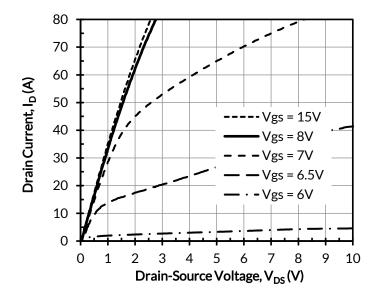


Figure 1. Typical output characteristics at T\_J = - 55°C, tp < 250 $\mu$ s

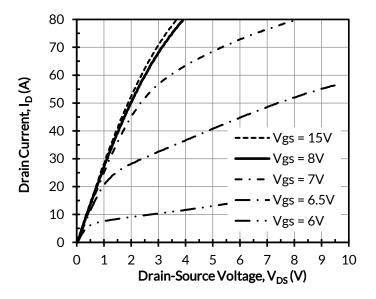


Figure 2. Typical output characteristics at T  $_{\rm J}$  = 25°C, tp < 250 $\mu s$ 

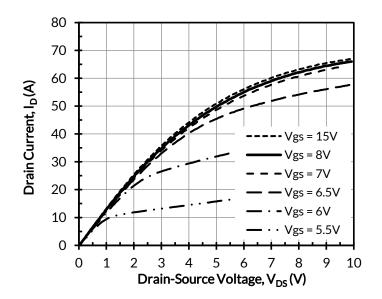


Figure 3. Typical output characteristics at T  $_{\rm J}$  = 175°C, tp < 250 $\mu s$ 

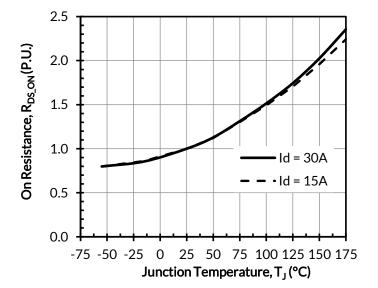


Figure 4. Normalized on-resistance vs. temperature at  $V_{GS}$  = 12V

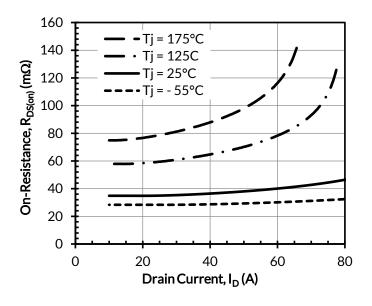
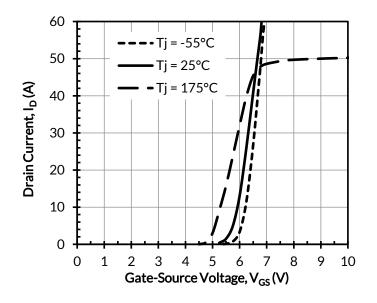


Figure 5. Typical drain-source on-resistances at  $V_{GS}$  = 12V



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Figure 6. Typical transfer characteristics at  $V_{DS}$  = 5V

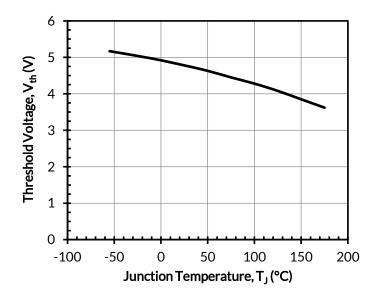


Figure 7. Threshold voltage vs. junction temperature at  $V_{\text{DS}}$  = 5V and  $I_{\text{D}}$  = 10mA

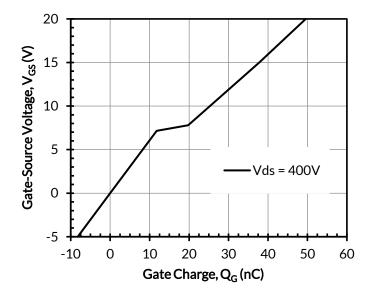


Figure 8. Typical gate charge at  $I_D$  = 30A

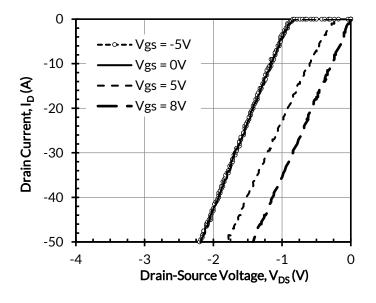
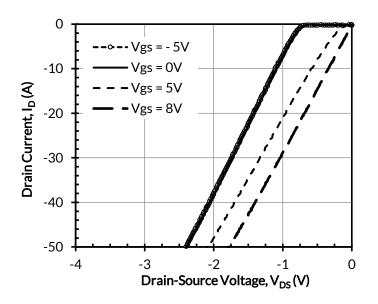


Figure 9. 3rd quadrant characteristics at  $T_J = -55^{\circ}C$ 



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Figure 10. 3rd quadrant characteristics at  $T_J = 25^{\circ}C$ 

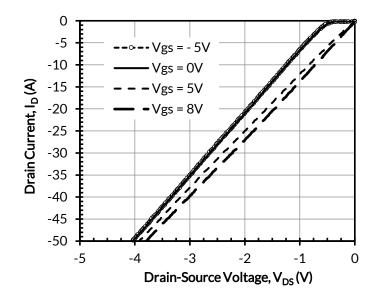


Figure 11. 3rd quadrant characteristics at T<sub>J</sub> = 175°C

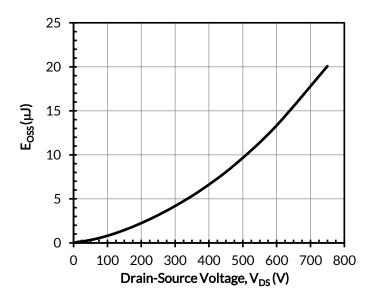


Figure 12. Typical stored energy in  $C_{OSS}$  at  $V_{GS}$  = 0V

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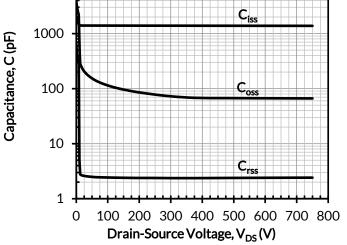
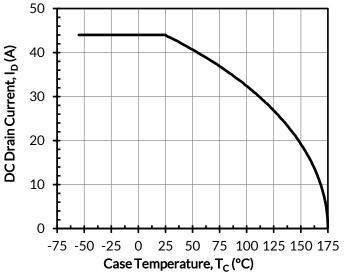


Figure 13. Typical capacitances at f = 100kHz and  $V_{GS}$  = 0V



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Figure 14. DC drain current derating

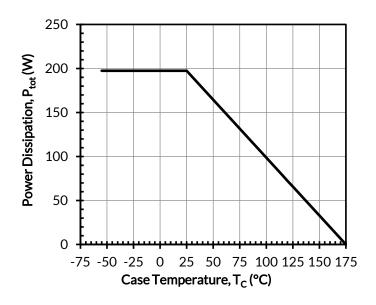


Figure 15. Total power dissipation

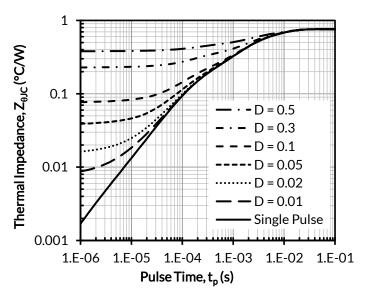


Figure 16. Maximum transient thermal impedance

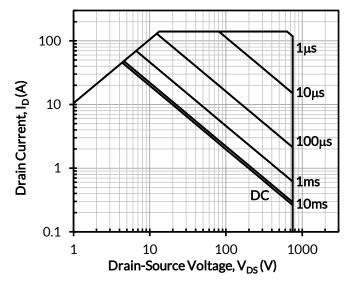
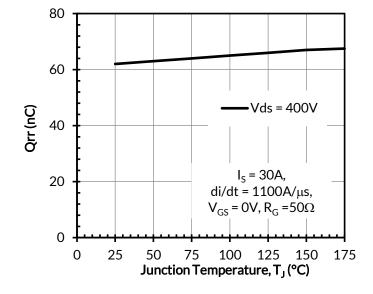


Figure 17. Safe operation area at  $T_C$  = 25°C, D = 0, Parameter  $t_p$ 



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Figure 18. Reverse recovery charge Qrr vs. junction temperature

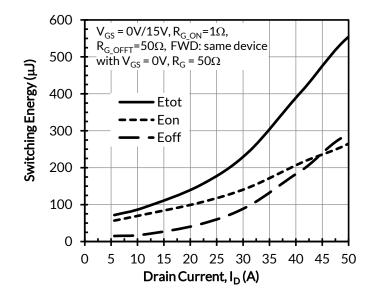


Figure 19. Clamped inductive switching energy vs. drain current at  $V_{DS}$  = 400V and  $T_J$  = 25°C

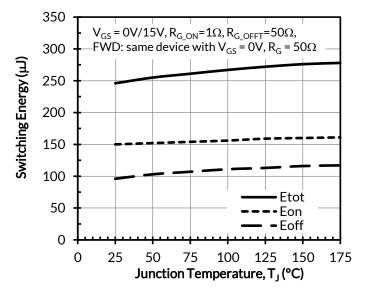


Figure 20. Clamped inductive switching energy vs. junction temperature at  $V_{DS}$  =400V and  $I_D$  = 30A

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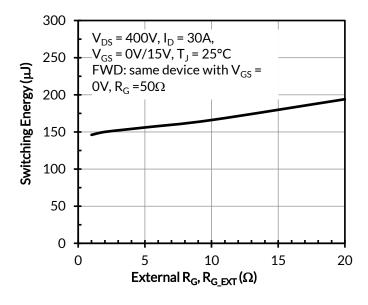
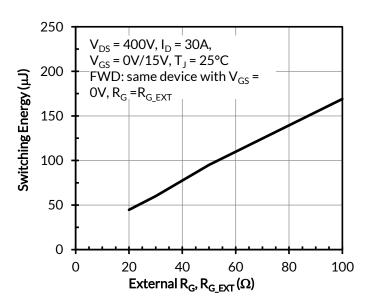


Figure 21. Clamped inductive switching turn-on energy vs.  $R_{G,\text{EXT\_ON}}$ 



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Figure 22. Clamped inductive switching turn-off energy vs.  $R_{G,\text{EXT\_OFF}}$ 

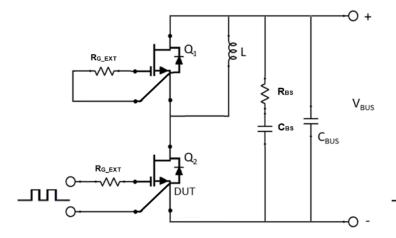


Figure 23. Schematic of the half-bridge mode switching test circuit. Note, a bus RC snubber ( $R_{BS} = 2.5\Omega$ ,  $C_{BS}=100$ nF) is used to reduce the power loop high frequency oscillations.

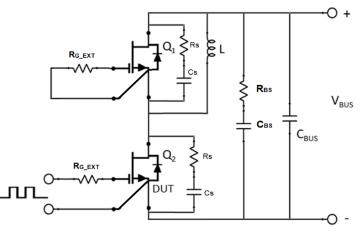


Figure 24. Schematic of the half-bridge mode switching test circuit with device RC snubbers ( $R_s = 10\Omega$ ,  $C_s = 100$ pF) and a bus RC snubber ( $R_{BS} = 2.5\Omega$ ,  $C_{BS} = 100$ nF).







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#### **Applications Information**

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ( $R_{DS(on)}$ ), output capacitance ( $C_{oss}$ ), gate charge ( $Q_G$ ), and reverse recovery charge ( $Q_{rr}$ ) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.unitedsic.com.

A snubber circuit with a small  $R_{(G)}$ , or gate resistor, provides better EMI suppression with higher efficiency compared to using a high  $R_{(G)}$  value. There is no extra gate delay time when using the snubber circuitry, and a small  $R_{(G)}$  will better control both the turn-off  $V_{(DS)}$  peak spike and ringing duration, while a high  $R_{(G)}$  will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high  $R_{(G)}$ , while greatly reducing  $E_{(OFF)}$  from mid-to-full load range with only a small increase in  $E_{(ON)}$ . Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the UnitedSiC website at www.unitedsic.com

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